

Ultra-low-power Arm[®]Cortex[®]-M4 32-bit MCU+FPU, 150 DMIPS, 1-MB Flash memory, 320-KB SRAM, LCD-TFT, AES+PKA

Datasheet - production data

Features

Ultra-low-power with FlexPowerControl

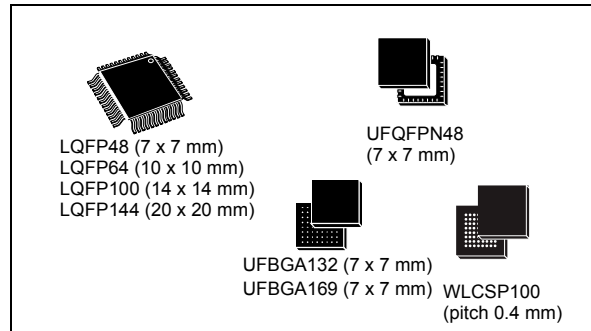
- 1.71 V to 3.6 V power supply
- -40 °C to 85/125 °C temperature range
- Batch acquisition mode (BAM)
- 150 nA in VBAT mode: supply for RTC and 32x32-bit backup registers
- 22 nA Shutdown mode (5 wakeup pins)
- 42 nA Standby mode (5 wakeup pins)
- 190 nA Standby mode with RTC
- 2.95 µA Stop 2 with RTC
- 110 µA/MHz Run mode
- 5 µs wakeup from Stop mode
- Brownout reset (BOR) in all modes except Shutdown

Core

- Arm[®] 32-bit Cortex[®]-M4 CPU with FPU, adaptive real-time accelerator (ART Accelerator) allowing 0-wait-state execution from Flash memory, frequency up to 120 MHz, MPU, 150 DMIPS/1.25 DMIPS/MHz (Dhrystone 2.1), and DSP instructions

Memories

- 1-Mbyte Flash memory, 2 banks read-while-write, proprietary code readout protection
- 320 Kbytes of SRAM including 64 Kbytes with hardware parity check
- External memory interface for static memories supporting SRAM, PSRAM, NOR, NAND and FRAM memories
- 2 x Octo-SPI memory interfaces



General-purpose inputs/outputs

- Up to 136 fast I/Os, most 5 V-tolerant, up to 14 I/Os with independent supply down to 1.08 V

Performance benchmark

- 1.25 DMIPS/MHz (Dhrystone 2.1)
- 409.20 CoreMark[®] (3.41 CoreMark/MHz @120 MHz)

Energy benchmark

- 285 ULPMark[™]CP score

Up to 24 capacitive sensing channels

- Support touchkey, linear and rotary touch sensors

Clock management

- 4 to 48 MHz crystal oscillator
- 32 kHz crystal oscillator for RTC (LSE)
- Internal 16 MHz factory-trimmed RC ($\pm 1\%$)
- Internal low-power 32 kHz RC ($\pm 5\%$)
- Internal multispeed 100 kHz to 48 MHz oscillator, auto-trimmed by LSE (better than $\pm 0.25\%$ accuracy)
- Internal 48 MHz with clock recovery
- 3 PLLs for system clock, USB, audio, ADC

Interconnect matrix

- 1 AHB bus matrix

14-channel DMA controller**23 communication peripherals**

- USB OTG 2.0 full-speed, LPM and BCD
- 2x SAIs (serial audio interface)
- 4x I2C FM+(1 Mbit/s), SMBus/PMBus
- 6x USARTs (ISO 7816, LIN, IrDA, modem)
- 3x SPIs (5x SPIs with the dual Octo-SPI)
- CAN (2.0B Active) and 2x SDMMC
- 8- to 14-bit camera interface up to 32 MHz (black and white) or 10 MHz (color)
- 8-/16-bit parallel synchronous data input/output slave interface (PSSI)

11 analog peripherals (independent supply)

- 2x 12-bit ADC 5 Msps, up to 16-bit with hardware oversampling, 200 μ A/Msps
- 2x 12-bit DAC, low-power sample and hold
- 2x operational amplifiers with built-in PGA
- 2x ultra-low-power comparators
- 2x digital filters for sigma delta modulator
- 1x temperature sensor

Advanced graphics features

- Chrom-ART Accelerator (DMA2D) for enhanced graphic content creation

- LCD-TFT controller

16x timers and watchdog

- 2 x 16-bit advanced motor-control
- 2 x 32-bit general purpose timers
- 5 x 16-bit general purpose timers
- 2x 16-bit basic timers
- 2x low-power 16-bit timers (available in Stop mode)
- 2x watchdogs
- 1x SysTick timer
- RTC with hardware calendar, alarms and calibration

Cryptographic acceleration

- AES (128/256-bit key)
- Public key accelerator (PKA)

True random generator**CRC calculation unit****HASH (SHA-256) hardware accelerator****Debug mode**

- Serial wire debug (SWD)
- JTAG
- Embedded Trace Macrocell™ (ETM)

96-bit unique ID**Table 1. Device summary**

Reference	Part numbers
STM32L4Q5xx	STM32L4Q5AG, STM32L4Q5CG, STM32L4Q5QG, STM32L4Q5RG, STM32L4Q5VG, STM32L4Q5ZG

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1 Introduction

This document provides the ordering information and mechanical device characteristics of the STM32L4Q5xx microcontrollers.

This document should be read in conjunction with the STM32L4Q5xx reference manual (RM0432). The reference manual is available from the STMicroelectronics website www.st.com.

For information on the Arm^{®(a)} Cortex[®]-M4 core, refer to the Cortex[®]-M4 Technical Reference Manual, available from the www.arm.com website.



arm

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2 Description

STM32L4Q5xx designates a family of ultra-low-power microcontroller devices (part of STM32L4+ Series) based on the high-performance Arm® Cortex®-M4 32-bit RISC core. They operate at a frequency of up to 120 MHz.

The Cortex-M4 core features a single-precision floating-point unit (FPU), which supports all the Arm® single-precision data-processing instructions and all the data types. The Cortex-M4 core also implements a full set of DSP (digital signal processing) instructions and a memory protection unit (MPU) that enhances the application's security.

These devices embed high-speed memories (1 Mbyte of Flash memory and 320 Kbytes of SRAM), a flexible external memory controller (FSMC) for static memories (for devices with packages of 100 pins and more), two OCTOSPI Flash memory interfaces (for devices with packages of 100 pins or more) and an extensive range of enhanced I/Os and peripherals connected to two APB buses, two AHB buses and a 32-bit multi-AHB bus matrix.

These devices embed several protection mechanisms for embedded Flash memory and SRAM: readout protection, write protection, proprietary code readout protection and a firewall.

These devices offer two fast 12-bit ADCs (5 Msps), two comparators, two operational amplifiers, two DAC channels, an internal voltage reference buffer, a low-power RTC, two general-purpose 32-bit timers, two 16-bit PWM timers dedicated to motor control, seven general-purpose 16-bit timers, and two 16-bit low-power timers. The devices support two digital filters for external sigma delta modulators (DFSDMs). In addition, up to 24 capacitive sensing channels are available.

They also feature standard and advanced communication interfaces such as:

- Four I2Cs
- Three SPIs
- Three USARTs, two UARTs and one low-power UART
- Two SAIs
- Two SDMMCs
- One CAN
- One USB OTG full-speed
- Camera interface
- Synchronous parallel data interface (PSSI)

STM32L4Q5xx devices embed an AES and a PKA hardware accelerator.

The devices operate in the -40 to +85 °C (+105 °C junction) and -40 to +125 °C (+130 °C junction) temperature ranges from a 1.71 to 3.6 V power supply. A comprehensive set of power-saving modes allows the design of low-power applications.

Some independent power supplies are supported such as an analog independent supply input for ADC, DAC, OPAMPs and comparators, a 3.3 V dedicated supply input for USB and up to 14 I/Os, which can be supplied independently down to 1.08 V. A VBAT input allows backup of the RTC and the registers.

The STM32L4Q5xx family offers eight packages from 48-pin to 169-pin.

Table 2. STM32L4Q5xx features and peripheral counts

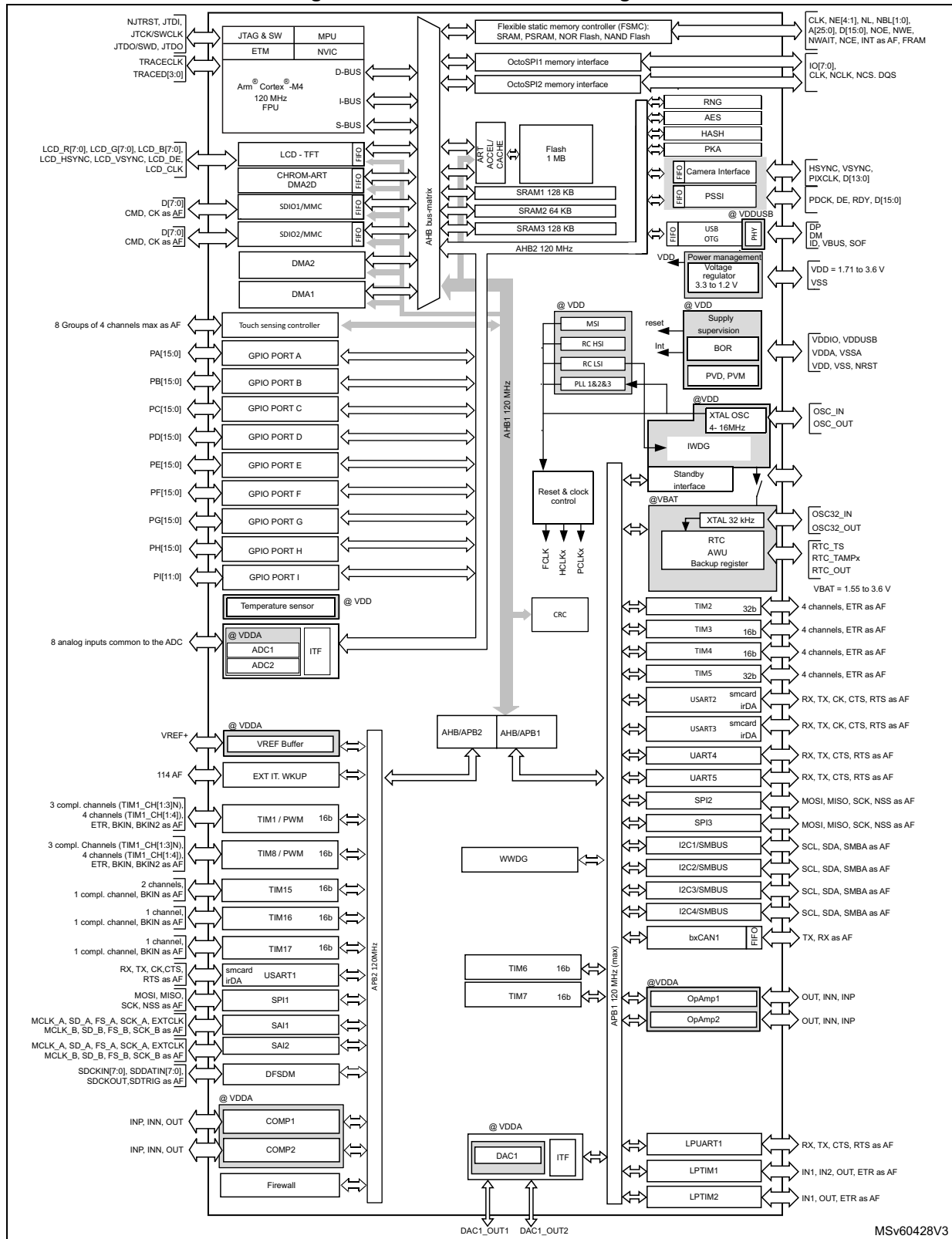
Peripheral		STM32L4 Q5CG (48 pins)	STM32L4 Q5RG (64 pins)	STM32L4 Q5VG (100 pins)	STM32L4 Q5QG (132 pins)	STM32L4 Q5ZG (144 pins)	STM32L4 Q5AG (169 pins)
Flash memory		1 Mbyte					
SRAM	System	320 (128 + 64 + 128) Kbytes					
	Backup	128 bytes					
External memory controller for static memories (FSMC)		No		Yes ⁽¹⁾		Yes	
OCTOSPI		2 ⁽²⁾			2		
Timers	Advanced control	2 (16-bit)					
	General purpose	5 (16-bit) 2 (32-bit)					
	Basic	2 (16-bit)					
	Low-power	2 (16-bit)					
	SysTick timer	1					
	Watchdog timers (independent, window)	2					
Comm. interfaces	SPI	3					
	I ² C	4					
	USART/UART UART LPUART	3 2 ⁽³⁾ 1					
	SAI	2					
	CAN	1					
	USB OTG FS	Yes					
	SDMMC	2 ⁽⁴⁾					
Digital filters for sigma-delta modulators		Yes (2 filters)					
Number of channels		4					
Real time clock (RTC)		Yes					
Tamper pins		3					
Camera interface		Yes ⁽⁵⁾					
PSSI		Yes ⁽⁶⁾					
Chrom-ART Accelerator		Yes					
LTDC		Yes ⁽⁷⁾		Yes			
True random number generator		Yes					

Table 2. STM32L4Q5xx features and peripheral counts (continued)

Peripheral	STM32L4 Q5CG (48 pins)	STM32L4 Q5RG (64 pins)	STM32L4 Q5VG (100 pins)	STM32L4 Q5QG (132 pins)	STM32L4 Q5ZG (144 pins)	STM32L4 Q5AG (169 pins)
AES	Yes					
HASH (SHA-256)	Yes					
Public key accelerator (PKA)	Yes					
GPIOs	38	52	83 ⁽⁸⁾	110	115	136
Wakeup pins	3	4	5	5	5	5
Nb of I/Os down to 1.08 V	0	0	0	14	14	14
Capacitive sensing Number of channels	9	21		24		
12-bit ADCs Number of channels	2					
	10	16	16	16	16	16
12-bit DAC channels	2					
Internal voltage reference buffer	Yes					
Analog comparator	2					
Operational amplifiers	2					
Max. CPU frequency	120 MHz					
Operating voltage (V _{DD})	1.71 to 3.6 V					
Operating temperature	Ambient operating temperature: -40 to 85 °C / -40 to 125 °C					
Packages	UFQFPN48/ LQFP48	LQFP64	LQFP100/ WLCSP100	UFBGA132	LQFP144	UFBGA169

- For the LQFP100 package, only FMC bank1 and NAND bank are available. Bank1 can only support a multiplexed NOR/PSRAM memory using the NE1 chip select.
- When multiplexed mode is enabled two OCTOSPIS are available, otherwise only one OCTOSPI is available, for further details on multiplexed mode refer to the RM0432 reference manual.
- For 48-pin and 64-pin packages only 1x UART is available.
- For 48 pins packages only one SDMMC2 with SDIO 4-bits is available. SDMMC1 and SDMMC2 are available for packages starting from 64 pins.
- DCMI is available on devices with packages starting from 64-pin.
- PSSI is available on devices with packages starting from 64-pin.
- For 48-pin and 64-pin packages, only RGB222 parallel output is supported.
- 81 GPIOs available for WLCSP100 package.

Figure 1. STM32L4Q5xx block diagram



1. AF: alternate function on I/O pins.

3 Functional overview

3.1 Arm[®] Cortex[®]-M4 core with FPU

The Arm[®] Cortex[®]-M4 with FPU processor is the latest generation of Arm[®] processors for embedded systems. It was developed to provide a low-cost platform that meets the needs of the MCU implementation, with a reduced pin count and with low-power consumption, while delivering outstanding computational performance and an advanced response to interrupts.

The Arm[®] Cortex[®]-M4 with FPU 32-bit RISC processor features an exceptional code-efficiency, delivering the expected high-performance from an Arm[®] core in a memory size usually associated with 8-bit and 16-bit devices.

The processor supports a set of DSP instructions which allows an efficient signal processing and a complex algorithm execution. Its single precision FPU speeds up the software development by using metalanguage development tools to avoid saturation.

With its embedded Arm[®] core, the family is compatible with all Arm[®] tools and software.

Figure 1 shows the general block diagram of the family devices.

3.2 Adaptive real-time memory accelerator (ART Accelerator)

The ART Accelerator is a memory accelerator that is optimized for the STM32 industry-standard Arm[®]Cortex[®]-M4 processors. It balances the inherent performance advantage of the Arm[®] Cortex[®]-M4 over Flash memory technologies, which normally requires the processor to wait for the Flash memory at higher frequencies.

To release the processor near 150 DMIPS performance at 120 MHz, the accelerator implements an instruction prefetch queue and a branch cache, which increases the program's execution speed from the Flash memory. Based on CoreMark benchmark, the performance achieved thanks to the ART accelerator is equivalent to 0 wait state program execution from the Flash memory at a CPU frequency of up to 120 MHz.

3.3 Memory protection unit (MPU)

The memory protection unit (MPU) is used to manage the CPU accesses to the memory and to prevent one task to accidentally corrupt the memory or the resources used by any other active task. This memory area is organized into up to eight protected areas, which can be divided in up into eight subareas each. The protection area sizes range between 32 bytes and the whole 4 gigabytes of addressable memory.

The MPU is especially helpful for applications where some critical or certified code has to be protected against the misbehavior of other tasks. It is usually managed by an RTOS (real-time operating system). If a program accesses a memory location that is prohibited by the MPU, the RTOS can detect it and take action. In an RTOS environment, the kernel can dynamically update the MPU area setting based on the process to be executed.

The MPU is optional and can be bypassed for applications that do not need it.

3.4 Memories

3.4.1 Embedded Flash memory

The devices feature 1 Mbyte of embedded Flash memory which is available for storing programs and data.

The Flash interface features:

- Single or dual bank operating modes
- Read-while-write (RWW) in dual bank mode

This feature allows to perform a read operation from one bank while an erase or program operation is performed to the other bank. The dual-bank boot is also supported. Each bank contains 128 pages of 4 Kbytes. In Single-bank mode the main Flash memory contains 128 pages of 8 Kbytes.

Flexible protections can be configured thanks to the option bytes:

- Readout protection (RDP) to protect the whole memory. Three levels of protection are available:
 - Level 0: no readout protection
 - Level 1: memory readout protection; the Flash memory cannot be read from or written to if either the debug features are connected or the boot in RAM or bootloader are selected
 - Level 2: chip readout protection; the debug features (Cortex-M4 JTAG and serial wire), the boot in RAM and the bootloader selection are disabled (JTAG fuse). This selection is irreversible.

Table 3. Access status versus readout protection level and execution modes

Area	Protection level	User execution			Debug, boot from RAM or boot from system memory (loader)		
		Read	Write	Erase	Read	Write	Erase
Main memory	1	Yes	Yes	Yes	No	No	No
	2	Yes	Yes	Yes	N/A	N/A	N/A
System memory	1	Yes	No	No	Yes	No	No
	2	Yes	No	No	N/A	N/A	N/A
Option bytes	1	Yes	Yes	Yes	Yes	Yes	Yes
	2	Yes	No	No	N/A	N/A	N/A
Backup registers	1	Yes	Yes	N/A ⁽¹⁾	No	No	N/A ⁽¹⁾
	2	Yes	Yes	N/A	N/A	N/A	N/A
SRAM2	1	Yes	Yes	Yes ⁽¹⁾	No	No	No ⁽¹⁾
	2	Yes	Yes	Yes	N/A	N/A	N/A
PKA RAM	1	Yes	Yes	Yes ⁽¹⁾	No	No	No ⁽¹⁾
	2	Yes	Yes	Yes	N/A	N/A	N/A

1. Erased when RDP change from Level 1 to Level 0.

- Write protection (WRP): the protected area is protected against erasing and programming:
 - In single bank mode, four areas can be selected with 8-Kbyte granularity.
 - In dual bank mode, two areas per bank can be selected with 4-Kbyte granularity.
- Proprietary code readout protection (PCROP): a part of the Flash memory can be protected against read and write from third parties. The protected area is execute-only and it can only be reached by the STM32 CPU as an instruction code, while all other accesses (DMA, debug and CPU data read, write and erase) are strictly prohibited:
 - In single bank mode, two areas can be selected with 128-bit granularity.
 - In dual bank mode, one area per bank can be selected with 64-bit granularity.

An additional option bit (PCROP_RDP) allows to select if the PCROP area is erased or not when the RDP protection is changed from Level 1 to Level 0.

The whole non-volatile memory embeds the error correction code (ECC) feature supporting:

- Single error detection and correction
- Double error detection
- The address of the ECC fail can be read in the ECC register.

3.4.2 Embedded SRAM

The devices feature 320 Kbytes of embedded SRAM. This SRAM is split into three blocks:

- 128 Kbytes mapped at address 0x2000 0000 (SRAM1).
- 64 Kbytes located at address 0x1000 0000 with hardware parity check (SRAM2). This memory is also mapped at address 0x2002 0000 offering a contiguous address space with the SRAM1.

This block is accessed through the ICode/DCode buses for maximum performance. Either 64 Kbytes or 4 Kbytes of SRAM2 can be retained in Standby mode. The SRAM2 can be write-protected with 1 Kbyte granularity.

- 128 Kbytes mapped at address 0x2003 0000 (SRAM3).

The memory can be accessed in read/write at CPU clock speed with 0 wait states.

3.5 Boot modes

At startup, a BOOT0 pin and an nBOOT1 option bit are used to select one of three boot options:

- Boot from user Flash
- Boot from system memory
- Boot from embedded SRAM

The BOOT0 value may come from the PH3-BOOT0 pin or from an option bit depending on the value of a user option bit to free the GPIO pad if needed.

A Flash empty-check mechanism is implemented to force the boot from system Flash if the first Flash memory location is not programmed and if the boot selection is configured to boot from main Flash.

The boot loader is located in the system memory. It is used to reprogram the Flash memory by using USART, I2C, SPI, CAN or USB OTG FS in device mode through the DFU (device firmware upgrade).

3.6 Firewall

These devices embed a firewall which protects code sensitive and secure data from any access performed by a code executed outside of the protected areas.

Each illegal access generates a reset which kills immediately the detected intrusion.

The main features of the firewall are the following:

- Three segments can be protected and defined thanks to the firewall registers:
 - Code segment (located in Flash or SRAM1 if defined as executable protected area)
 - Non-volatile data segment (located in Flash)
 - Volatile data segment (located in SRAM1)
- The start address and the length of each segment are configurable:
 - Code segment: up to 1024 Kbytes with granularity of 256 bytes
 - Non-volatile data segment: up to 1024 Kbytes with granularity of 256 bytes
 - Volatile data segment: up to 128 Kbytes of SRAM1 with a granularity of 64 bytes
- Specific mechanism implemented to open the firewall to get access to the protected areas (call gate entry sequence)
- Volatile data segment can be shared or not with the non-protected code
- Volatile data segment can be executed or not depending on the firewall configuration

The Flash readout protection must be set to level 2 in order to reach the expected level of protection.

3.7 Power supply management

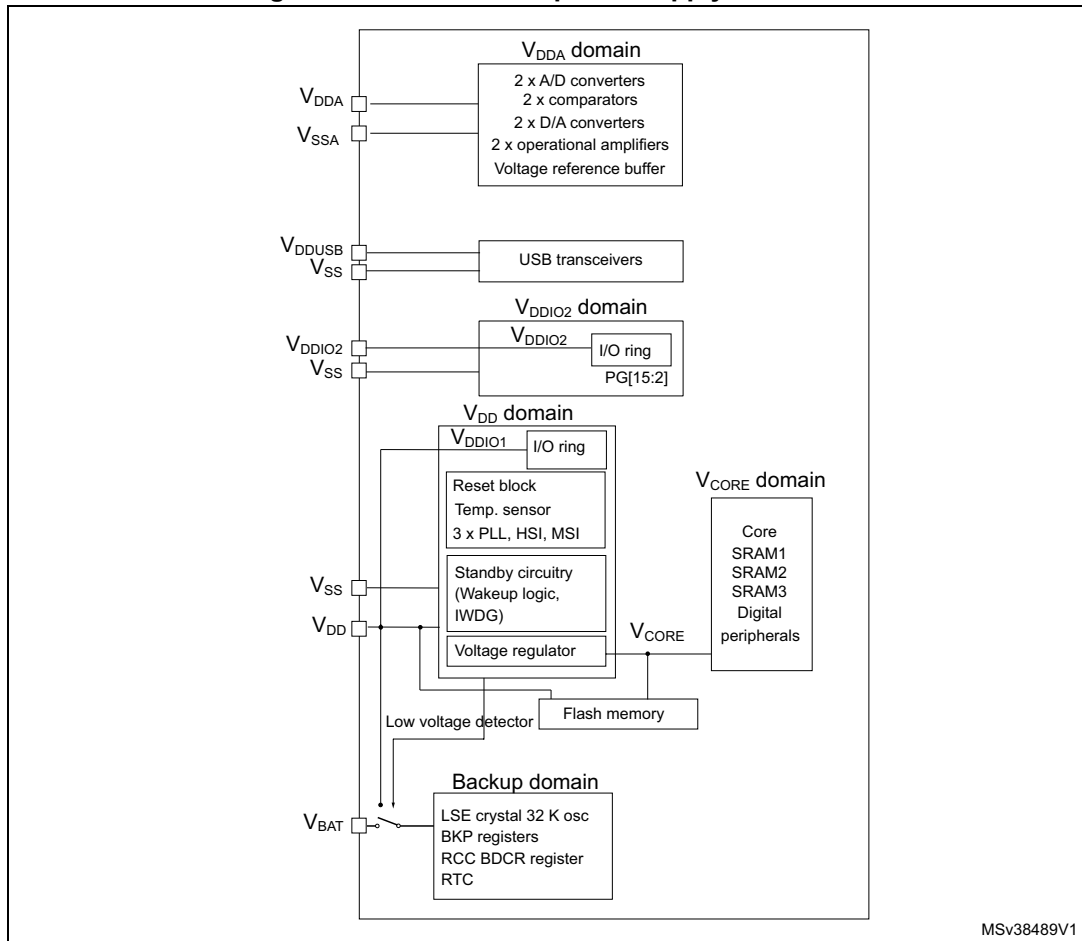
3.7.1 Power supply schemes

The STM32L4Q5xx devices require a 1.71 V to 3.6 V V_{DD} operating voltage supply. Several independent supplies can be provided for specific peripherals:

- $V_{DD} = 1.71 \text{ V to } 3.6 \text{ V}$
 V_{DD} is the external power supply for the I/Os, the internal regulator and the system analog such as reset, power management and internal clocks. It is provided externally through the VDD pins.
- $V_{DDA} = 1.62 \text{ V (ADCs/COMP)} / 1.8 \text{ V (DACs/OPAMP)} \text{ to } 2.4 \text{ V (VREFBUF)} \text{ to } 3.6 \text{ V}$
 V_{DDA} is the external analog power supply for A/D converters, D/A converters, voltage reference buffer, operational amplifiers and comparators. The V_{DDA} voltage level is independent from the V_{DD} voltage and should preferably be connected to V_{DD} when these peripherals are not used.
- $V_{DDUSB} = 3.0 \text{ V to } 3.6 \text{ V}$
 V_{DDUSB} is the external independent power supply for USB transceivers. The V_{DDUSB} voltage level is independent from the VDD voltage and should preferably be connected to VDD when the USB is not used.
- $V_{DDIO2} = 1.08 \text{ V to } 3.6 \text{ V}$
- V_{DDIO2} is the external power supply for 14 I/Os (port G[15:2]). The V_{DDIO2} voltage level is independent from the VDD voltage and should preferably be connected to VDD when PG[15:2] are not used.
- $V_{BAT} = 1.55 \text{ V to } 3.6 \text{ V}$
 V_{BAT} is the power supply for RTC, external clock 32 kHz oscillator and backup registers (through power switch) when V_{DD} is not present.
- V_{REF-}, V_{REF+}
 V_{REF+} is the input reference voltage for ADCs and DACs. It is also the output of the internal voltage reference buffer when enabled.
 When $V_{DDA} < 2 \text{ V}$ V_{REF+} must be equal to V_{DDA} .
 When $V_{DDA} \geq 2 \text{ V}$ V_{REF+} must be between 2 V and V_{DDA} .
 V_{REF+} can be grounded when ADC and DAC are not active.
 The internal voltage reference buffer supports two output voltages, which are configured with VRS bit in the VREFBUF_CSR register:
 - V_{REF+} around 2.048 V. This requires V_{DDA} equal to or higher than 2.4 V.
 - V_{REF+} around 2.5 V. This requires V_{DDA} equal to or higher than 2.8 V. V_{REF-} and V_{REF+} pins are not available on all packages. When not available, they are bonded to VSSA and VDDA, respectively.
 When the V_{REF+} is double-bonded with VDDA in a package, the internal voltage reference buffer is not available and must be kept disable (refer to datasheet for packages pinout description).
 V_{REF-} must always be equal to V_{SSA} .

An embedded linear voltage-regulator is used to supply the internal digital power V_{CORE} . V_{CORE} is the power supply for digital peripherals, SRAM1, SRAM2 and SRAM3. The Flash is supplied by V_{CORE} and V_{DD} .

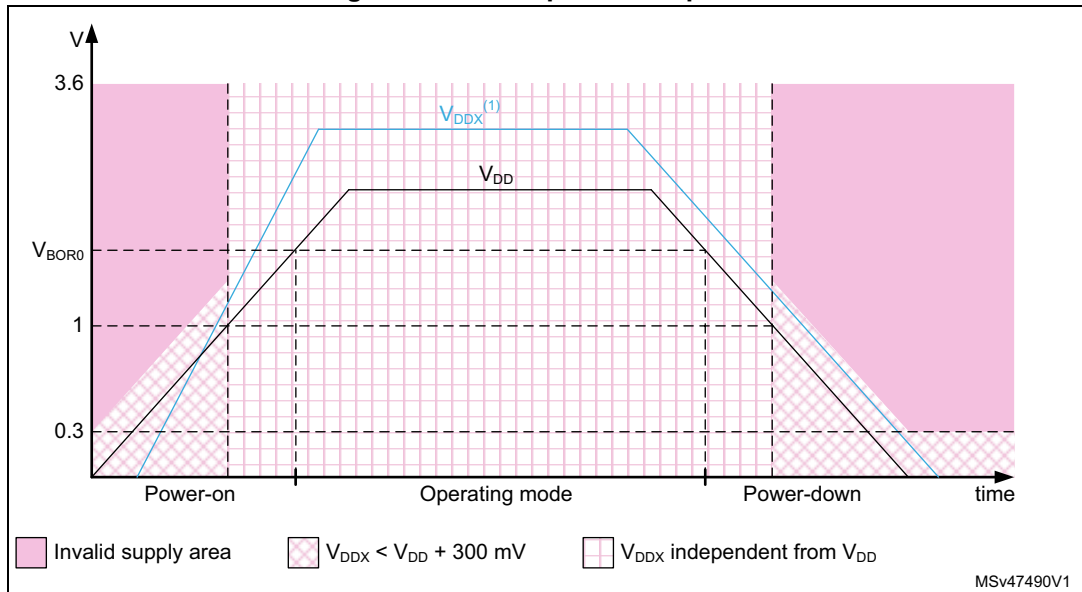
Figure 2. STM32L4Q5xx power supply overview



During power-up and power-down phases, the following power sequence requirements must be respected:

- When V_{DD} is below 1 V, other power supplies (V_{DDA}, V_{DDIO2} and V_{DDUSB}) must remain below V_{DD} +300 mV.
- When V_{DD} is above 1 V, all power supplies are independent.
- During the power-down phase, V_{DD} can temporarily become lower than other supplies only if the energy provided to the MCU remains below 1 mJ; this allows external decoupling capacitors to be discharged with different time constants during the power-down transient phase.

Figure 3. Power-up/down sequence



1. V_{DDX} refers to any power supply among V_{DDA}, V_{DDIO2} and V_{DDUSB}.

3.7.2 Power supply supervisor

The STM32L4Q5xx devices have an integrated ultra-low-power Brownout reset (BOR) active in all modes (except for Shutdown mode). The BOR ensures proper operation of the devices after power-on and during power-down. The devices remain in reset mode when the monitored supply voltage V_{DD} is below a specified threshold, without the need for an external reset circuit.

The lowest BOR level is 1.71 V at power on, and other higher thresholds can be selected through option bytes. The devices feature an embedded programmable voltage detector (PVD) that monitors the V_{DD} power supply and compares it to the VPVD threshold.

An interrupt can be generated when V_{DD} drops below the VPVD threshold and/or when V_{DD} is higher than the VPVD threshold. The interrupt service routine can then generate a warning message and/or put the MCU into a safe state. The PVD is enabled by software.

In addition, the devices embed a peripheral voltage monitor which compares the independent supply voltages V_{DDA}, V_{DDUSB}, V_{DDIO2} with a fixed threshold in order to ensure that the peripheral is in its functional supply range.

3.7.3 Voltage regulator

Two embedded linear voltage regulators supply most of the digital circuitries: the main regulator (MR) and the low-power regulator (LPR).

- The MR is used in the Run and Sleep modes and in the Stop 0 mode.
- The LPR is used in Low-power run, Low-power sleep, Stop 1 and Stop 2 modes. It is also used to supply the 64 Kbytes SRAM2 in standby with RAM2 retention.
- Both regulators are in power-down while they are in standby and Shutdown modes: the regulator output is in high impedance, and the kernel circuitry is powered down thus inducing zero consumption.

The ultra-low-power devices support dynamic voltage scaling to optimize its power consumption in Run mode. The voltage from the main regulator that supplies the logic (VCORE) can be adjusted according to the system's maximum operating frequency.

The main regulator operates in the following ranges:

- Range 1 boost mode with the CPU running at up to 120 MHz.
- Range 1 normal mode with the CPU running at up to 80 MHz.
- Range 2 with a maximum CPU frequency of 26 MHz. All peripheral clocks are also limited to 26 MHz.

The VCORE can be supplied by the low-power regulator, the main regulator being switched off. The system is then in Low-power run mode.

- Low-power run mode with the CPU running at up to 2 MHz. Peripherals with independent clock can be clocked by the HSI16.

Note: The USB can only be used when the main regulator is in Range 1 boost mode.

3.7.4 Low-power modes

The ultra-low-power devices support seven low-power modes to achieve the best compromise between low-power consumption, short startup time, available peripherals and available wake-up sources. The table below shows the related STM32L4Q5xx modes overview.

Table 4. STM32L4Q5xx modes overview

Mode	Regulator ⁽¹⁾	CPU	Flash	SRAM	Clocks	DMA and Peripherals ⁽²⁾	Wakeup source
Run	Range 1	Yes	ON ⁽³⁾	ON	Any	All	N/A
	Range 2					All except OTG_FS, RNG, LCD-TFT	
LPRun	LPR	Yes	ON ⁽³⁾	ON	Any except PLL	All except OTG_FS, RNG, LCD-TFT	N/A
Sleep	Range 1	No	ON ⁽³⁾	ON ⁽⁴⁾	Any	All	Any interrupt or event
	Range 2					All except OTG_FS, RNG, LCD-TFT	
LPSleep	LPR	No	ON ⁽³⁾	ON ⁽⁴⁾	Any except PLL	All except OTG_FS, RNG, LCD-TFT	Any interrupt or event
Stop 0	Range 1	No	Off	ON	LSE LSI	BOR, PVD, PVM RTC, IWDG COMPx (x=1,2) DAC1 OPAMPx (x=1,2) USARTx (x=1...5) ⁽⁵⁾ LPUART1 ⁽⁵⁾ I2Cx (x=1...4) ⁽⁶⁾ LPTIMx (x=1,2) *** All other peripherals are frozen	Reset pin, all I/Os BOR, PVD, PVM RTC, IWDG COMPx (x=1..2) USARTx (x=1...5) ⁽⁵⁾ LPUART1 ⁽⁵⁾ I2Cx (x=1...4) ⁽⁶⁾ LPTIMx (x=1,2) OTG_FS ⁽⁷⁾
	Range 2						
Stop 1	LPR	No	Off	ON	LSE LSI	BOR, PVD, PVM RTC, IWDG COMPx (x=1,2) DAC1 OPAMPx (x=1,2) USARTx (x=1...5) ⁽⁵⁾ LPUART1 ⁽⁵⁾ I2Cx (x=1...4) ⁽⁶⁾ LPTIMx (x=1,2) *** All other peripherals are frozen	Reset pin, all I/Os BOR, PVD, PVM RTC, IWDG COMPx (x=1..2) USARTx (x=1...5) ⁽⁵⁾ LPUART1 ⁽⁵⁾ I2Cx (x=1...4) ⁽⁶⁾ LPTIMx (x=1,2) OTG_FS ⁽⁷⁾

Table 4. STM32L4Q5xx modes overview (continued)

Mode	Regulator ⁽¹⁾	CPU	Flash	SRAM	Clocks	DMA and Peripherals ⁽²⁾	Wakeup source
Stop 2	LPR	No	Off	ON	LSE LSI	BOR, PVD, PVM RTC, IWDG COMPx (x=1..2) I2C3 ⁽⁶⁾ LPUART1 ⁽⁵⁾ LPTIM1x (x=1,2) *** All other peripherals are frozen	Reset pin, all I/Os BOR, PVD, PVM RTC, IWDG COMPx (x=1..2) I2C3 ⁽⁶⁾ LPUART1 ⁽⁵⁾ LPTIM1x (x=1,2)
Standby	LPR	Powered Off	Off	SRAM2 ON	LSE LSI	BOR, RTC, IWDG *** All other peripherals are powered off *** I/O configuration can be floating, pull-up or pull-down	Reset pin 5 I/Os (WKUPx) ⁽⁸⁾ BOR, RTC, IWDG
	OFF			Powered Off		I/O configuration can be floating, pull-up or pull-down ⁽⁹⁾	
Shutdown	OFF	Powered Off	Off	Powered Off	LSE	RTC *** All other peripherals are powered off *** I/O configuration can be floating, pull-up or pull-down ⁽⁹⁾	Reset pin 5 I/Os (WKUPx) ⁽⁸⁾ RTC

1. LPR means Main regulator is OFF and Low-power regulator is ON.
2. All peripherals can be active or clock gated to save power consumption.
3. The Flash memory can be put in power-down and its clock can be gated off when executing from SRAM.
4. The SRAM1, SRAM2 and SRAM3 clocks can be gated on or off independently.
5. U(S)ART and LPUART reception is functional in Stop mode, and generates a wakeup interrupt on Start, address match or received frame event.
6. I2C address detection is functional in Stop mode, and generates a wakeup interrupt in case of address match.
7. OTG_FS wakeup by resume from suspend and attach detection protocol event.
8. The I/Os with wakeup from Standby/Shutdown capability are: PA0, PC13, PE6, PA2, PC5.
9. I/Os can be configured with internal pull-up, pull-down or floating in Shutdown mode but the configuration is lost when exiting the Shutdown mode.

By default, the microcontroller is in Run mode after a system or a power reset. It is up to the user to select one of the low-power modes described below:

- **Sleep mode**
In Sleep mode, only the CPU is stopped. All peripherals continue to operate and can wake up the CPU when an interrupt/event occurs.
- **Low-power run mode**
This mode is achieved with VCORE supplied by the low-power regulator to minimize the regulator's operating current. The code can be executed from SRAM or from Flash,

and the CPU frequency is limited to 2 MHz. The peripherals with independent clock can be clocked by HSI16.

- **Low-power sleep mode**

This mode is entered from the Low-power run mode. Only the CPU clock is stopped. When wakeup is triggered by an event or an interrupt, the system reverts to the Low-power run mode.

- **Stop 0, Stop 1 and Stop 2 modes**

Stop mode achieves the lowest power consumption while retaining the content of SRAM and registers. All clocks in the VCORE domain are stopped, the PLL, the MSI RC, the HSI16 RC and the HSE crystal oscillators are disabled. The LSE or LSI is still running.

The RTC can remain active (Stop mode with RTC, Stop mode without RTC).

Some peripherals with wake-up capability can enable the HSI16 RC during Stop mode to detect their wake-up condition.

Three Stop modes are available: Stop 0, Stop 1 and Stop 2 modes. In Stop 2 mode, most of the VCORE domain is put in a lower leakage mode.

Stop 1 offers the largest number of active peripherals and wakeup sources, a smaller wakeup time but a higher consumption than Stop 2. In Stop 0 mode, the main regulator remains ON, allowing a very fast wakeup time but with much higher consumption.

The system clock when exiting from Stop 0, Stop 1 or Stop 2 modes can be either MSI up to 48 MHz or HSI16, depending on software configuration.

- **Standby mode**

The Standby mode is used to achieve the lowest power consumption with BOR. The internal regulator is switched off so that the VCORE domain is powered off. The PLL, the MSI RC, the HSI16 RC and the HSE crystal oscillators are also switched off.

The RTC can remain active (Standby mode with RTC, Standby mode without RTC).

The Brownout reset (BOR) always remains active in Standby mode.

The state of each I/O during Standby mode can be selected by software: I/O with internal pull-up, internal pull-down or floating.

After entering Standby mode, SRAM1, SRAM3 and register contents are lost except for registers in the Backup domain and Standby circuitry. Optionally, the full SRAM2 (or

only the upper 4 Kbytes) can be retained in Standby mode, supplied by the low-power regulator (standby with RAM2 retention mode).

The device exits Standby mode when an external reset (NRST pin), an IWDG reset, WKUP pin event (configurable rising or falling edge), or an RTC event occurs (alarm, periodic wakeup, timestamp, tamper) or a failure is detected on LSE (CSS on LSE).

The system clock after wakeup is MSI up to 8 MHz.

- **Shutdown mode**

The Shutdown mode allows to achieve the lowest power consumption. The internal regulator is switched off so that the VCORE domain is powered off. The PLL, the HSI16, the MSI, the LSI and the HSE oscillators are also switched off.

The RTC can remain active (Shutdown mode with RTC, Shutdown mode without RTC).

The BOR is not available in Shutdown mode. No power voltage monitoring is possible in this mode, therefore the switch to Backup domain is not supported.

SRAM1, SRAM2, SRAM3 and register contents are lost except for registers in the Backup domain.

The device exits Shutdown mode when an external reset (NRST pin), a WKUP pin event (configurable rising or falling edge), or an RTC event occurs (alarm, periodic wakeup, timestamp, tamper).

The system clock after wakeup is MSI at 4 MHz.

Table 5. Functionalities depending on the working mode⁽¹⁾

Peripheral	Run	Sleep	Low-power run	Low-power sleep	Stop 0/1		Stop 2		Standby		Shutdown		VBAT
					-	Wakeup capability	-	Wakeup capability	-	Wakeup capability	-	Wakeup capability	
CPU	Y	-	Y	-	-	-	-	-	-	-	-	-	-
Flash memory (2 Mbytes)	O ⁽²⁾	O ⁽²⁾	O ⁽²⁾	O ⁽²⁾	-	-	-	-	-	-	-	-	-
SRAM1 (128 Kbytes)	Y	Y ⁽³⁾	Y	Y ⁽³⁾	Y	-	Y	-	-	-	-	-	-
SRAM2 (64 Kbytes)	Y	Y ⁽³⁾	Y	Y ⁽³⁾	Y	-	Y	-	O ⁽⁴⁾	-	-	-	-
SRAM3 (128 Kbytes)	Y	Y ⁽³⁾	Y	Y ⁽³⁾	Y	-	Y ⁽³⁾	-	-	-	-	-	-
FSMC	O	O	O	O	-	-	-	-	-	-	-	-	-
OCTOSPIx (x=1,2)	O	O	O	O	-	-	-	-	-	-	-	-	-
Backup registers	Y	Y	Y	Y	Y	-	Y	-	Y	-	Y	-	Y
Brownout reset (BOR)	Y	Y	Y	Y	Y	Y	Y	Y	Y	Y	-	-	-
Programmable voltage detector (PVD)	O	O	O	O	O	O	O	O	-	-	-	-	-
Peripheral voltage monitor (PVMx; x=1,2,3,4)	O	O	O	O	O	O	O	O	-	-	-	-	-
DMA	O	O	O	O	-	-	-	-	-	-	-	-	-
DMA2D	O	O	O	O	-	-	-	-	-	-	-	-	-
High speed internal (HSI16)	O	O	O	O	(5)	-	(5)	-	-	-	-	-	-
Oscillator HSI48	O	O	-	-	-	-	-	-	-	-	-	-	-
High speed external (HSE)	O	O	O	O	-	-	-	-	-	-	-	-	-
Low speed internal (LSI)	O	O	O	O	O	-	O	-	O	-	-	-	-
Low speed external (LSE)	O	O	O	O	O	-	O	-	O	-	O	-	O
Multi speed internal (MSI)	O	O	O	O	-	-	-	-	-	-	-	-	-
Clock security system (CSS)	O	O	O	O	-	-	-	-	-	-	-	-	-

Table 5. Functionalities depending on the working mode⁽¹⁾ (continued)

Peripheral	Run	Sleep	Low-power run	Low-power sleep	Stop 0/1		Stop 2		Standby		Shutdown		VBAT
					-	Wakeup capability	-	Wakeup capability	-	Wakeup capability	-	Wakeup capability	
Clock security system on LSE	0	0	0	0	0	0	0	0	0	0	-	-	-
RTC / Auto wakeup	0	0	0	0	0	0	0	0	0	0	0	0	0
Number of RTC Tamper pins	3	3	3	3	3	0	3	0	3	0	3	0	3
Camera interface (DCMI)	0	0	0	0	-	-	-	-	-	-	-	-	-
PSSI	0	0	0	0	-	-	-	-	-	-	-	-	-
LCD-TFT	0	0	-	-	-	-	-	-	-	-	-	-	-
USB OTG FS	0 ⁽⁸⁾	0 ⁽⁸⁾	-	-	-	0	-	-	-	-	-	-	-
USARTx (x=1,2,3,4,5)	0	0	0	0	0 ⁽⁶⁾	0 ⁽⁶⁾	-	-	-	-	-	-	-
Low-power UART (LPUART)	0	0	0	0	0 ⁽⁶⁾	0 ⁽⁶⁾	0 ⁽⁶⁾	0 ⁽⁶⁾	-	-	-	-	-
I2Cx (x=1,2,4)	0	0	0	0	0 ⁽⁷⁾	0 ⁽⁷⁾	-	-	-	-	-	-	-
I2C3	0	0	0	0	0 ⁽⁷⁾	0 ⁽⁷⁾	0 ⁽⁷⁾	0 ⁽⁷⁾	-	-	-	-	-
SPIx (x=1,2,3)	0	0	0	0	-	-	-	-	-	-	-	-	-
CAN(x=1,2)	0	0	0	0	-	-	-	-	-	-	-	-	-
SDMMC1	0	0	0	0	-	-	-	-	-	-	-	-	-
SDMMC2	0	0	0	0	-	-	-	-	-	-	-	-	-
SAIx (x=1,2)	0	0	0	0	-	-	-	-	-	-	-	-	-
DFSDM1	0	0	0	0	-	-	-	-	-	-	-	-	-
ADCx (x=1,2)	0	0	0	0	-	-	-	-	-	-	-	-	-
DAC1	0	0	0	0	0	-	-	-	-	-	-	-	-
VREFBUF	0	0	0	0	0	-	-	-	-	-	-	-	-
OPAMPx (x=1,2)	0	0	0	0	0	-	-	-	-	-	-	-	-
COMPx (x=1,2)	0	0	0	0	0	0	0	0	-	-	-	-	-
Temperature sensor	0	0	0	0	-	-	-	-	-	-	-	-	-
Timers (TIMx)	0	0	0	0	-	-	-	-	-	-	-	-	-
Low-power timer 1 (LPTIM1)	0	0	0	0	0	0	0	0	-	-	-	-	-

Table 5. Functionalities depending on the working mode⁽¹⁾ (continued)

Peripheral	Run	Sleep	Low-power run	Low-power sleep	Stop 0/1		Stop 2		Standby		Shutdown		VBAT
					-	Wakeup capability	-	Wakeup capability	-	Wakeup capability	-	Wakeup capability	
Low-power timer 2 (LPTIM2)	O	O	O	O	O	O	O	O	-	-	-	-	-
Independent watchdog (IWDG)	O	O	O	O	O	O	O	O	O	O	-	-	-
Window watchdog (WWDG)	O	O	O	O	-	-	-	-	-	-	-	-	-
SysTick timer	O	O	O	O	-	-	-	-	-	-	-	-	-
Touch sensing controller (TSC)	O	O	O	O	-	-	-	-	-	-	-	-	-
Random number generator (RNG)	O ⁽⁸⁾	O ⁽⁸⁾	-	-	-	-	-	-	-	-	-	-	-
AES hardware accelerator	O	O	O	O	-	-	-	-	-	-	-	-	-
PKA	O	O	O	O	-	-	-	-	-	-	-	-	-
HASH hardware accelerator	O	O	O	O	-	-	-	-	-	-	-	-	-
CRC calculation unit	O	O	O	O	-	-	-	-	-	-	-	-	-
GPIOs	O	O	O	O	O	O	O	O	⁽⁹⁾ 5 pins ⁽¹⁰⁾	⁽¹¹⁾ 5 pins ⁽¹⁰⁾	-	-	-

1. Legend: Y = yes (enable). O = optional (disable by default, can be enabled by software). - = not available.

Gray cells highlight the wakeup capability in each mode.

- The Flash can be configured in power-down mode. By default, it is not in power-down mode.
- The SRAM clock can be gated on or off. In Stop 2 mode, the content of SRAM3 is preserved or not depending on the RRSTP bit in PWR_CR1 register.
- The upper 4 Kbytes or the full SRAM2 content is preserved depending on RRS[1:0] bits configuration in PWR_CR3 register.
- Some peripherals with wakeup from Stop capability can request HSI16 to be enabled. In this case, HSI16 is woken up by the peripheral, and only feeds the peripheral which requested it. HSI16 is automatically put off when the peripheral does not need it anymore.
- JART and LPUART reception is functional in Stop mode, and generates a wakeup interrupt on Start, address match or received frame event.
- I2C address detection is functional in Stop mode, and generates a wakeup interrupt in case of address match.
- Voltage scaling range 1 only.
- I/Os can be configured with internal pull-up, pull-down or floating in Standby mode.
- The I/Os with wakeup from standby/shutdown capability are: PA0, PC13, PE6, PA2, PC5.
- I/Os can be configured with internal pull-up, pull-down or floating in Shutdown mode but the configuration is lost when exiting the Shutdown mode.

3.7.5 Reset mode

In order to improve the consumption under reset, the I/Os state under and after reset is “analog state” (the I/O schmitt trigger is disable). In addition, the internal reset pull-up is deactivated when the reset source is internal.

3.7.6 VBAT operation

The VBAT pin allows the device VBAT domain to be powered from an external battery, an external supercapacitor, or from V_{DD} when there is no external battery and when an external supercapacitor is present. The VBAT pin supplies the RTC with LSE and the backup registers. Three anti-tamper detection pins are available in VBAT mode.

The VBAT operation is automatically activated when V_{DD} is not present. An internal VBAT battery charging circuit is embedded and can be activated when V_{DD} is present.

Note: When the microcontroller is supplied from VBAT, neither external interrupts nor RTC alarm/events exit the microcontroller from the VBAT operation.

3.8 Peripheral and interconnect matrix

Several peripherals have direct connections between them, which allow autonomous communication between them and support the saving of CPU resources (thus power supply consumption). In addition, these hardware connections allow fast and predictable latency.

Depending on the peripherals, these interconnections can operate in Run, Sleep, Low-power run, Low-power Sleep, Stop 0, Stop 1 and Stop 2 modes. See [Table 6](#) for more details.

Table 6. STM32L4Q5xx peripherals interconnect matrix

Interconnect source	Interconnect destination	Interconnect action	Run	Sleep	Low-power run	Low-power sleep	Stop 0 / Stop 1	Stop 2
TIMx	TIMx	Timers synchronization or chaining	Y	Y	Y	Y	-	-
	ADCx DAC1 DFSDM1	Conversion triggers	Y	Y	Y	Y	-	-
	DMA	Memory to memory transfer trigger	Y	Y	Y	Y	-	-
	COMPx	Comparator output blanking	Y	Y	Y	Y	-	-
COMPx	TIM1, 8 TIM2, 3	Timer input channel, trigger, break from analog signals comparison	Y	Y	Y	Y	-	-
	LPTIMERx	Low-power timer triggered by analog signals comparison	Y	Y	Y	Y	Y	Y
ADCx	TIM1, 8	Timer triggered by analog watchdog	Y	Y	Y	Y	-	-

Table 6. STM32L4Q5xx peripherals interconnect matrix (continued)

Interconnect source	Interconnect destination	Interconnect action	Run	Sleep	Low-power run	Low-power sleep	Stop 0 / Stop 1	Stop 2
RTC	TIM16	Timer input channel from RTC events	Y	Y	Y	Y	-	-
	LPTIMERx	Low-power timer triggered by RTC alarms or tampers	Y	Y	Y	Y	Y	Y
All clocks sources (internal and external)	TIM2 TIM15, 16, 17	Clock source used as input channel for RC measurement and trimming	Y	Y	Y	Y	-	-
USB	TIM2	Timer triggered by USB SOF	Y	Y	-	-	-	-
CSS CPU (hard fault) RAM (parity error) Flash memory (ECC error) COMPx PVD DFSDM1 (analog watchdog, short circuit detection)	TIM1,8 TIM15,16,17	Timer break	Y	Y	Y	Y	-	-
GPIO	TIMx	External trigger	Y	Y	Y	Y	-	-
	LPTIMERx	External trigger	Y	Y	Y	Y	Y	Y
	ADC DAC1 DFSDM1	Conversion external trigger	Y	Y	Y	Y	-	-

3.9 Reset and clock controller (RCC)

The clock controller (see [Figure 4](#)) distributes the clocks coming from the different oscillators to the core and to the peripherals. It also manages the clock gating for low-power modes and ensures the clock robustness. It features:

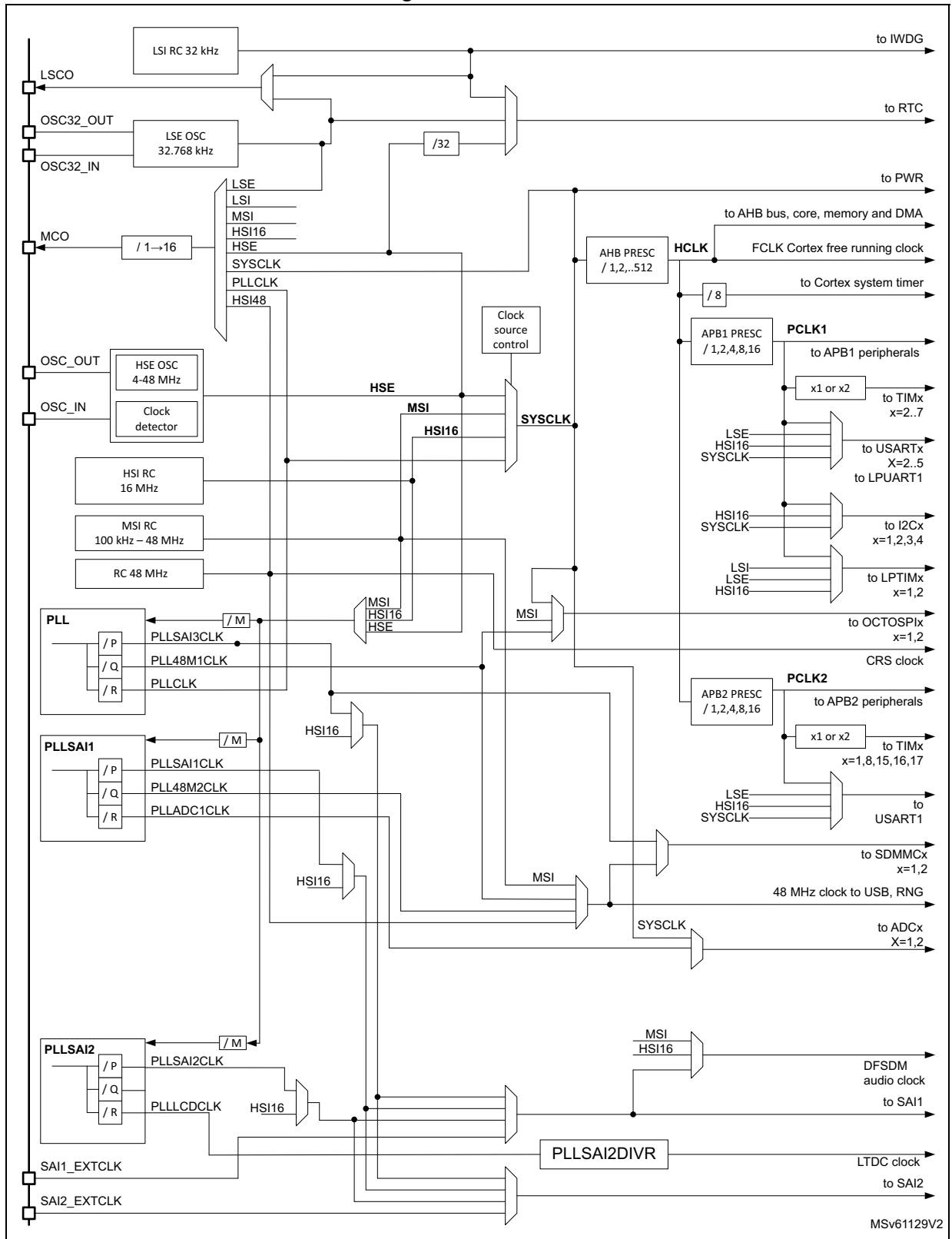
- **Clock prescaler:** to get the best trade-off between speed and current consumption, the clock frequency to the CPU and peripherals can be adjusted by a programmable prescaler.
- **Safe clock switching:** clock sources can be changed safely on the fly in Run mode through a configuration register.
- **Clock management:** to reduce the power consumption, the clock controller can stop the clock to the core, individual peripherals or memory.
- **System clock source:** four different clock sources can be used to drive the master clock SYSCLK:
 - 4 to 48 MHz high-speed external crystal or ceramic resonator (HSE), that can supply a PLL. The HSE can also be configured in bypass mode for an external clock.
 - 16 MHz high-speed internal RC oscillator (HSI16), trimmable by software, that can supply a PLL
 - Multispeed internal RC oscillator (MSI), trimmable by software, able to generate 12 frequencies from 100 kHz to 48 MHz. When a 32.768 kHz clock source is available in the system (LSE), the MSI frequency can be automatically trimmed by hardware to reach better than $\pm 0.25\%$ accuracy. In this mode the MSI can feed the USB device, saving the need of an external high-speed crystal (HSE). The MSI can supply a PLL.
 - System PLL which can be fed by HSE, HSI16 or MSI, with a maximum frequency at 120 MHz.
- **RC48 with clock recovery system (HSI48):** internal 48 MHz clock source (HSI48) can be used to drive the USB, the SDMMC or the RNG peripherals. This clock can be output on the MCO.
- **Auxiliary clock source:** two ultra-low-power clock sources that can be used to drive the real-time clock:
 - 32.768 kHz low-speed external crystal (LSE), supporting four drive capability modes. The LSE can also be configured in bypass mode for an external clock.
 - 32 kHz low-speed internal RC (LSI), also used to drive the independent watchdog. The LSI clock accuracy is $\pm 5\%$ accuracy.
- **Peripheral clock sources:** several peripherals (USB, SDMMC, RNG, SAI, USARTs, I2Cs, LPTimers, ADC) have their own independent clock whatever the system clock. Three PLLs, each having three independent outputs allowing the highest flexibility, can generate independent clocks for the ADC, the USB/SDMMC/RNG, the two SAIs, and LCD-TFT.
- **Startup clock:** after reset, the microcontroller restarts by default with an internal 4 MHz clock (MSI). The prescaler ratio and clock source can be changed by the application program as soon as the code execution starts.
- **Clock security system (CSS):** this feature can be enabled by software. If a HSE clock failure occurs, the master clock is automatically switched to HSI16 and a software

interrupt is generated if enabled. LSE failure can also be detected and generated an interrupt.

- Clock-out capability:
 - **MCO (microcontroller clock output)**: it outputs one of the internal clocks for external use by the application
 - **LSCO (low-speed clock output)**: it outputs LSI or LSE in all low-power modes (except VBAT).

Several prescalers allow the AHB frequency, the high speed APB (APB2) and the low-speed APB (APB1) domains to be configured. The maximum frequency of the AHB and the APB domains is 120 MHz.

Figure 4. Clock tree



3.10 Clock recovery system (CRS)

The devices embed a special block which allows automatic trimming of the internal 48 MHz oscillator to guarantee its optimal accuracy over the whole device operational range. This automatic trimming is based on the external synchronization signal, which can be either derived from USB SOF signalization, from LSE oscillator, from an external signal on CRS_SYNC pin or generated by user software. For faster lock-in during startup it is also possible to combine automatic trimming with manual trimming action.

3.11 General-purpose inputs/outputs (GPIOs)

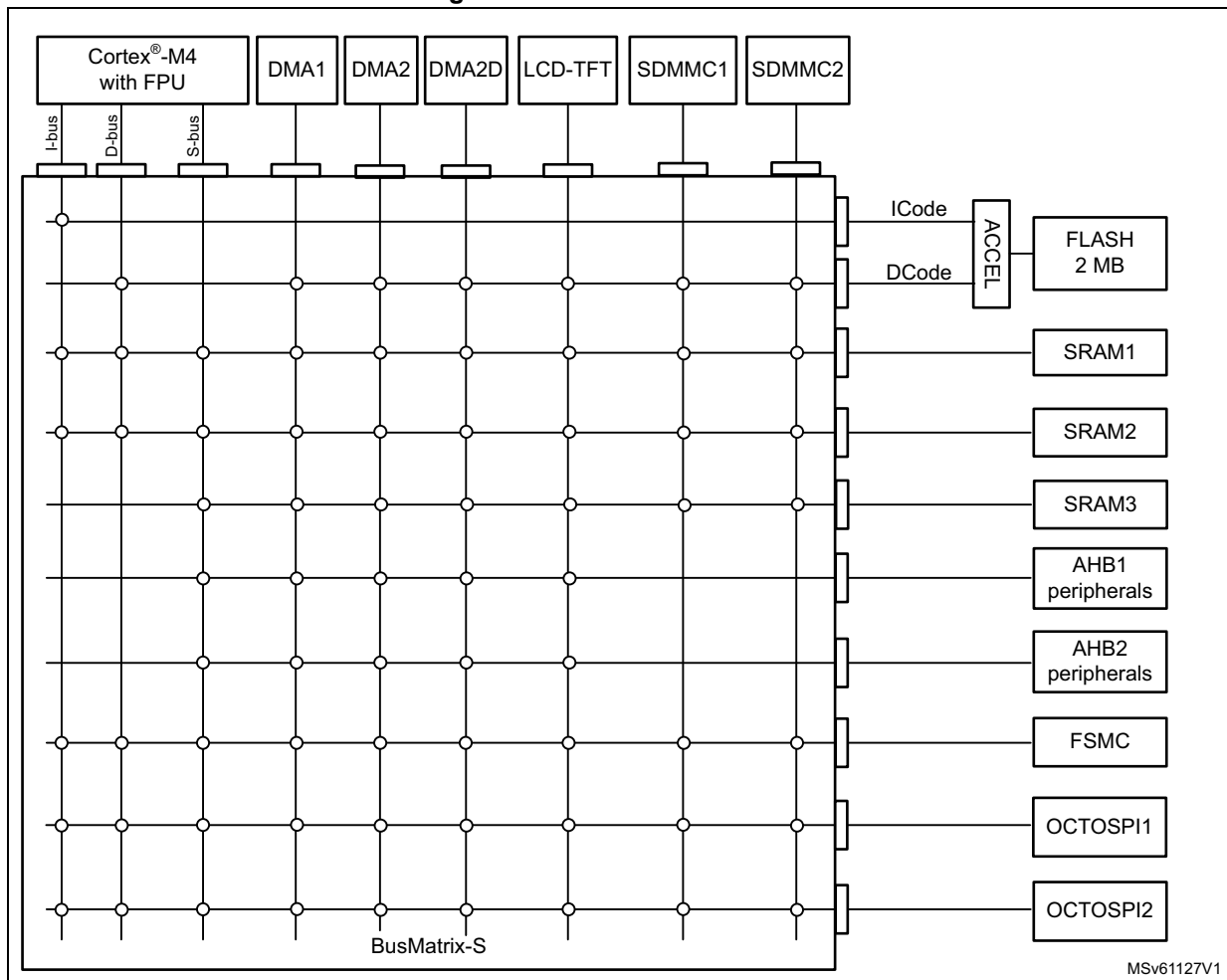
Each of the GPIO pins can be configured by software as output (push-pull or open-drain), as input (with or without pull-up or pull-down) or as peripheral alternate function. Most of the GPIO pins are shared with digital or analog alternate functions. Fast I/O toggling can be achieved thanks to their mapping on the AHB2 bus.

The I/Os alternate function configuration can be locked if needed following a specific sequence in order to avoid spurious writing to the I/Os registers.

3.12 Multi-AHB bus matrix

The 32-bit multi-AHB bus matrix interconnects all the masters (CPU, DMAs, DMA2D, SDMMC1, SDMMC2 and LCD-TFT) and the slaves (Flash memory, RAM, FSMC, OCTOSPI, AHB and APB peripherals). It also ensures a seamless and efficient operation even when several high-speed peripherals work simultaneously.

Figure 5. Multi-AHB bus matrix



3.13 Direct memory access controller (DMA)

The device embeds 2 DMAs. Refer to [Table 7: DMA implementation](#) for the features implementation.

Direct memory access (DMA) is used in order to provide a high-speed data transfer between peripherals and memory as well as from memory to memory. Data can be quickly moved by DMA without any CPU actions. This keeps the CPU resources free for other operations.

The two DMA controllers have 14 channels in total, each one dedicated to manage memory access requests from one or more peripherals. Each controller has an arbiter for handling the priority between DMA requests.

The DMA supports:

- 14 independently configurable channels (requests)
 - Each channel is connected to a dedicated hardware DMA request, a software trigger is also supported on each channel. This configuration is done by software.
- Priorities between requests from channels of one DMA are both software programmable (4 levels: very high, high, medium, low) or hardware programmable in case of equality (request 1 has priority over request 2, etc.)
- Independent source and destination transfer size (byte, half word, word), emulating packing and unpacking. Source/destination addresses must be aligned on the data size
- Support for circular buffer management
- 3 event flags (DMA half transfer, DMA transfer complete and DMA transfer error) logically ORed together in a single interrupt request for each channel
- Memory-to-memory transfer
- Peripheral-to-memory, memory-to-peripheral, and peripheral-to-peripheral transfers
- Access to Flash, SRAM, APB and AHB peripherals as source and destination
- Programmable number of data to be transferred: up to 65536

Table 7. DMA implementation

DMA features	DMA1	DMA2
Number of regular channels	7	7

3.14 DMA request router (DMAMux)

When a peripheral indicates a request for DMA transfer by setting its DMA request line, the DMA request is pending until it is served and the corresponding DMA request line is reset. The DMA request router allows to route the DMA control lines between the peripherals and the DMA controllers of the product.

An embedded multi-channel DMA request generator can be considered as one of such peripherals. The routing function is ensured by a multi-channel DMA request line multiplexer. Each channel selects a unique set of DMA control lines, unconditionally or synchronously with events on synchronization inputs.

For simplicity, the functional description is limited to DMA request lines. The other DMA control lines are not shown in figures or described in the text. The DMA request generator produces DMA requests following events on DMA request trigger inputs.

3.15 Chrom-ART Accelerator (DMA2D)

Chrom-ART Accelerator (DMA2D) is a graphic accelerator that offers an advanced bit blitting, row data copy and pixel format conversion. It supports the following functions:

- Rectangle filling with a fixed color
- Rectangle copy
- Rectangle copy with pixel format conversion
- Rectangle composition with blending and pixel format conversion.

Various image format coding are supported, from indirect 4 bpp color mode up to 32 bpp direct color. It embeds a dedicated memory to store color lookup tables.

An interrupt can be generated when an operation is complete or at a programmed watermark.

All the operations are fully automatized and are running independently from the CPU or the DMAs.

3.16 Interrupts and events

3.16.1 Nested vectored interrupt controller (NVIC)

The STM32L4Q5xx devices embed a nested vectored interrupt controller which is able to manage 16 priority levels, and to handle up to 95 maskable interrupt channels plus the 16 interrupt lines of the Cortex[®]-M4.

The NVIC benefits are the following:

- Closely coupled NVIC gives low latency interrupt processing
- Interrupt entry vector table address passed directly to the core
- Allows early processing of interrupts
- Processing of late arriving higher priority interrupts
- Support for tail chaining
- Processor state automatically saved
- Interrupt entry restored on interrupt exit with no instruction overhead

The NVIC hardware block provides flexible interrupt management features with minimal interrupt latency.

3.16.2 Extended interrupt/event controller (EXTI)

The extended interrupt/event controller consists of 36 edge detector lines used to generate interrupt/event requests and to wake up the system from the Stop mode. Each external line can be independently configured to select the trigger event (rising edge, falling edge, both) and can be masked independently.

A pending register maintains the status of the interrupt requests. The internal lines are connected to peripherals with wakeup from Stop mode capability. The EXTI can detect an external line with a pulse width shorter than the internal clock period. Up to 114 GPIOs can be connected to the 16 external interrupt lines.

3.17 Cyclic redundancy check calculation unit (CRC)

The CRC (cyclic redundancy check) calculation unit is used to get a CRC code using a configurable generator with polynomial value and size.

Among other applications, the CRC-based techniques are used to verify data transmission or storage integrity. In the scope of the EN/IEC 60335-1 standard, they offer a mean to verify the Flash memory integrity.

The CRC calculation unit helps to compute a signature of the software during runtime, which can be ulteriorly compared with a reference signature generated at link-time and which can be stored at a given memory location.

3.18 Flexible static memory controller (FSMC)

The flexible static memory controller (FSMC) includes two memory controllers:

- The NOR/PSRAM memory controller
- The NAND/memory controller

This memory controller is also named flexible memory controller (FMC).

The main features of the FSMC controller are the following:

- Interface with static-memory mapped devices including:
 - Static random access memory (SRAM)
 - NOR Flash memory/OneNAND Flash memory
 - PSRAM (four memory banks)
 - NAND Flash memory with ECC hardware to check up to 8 Kbytes of data
 - Ferroelectric RAM (FRAM)
- 8-,16- bit data bus width
- Independent chip select control for each memory bank
- Independent configuration for each memory bank
- Write FIFO
- The Maximum FMC_CLK frequency for synchronous accesses is HCLK/2.

LCD parallel interface

The FMC can be configured to interface seamlessly with most graphic LCD controllers. It supports the Intel 8080 and Motorola 6800 modes, and is flexible enough to adapt to specific LCD interfaces. This LCD parallel interface capability makes it easy to build cost effective graphic applications using LCD modules with embedded controllers or high-performance solutions using external controllers with dedicated acceleration.

3.19 Octo-SPI interface (OCTOSPI)

The OCTOSPI is a specialized communication interface targetting single, dual, quad or octal SPI memories. It can operate in any of the three following modes:

- Indirect mode: all the operations are performed using the OCTOSPI registers
- Status polling mode: the external memory status register is periodically read and an interrupt can be generated in case of flag setting
- Memory-mapped mode: the external memory is memory mapped and is seen by the system as if it were an internal memory supporting read and write operation

The OCTOSPI supports two frame formats:

- Classical frame format with command, address, alternate byte, dummy cycles and data phase over 1, 2, 4 or 8 data pins
- Hyperbus™ frame format

The OCTOSPI offers the following features:

- Three functional modes: indirect, status-polling, and memory-mapped
- Read and write support in memory-mapped mode
- Supports for single, dual, quad and octal communication
- Dual-quad mode, where 8 bits can be sent/received simultaneously by accessing two quad memories in parallel.
- SDR and DTR support
- Data strobe support
- Fully programmable opcode for both indirect and memory mapped mode
- Fully programmable frame format for both indirect and memory mapped mode
- Each of the five following phases can be configured independently (enable, length, single/dual/quad communication)
 - Instruction phase
 - Address phase
 - Alternate bytes phase
 - Dummy cycles phase
 - Data phase
- Hyperbus™ support
- Integrated FIFO for reception and transmission
- 8, 16, and 32-bit data accesses are allowed
- DMA channel for indirect mode operations

- Timeout management
- Interrupt generation on FIFO threshold, timeout, status match, operation complete, and access error

3.20 Octo-SPI IO manager (OCTOSPIIOM)

The OCTOSPI IO manager is a low level interface allowing:

- Efficient OCTOSPI pin assignment with a full IO Matrix (before alternate function map)
- Multiplexing single/dual/quad/octal SPI interface over the same bus

The OCTOSPI IO manager has the following features:

- Support up to two single/dual/quad/octal SPI Interface
- Support up to eight ports for pin assignment
- Fully programmable IO matrix for pin assignment by function (data/control/clock)
- Muxer for Single/Dual/Quad/Octal SPI interface multiplexing over the same bus

3.21 Analog-to-digital converter (ADC)

The device embeds two successive approximation analog-to-digital converters with the following features:

- 12-bit native resolution, with built-in calibration
- 5.33 Msps maximum conversion rate with full resolution
 - Down to 18.75 ns sampling time
 - Increased conversion rate for lower resolution (up to 8.88 Msps for 6-bit resolution)
- Up to 16 external channels
- 5 internal channels: internal reference voltage, temperature sensor, VBAT/3, DAC1_OUT1 and DAC1_OUT2 outputs
- One external reference pin is available on some package, allowing the input voltage range to be independent from the power supply
- Single-ended and differential mode inputs
- Low-power design
 - Capable of low-current operation at low conversion rate (consumption decreases linearly with speed)
 - Dual clock domain architecture: ADC speed independent from CPU frequency
- Highly versatile digital interface
 - Single-shot or continuous/discontinuous sequencer-based scan mode: 2 groups of analog signals conversions can be programmed to differentiate background and high-priority real-time conversions
 - Each ADC support multiple trigger inputs for synchronization with on-chip timers and external signals
 - Results stored into a data register or in RAM with DMA controller support
 - Data pre-processing: left/right alignment and per channel offset compensation
 - Built-in oversampling unit for enhanced SNR
 - Channel-wise programmable sampling time
 - Analog watchdog for automatic voltage monitoring, generating interrupts and trigger for selected timers
 - Hardware assistant to prepare the context of the injected channels to allow fast context switching

3.21.1 Temperature sensor

The temperature sensor (TS) generates a voltage V_{TS} that varies linearly with temperature. The temperature sensor is internally connected to the ADC1_IN17 input channels which is used to convert the sensor output voltage into a digital value.

The sensor provides good linearity but it has to be calibrated to obtain good overall accuracy of the temperature measurement. As the offset of the temperature sensor varies from chip to chip due to process variation, the uncalibrated internal temperature sensor is suitable for applications that detect temperature changes only.

To improve the accuracy of the temperature sensor measurement, each device is individually factory-calibrated by ST. The temperature sensor factory calibration data are stored by ST in the system memory area, accessible in read-only mode.

Table 8. Temperature sensor calibration values

Calibration value name	Description	Memory address
TS_CAL1	TS ADC raw data acquired at a temperature of 30 °C (± 5 °C), $V_{DDA} = V_{REF+} = 3.0$ V (± 10 mV)	0x1FFF 75A8 - 0x1FFF 75A9
TS_CAL2	TS ADC raw data acquired at a temperature of 130 °C (± 5 °C), $V_{DDA} = V_{REF+} = 3.0$ V (± 10 mV)	0x1FFF 75CA - 0x1FFF 75CB

3.21.2 Internal voltage reference (V_{REFINT})

The internal voltage reference (V_{REFINT}) provides a stable (bandgap) voltage output for the ADC and the comparators. The V_{REFINT} is internally connected to the ADC1_IN0 input channel. The precise voltage of V_{REFINT} is individually measured for each part by ST during production test and stored in the system memory area. It is accessible in read-only mode.

Table 9. Internal voltage reference calibration values

Calibration value name	Description	Memory address
VREFINT	Raw data acquired at a temperature of 30 °C (± 5 °C), $V_{DDA} = V_{REF+} = 3.0$ V (± 10 mV)	0x1FFF 75AA - 0x1FFF 75AB

3.21.3 V_{BAT} battery voltage monitoring

This embedded hardware enables the application to measure the V_{BAT} battery voltage using the internal ADC channel ADC1_IN18. As the V_{BAT} voltage may be higher than the V_{DDA} , and thus outside the ADC input range, the V_{BAT} pin is internally connected to a bridge divider by 3. As a consequence, the converted digital value is one third of the V_{BAT} voltage.

3.22 Digital to analog converter (DAC)

Two 12-bit buffered DAC channels can be used to convert digital signals into analog voltage signal outputs. The chosen design structure is composed of integrated resistor strings and an amplifier in inverting configuration.

This digital interface supports the following features:

- Up to two DAC output channels
- 8-bit or 12-bit output mode
- Buffer offset calibration (factory and user trimming)
- Left or right data alignment in 12-bit mode
- Synchronized update capability
- Noise-wave generation
- Triangular-wave generation

- Dual DAC channel independent or simultaneous conversions
- DMA capability for each channel
- External triggers for conversion
- Sample and hold low-power mode, with internal or external capacitor

The DAC channels are triggered through the timer update outputs that are also connected to different DMA channels.

3.23 Voltage reference buffer (VREFBUF)

The devices embed a voltage reference buffer which can be used as voltage reference for ADC, DACs and also as voltage reference for external components through the VREF+ pin.

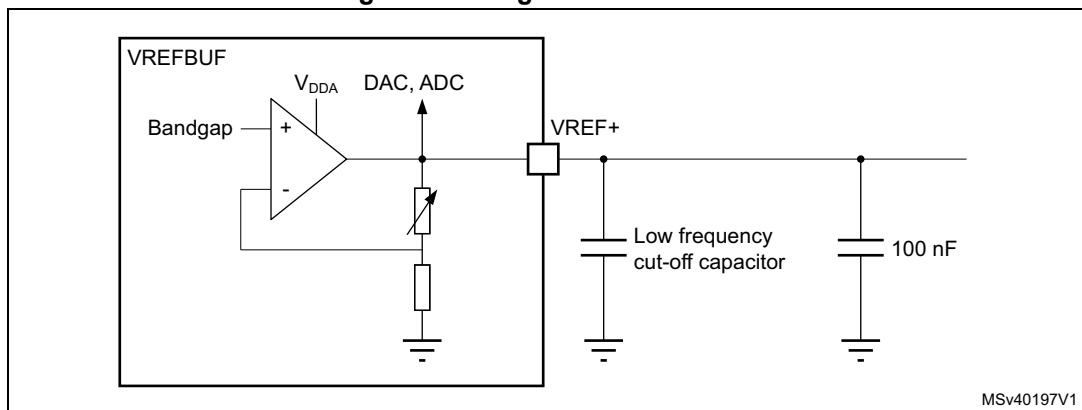
The internal voltage reference buffer supports two voltages:

- 2.042 V
- 2.5 V

An external voltage reference can be provided through the VREF+ pin when the internal voltage reference buffer is off.

The VREF+ pin is double-bonded with VDDA on some packages. In these packages the internal voltage reference buffer is not available.

Figure 6. Voltage reference buffer



MSv40197V1

3.24 Comparators (COMP)

The devices embed two rail-to-rail comparators with programmable reference voltage (internal or external), hysteresis and speed (low speed for low-power) and with selectable output polarity.

The reference voltage can be one of the following:

- External I/O
- DAC output channels
- Internal reference voltage or submultiple (1/4, 1/2, 3/4).

All comparators can wake up from Stop mode, generate interrupts and breaks for the timers and can also be combined into a window comparator.

3.25 Operational amplifier (OPAMP)

The devices embed two operational amplifiers with external or internal follower routing and PGA capability.

The operational amplifier features:

- Low input bias current
- Low offset voltage
- Low-power mode
- Rail-to-rail input

3.26 Digital filter for sigma-delta modulators (DFSDM)

The STM32L4Q5xx devices embed one DFSDM with two digital filters modules and four external input serial channels (transceivers) or alternately four internal parallel inputs support.

The DFSDM peripheral is dedicated to interface the external $\Sigma\Delta$ modulators to the microcontroller and then to perform digital filtering of the received data streams (which represent analog value on $\Sigma\Delta$ modulators inputs).

The DFSDM can also interface the PDM (pulse density modulation) microphones and perform PDM to PCM conversion and filtering in hardware. The DFSDM features optional parallel data stream inputs from microcontrollers memory (through DMA/CPU transfers into DFSDM).

The DFSDM transceivers support several serial interface formats (to support various $\Sigma\Delta$ modulators) and the DFSDM digital filter modules perform digital processing according to the user's selected filter parameters with up to 24-bit final ADC resolution.

The DFSDM peripheral supports:

- 4 multiplexed input digital serial channels:
 - Configurable SPI interface to connect various SD modulator(s)
 - Configurable Manchester coded 1 wire interface support
 - PDM (pulse density modulation) microphone input support
 - Maximum input clock frequency up to 20 MHz (10 MHz for Manchester coding)
 - Clock output for SD modulator(s): 0..20 MHz
- Alternative inputs from 8 internal digital parallel channels (up to 16-bit input resolution):
 - Internal sources: device memory data streams (DMA)
- 2 digital filter modules with adjustable digital signal processing:
 - Sinc^x filter: filter order/type (1..5), oversampling ratio (up to 1..1024)
 - Integrator: oversampling ratio (1..256)
- Up to 24-bit output data resolution, signed output data format
- Automatic data offset correction (offset stored in register by user)
- Continuous or single conversion
- Start-of-conversion triggered by:
 - Software trigger
 - Internal timers

- External events
- Start-of-conversion synchronously with first digital filter module (DFSDM0)
- Analog watchdog feature:
 - Low value and high-value data threshold registers
 - Dedicated configurable Sincx digital filter (order = 1..3, oversampling ratio = 1..32)
 - Input from final output data or from selected input digital serial channels
 - Continuous monitoring independently from standard conversion
- Short circuit detector to detect saturated analog input values (bottom and top range):
 - Up to 8-bit counter to detect 1..256 consecutive 0's or 1's on serial data stream
 - Monitoring continuously each input serial channel
- Break signal generation on analog watchdog event or on short circuit detector event
- Extremes detector:
 - Storage of minimum and maximum values of final conversion data
 - Refreshed by software
- DMA capability to read the final conversion data
- Interrupts: end of conversion, overrun, analog watchdog, short circuit, input serial channel clock absence
- “Regular” or “injected” conversions:
 - “Regular” conversions can be requested at any time or even in continuous mode without having any impact on the timing of “injected” conversions
 - “Injected” conversions for precise timing and with high conversion priority

3.27 Digital camera interface (DCMI)

The devices embed a camera interface that can connect with any camera modules and CMOS sensors through an 8-bit to 14-bit parallel interface in order to receive video data.

The camera interface can sustain a data transfer rate up to 54 Mbytes/s at 54 MHz. It features:

- Programmable polarity for the input pixel clock and synchronization signals
- Parallel data communication of 8-, 10-, 12- or 14-bit
- Supports 8-bit progressive video monochrome or raw bayer format, YCbCr 4:2:2 progressive video, RGB 565 progressive video or compressed data (like JPEG)
- Supports continuous mode or snapshot (a single frame) mode
- Capability to automatically crop the image.

3.28 LCD-TFT controller (LTDC)

The LCD-TFT display controller provides a 24-bit parallel digital RGB (red, green, blue) and delivers all signals to interface directly to a broad range of LCD and TFT panels with the following features:

- One display layer with dedicated FIFO (64 x 32-bit)
- Color look-up table (CLUT) up to 256 colors (256 x 24-bit) per layer
- Up to 8 input color formats selectable per layer
- Flexible blending between two layers using alpha value (per pixel or constant)
- Flexible programmable parameters for each layer
- Color keying (transparency color)
- Up to four programmable interrupt events

3.29 Touch sensing controller (TSC)

The touch sensing controller provides a simple solution to add capacitive sensing functionality to any application. A capacitive sensing technology is able to detect finger presence near an electrode that is protected from direct touch by a dielectric (glass, plastic or other). The capacitive variation introduced by the finger (or any conductive object) is measured using a proven implementation based on a surface charge transfer acquisition principle.

The touch sensing controller is fully supported by the STMTouch touch sensing firmware library which is free to use and allows touch sensing functionality to be implemented reliably in the end application.

The main features of the touch sensing controller are the following:

- Proven and robust surface charge transfer acquisition principle
- Supports up to 24 capacitive sensing channels
- Up to 3 capacitive sensing channels can be acquired in parallel offering a very good response time
- Spread spectrum feature to improve system robustness in noisy environments
- Full hardware management of the charge transfer acquisition sequence
- Programmable charge transfer frequency
- Programmable sampling capacitor I/O pin
- Programmable channel I/O pin
- Programmable max count value to avoid long acquisition when a channel is faulty
- Dedicated end of acquisition and max count error flags with interrupt capability
- One sampling capacitor for up to 3 capacitive sensing channels to reduce the system components
- Compatible with proximity, touchkey, linear and rotary touch sensor implementation
- Designed to operate with STMTouch touch sensing firmware library

Note: The number of capacitive sensing channels is dependent on the size of the packages and subject to I/O availability.

3.30 Random number generator (RNG)

All devices embed an RNG that delivers 32-bit random numbers generated by an integrated analog circuit.

3.31 Parallel synchronous slave interface (PSSI)

The PSSI is a generic synchronous 8-/16-bit parallel data input/output slave interface. It allows the transmitter to send a data valid signal to indicate when the data is valid, and the receiver to output a flow control signal to indicate when it is ready to sample the data.

The PSSI main features are:

- The PSSI shares most of the circuitry with the digital camera interface (DCMI). It thus cannot be used simultaneously with the DCMI.
- Slave mode operation
- 8- or 16-bit parallel data input or output
- 4-word (16-byte) FIFO
- Data enable (DE) alternate function input and Ready (RDY) alternate function output. When enabled, these signals can either allow the transmitter to indicate when the data is valid or the receiver to indicate when it is ready to sample the data, or both.

3.32 Advanced encryption standard hardware accelerator (AES)

The devices embed an AES hardware accelerator that can be used both to encipher and to decipher data using an AES algorithm.

The AES peripheral supports:

- Encryption/decryption using AES Rijndael block cipher algorithm
- NIST FIPS 197 compliant implementation of AES encryption/decryption algorithm
- 128-bit and 256-bit register for storing the encryption, decryption or derivation key (4x 32-bit registers)
- Electronic codebook (ECB), cipher block chaining (CBC), Counter mode (CTR), Galois Counter Mode (GCM), Galois Message Authentication Code mode (GMAC) and Cipher Message Authentication Code mode (CMAC) supported
- Key scheduler
- Key derivation for decryption
- 128-bit data block processing
- 128-bit, 256-bit key length
- 1x32-bit INPUT buffer and 1x32-bit OUTPUT buffer
- Register access supporting 32-bit data width only
- One 128-bit Register for the initialization vector when AES is configured in CBC mode or for the 32-bit counter initialization when CTR mode is selected, GCM mode or CMAC mode
- Automatic data flow control with support of direct memory access (DMA) using 2 channels, one for incoming data, and one for outgoing data
- Suspend a message if another message with a higher priority needs to be processed.

3.33 HASH hardware accelerator (HASH)

The hash processor is a fully compliant implementation of the secure hash algorithm (SHA-1, SHA-224, SHA-256), the MD5 (message-digest algorithm 5) hash algorithm and the HMAC (keyed-hash message authentication code) algorithm suitable for a variety of applications.

It computes a message digest (160 bits for the SHA-1 algorithm, 256 bits for the SHA-256 algorithm and 224 bits for the SHA-224 algorithm, 128 bits for the MD5 algorithm) for messages of up to (264 - 1) bits, while the HMAC algorithms provide a way of authenticating messages by means of hash functions. The HMAC algorithms consist in calling the SHA-1, SHA-224, SHA-256 or MD5 hash function twice.

3.34 Public key accelerator (PKA)

The public key accelerator (PKA) is intended for the computation of cryptographic public key primitives, specifically those related to RSA, Diffie-Hellmann or ECC (elliptic curve cryptography) over GF(p) (Galois fields). To achieve high performance at a reasonable cost, these operations are executed in the Montgomery domain.

All needed computations are performed within the accelerator, so no further hardware/software elaboration is needed to process the inputs or the outputs.

PKA main features:

- Acceleration of RSA, DH and ECC over GF(p) operations, based on the Montgomery method for fast modular multiplications. More specifically:
 - RSA modular exponentiation, RSA Chinese remainder theorem (CRT) exponentiation
 - ECC scalar multiplication, point on curve check
 - ECDSA signature generation and verification
- Capability to handle operands up to 3136 bits for RSA/DH and 640 bits for ECC.
- Arithmetic and modular operations such as addition, subtraction, multiplication, modular reduction, modular inversion, comparison, and Montgomery multiplication.
- Built-in Montgomery domain inward and outward transformations
- AMBA AHB slave peripheral, accessible through 32-bit word single accesses only (otherwise, for writes, an AHB bus error is generated, and write accesses are ignored).

3.35 Timers and watchdogs

The devices include two advanced control timers, up to nine general-purpose timers, two basic timers, two low-power timers, two watchdog timers and a SysTick timer.

[Table 10](#) below compares the features of the advanced control, general-purpose and basic timers.

Table 10. Timer feature comparison

Timer type	Timer	Counter resolution	Counter type	Prescaler factor	DMA request generation	Capture/compare channels	Complementary outputs
Advanced control	TIM1, TIM8	16-bit	Up, down, Up/down	Any integer between 1 and 65536	Yes	4	3
General-purpose	TIM2, TIM5	32-bit	Up, down, Up/down	Any integer between 1 and 65536	Yes	4	No
General-purpose	TIM3, TIM4	16-bit	Up, down, Up/down	Any integer between 1 and 65536	Yes	4	No
General-purpose	TIM15	16-bit	Up	Any integer between 1 and 65536	Yes	2	1
General-purpose	TIM16, TIM17	16-bit	Up	Any integer between 1 and 65536	Yes	1	1
Basic	TIM6, TIM7	16-bit	Up	Any integer between 1 and 65536	Yes	0	No

3.35.1 Advanced-control timer (TIM1, TIM8)

The advanced-control timers can each be seen as a three-phase PWM multiplexed on six channels. They have complementary PWM outputs with programmable inserted dead-times. They can also be seen as complete general-purpose timers.

The four independent channels can be used for:

- Input capture
- Output compare
- PWM generation (edge or center-aligned modes) with full modulation capability (0-100%)
- One-pulse mode output

In debug mode, the advanced-control timer counter can be frozen and the PWM outputs disabled in order to turn off any power switches driven by these outputs.

Many features are shared with the general-purpose TIMx timers (described in [Section 3.35.2](#)) using the same architecture, so the advanced-control timers can work together with the TIMx timers via the Timer Link feature for synchronization or event chaining.

3.35.2 General-purpose timers (TIM2, TIM3, TIM4, TIM5, TIM15, TIM16, TIM17)

There are up to seven synchronizable general-purpose timers embedded in the devices (see [Table 10](#) for differences).

Each general-purpose timer can be used to generate PWM outputs, or act as a simple time base.

- TIM2, TIM3, TIM4 and TIM5

They are full-featured general-purpose timers:

- TIM2 and TIM5 have a 32-bit auto-reload up/downcounter and 32-bit prescaler
- TIM3 and TIM4 have 16-bit auto-reload up/downcounter and 16-bit prescaler.

These timers feature four independent channels for input capture/output compare, PWM or one-pulse mode output. They can work together, or with the other general-purpose timers via the Timer Link feature for synchronization or event chaining.

The counters can be frozen in debug mode.

All have independent DMA request generation and support quadrature encoders.

- TIM15, 16 and 17

They are general-purpose timers with mid-range features:

They have 16-bit auto-reload upcounters and 16-bit prescalers.

- TIM15 has two channels and one complementary channel
- TIM16 and TIM17 have one channel and one complementary channel

All channels can be used for input capture/output compare, PWM or one-pulse mode output.

The timers can work together via the Timer Link feature for synchronization or event chaining. The timers have independent DMA request generation.

The counters can be frozen in debug mode.

3.35.3 Basic timers (TIM6 and TIM7)

The basic timers are mainly used for DAC trigger generation. They can also be used as generic 16-bit timebases.

3.35.4 Low-power timer (LPTIM1 and LPTIM2)

The devices embed two low-power timers. These timers have an independent clock and are running in Stop mode if they are clocked by LSE, LSI or an external clock. They are able to wake up the system from Stop mode.

LPTIM1 and LPTIM2 can be active in Stop 0, Stop 1 and Stop 2 modes.

This low-power timer supports the following features:

- 16-bit up counter with 16-bit autoreload register
- 16-bit compare register
- Configurable output: pulse, PWM
- Continuous/ one shot mode
- Selectable software/hardware input trigger
- Selectable clock source
 - Internal clock sources: LSE, LSI, HSI16 or APB clock
 - External clock source over LPTIM input (working even with no internal clock source running, used by pulse counter application).
- Programmable digital glitch filter
- Encoder mode (LPTIM1 only)
- Repetition counter.

3.35.5 Independent watchdog (IWDG)

The independent watchdog is based on a 12-bit downcounter and an 8-bit prescaler. It is clocked from an independent 32 kHz internal RC (LSI) and as it operates independently from the main clock, it can operate in Stop and Standby modes. It can be used either as a watchdog to reset the device when a problem occurs, or as a free running timer for application timeout management. It is hardware or software configurable through the option bytes. The counter can be frozen in debug mode.

3.35.6 Window watchdog (WWDG)

The window watchdog is based on a 7-bit downcounter that can be set as free running. It can be used as a watchdog to reset the device when a problem occurs. It is clocked from the main clock. It has an early warning interrupt capability and the counter can be frozen in debug mode.

3.35.7 SysTick timer

This timer is dedicated to real-time operating systems, but could also be used as a standard down counter. It features:

- A 24-bit down counter
- Autoreload capability
- Maskable system interrupt generation when the counter reaches 0.
- Programmable clock source

3.36 Real-time clock (RTC) and backup registers

The RTC is an independent BCD timer/counter. It supports the following features:

- Calendar with subsecond, seconds, minutes, hours (12 or 24 format), week day, date, month, year, in BCD (binary-coded decimal) format
- Automatic correction for 28, 29 (leap year), 30, and 31 days of the month
- Two programmable alarms
- On-the-fly correction from 1 to 32767 RTC clock pulses. This can be used to synchronize it with a master clock
- Reference clock detection: a more precise second source clock (50 or 60 Hz) can be used to enhance the calendar precision
- Digital calibration circuit with 0.95 ppm resolution, to compensate for quartz crystal inaccuracy
- Three anti-tamper detection pins with programmable filter
- Timestamp feature which can be used to save the calendar content. This function can be triggered by an event on the timestamp pin, or by a tamper event, or by a switch to VBAT mode
- 17-bit auto-reload wakeup timer (WUT) for periodic events with programmable resolution and period

The RTC and the 32 backup registers are supplied through a switch that takes power either from the V_{DD} supply when present or from the VBAT pin.

The backup registers are 32-bit registers used to store 128 bytes of user application data when VDD power is not present. They are not reset by a system or power reset, or when the device wakes up from standby or Shutdown mode.

The RTC clock sources can be:

- A 32.768 kHz external crystal (LSE)
- An external resonator or oscillator (LSE)
- The internal low-power RC oscillator (LSI, with typical frequency of 32 kHz)
- The high-speed external clock (HSE) divided by 32

The RTC is functional in VBAT mode and in all low-power modes when it is clocked by the LSE. When clocked by the LSI, the RTC is not functional in VBAT mode, but is functional in all low-power modes except Shutdown mode.

All RTC events (alarm, wake-up timer, Timestamp or Tamper) can generate an interrupt and wakeup the device from the low-power modes.

3.37 Inter-integrated circuit interface (I2C)

The device embeds four I2C. Refer to [Table 11: I2C implementation](#) for the features implementation.

The I²C bus interface handles communications between the microcontroller and the serial I²C bus. It controls all I²C bus-specific sequencing, protocol, arbitration and timing.

The I2C peripheral supports:

- I²C-bus specification and user manual rev. 5 compatibility:
 - Slave and master modes, multimaster capability
 - Standard-mode (Sm), with a bitrate up to 100 kbit/s
 - Fast-mode (Fm), with a bitrate up to 400 kbit/s
 - Fast-mode plus (Fm+), with a bitrate up to 1 Mbit/s and 20 mA output drive I/Os
 - 7-bit and 10-bit addressing mode, multiple 7-bit slave addresses
 - Programmable setup and hold times
 - Optional clock stretching
- System management bus (SMBus) specification rev 2.0 compatibility:
 - Hardware PEC (packet error checking) generation and verification with ACK control
 - Address resolution protocol (ARP) support
 - SMBus alert
- Power system management protocol (PMBusTM) specification rev 1.1 compatibility
- Independent clock: a choice of independent clock sources allowing the I2C communication speed to be independent from the PCLK reprogramming. Refer to [Figure 4: Clock tree](#)
- Wakeup from Stop mode on address match
- Programmable analog and digital noise filters
- 1-byte buffer with DMA capability

Table 11. I2C implementation

I2C features ⁽¹⁾	I2C1	I2C2	I2C3	I2C4
Standard-mode (up to 100 kbit/s)	X	X	X	X
Fast-mode (up to 400 kbit/s)	X	X	X	X
Fast-mode Plus with 20 mA output drive I/Os (up to 1 Mbit/s)	X	X	X	X
Programmable analog and digital noise filters	X	X	X	X
SMBus/PMBus hardware support	X	X	X	X
Independent clock	X	X	X	X
Wakeup from Stop 0, Stop 1 mode on address match	X	X	X	X
Wakeup from Stop 2 mode on address match	-	-	X	-

1. X: supported

3.38 Universal synchronous/asynchronous receiver transmitter (USART)

The devices have three embedded universal synchronous receiver transmitters (USART1, USART2 and USART3) and two universal asynchronous receiver transmitters (UART4, UART5).

These interfaces provide asynchronous communication, IrDA SIR ENDEC support, multiprocessor communication mode, single-wire half-duplex communication mode and have LIN master/slave capability. They provide hardware management of the CTS and RTS signals, and RS485 driver enable. They are able to communicate at speeds of up to 10 Mbit/s.

The USART1, USART2 and USART3 also provide a Smartcard mode (ISO 7816 compliant) and an SPI-like communication capability.

All USART have a clock domain independent from the CPU clock, allowing the USARTx (x=1,2,3,4,5) to wake up the MCU from Stop mode using baudrates up to 200 Kbaud. The wake up events from Stop mode are programmable and can be:

- Start bit detection
- Any received data frame
- A specific programmed data frame

All USART interfaces can be served by the DMA controller.

Table 12. USART/UART/LPUART features

USART modes/features ⁽¹⁾	USART1	USART2	USART3	UART4	UART5	LPUART1
Hardware flow control for modem	X	X	X	X	X	X
Continuous communication using DMA	X	X	X	X	X	X
Multiprocessor communication	X	X	X	X	X	X
Synchronous mode	X	X	X	-	-	-
Smartcard mode	X	X	X	-	-	-
Single-wire half-duplex communication	X	X	X	X	X	X
IrDA SIR ENDEC block	X	X	X	X	X	-
LIN mode	X	X	X	X	X	-
Dual clock domain	X	X	X	X	X	X
Wakeup from Stop 0 / Stop 1 modes	X	X	X	X	X	X
Wakeup from Stop 2 mode	-	-	-	-	-	X
Receiver timeout interrupt	X	X	X	X	X	-
Modbus communication	X	X	X	X	X	-
Auto baud rate detection	X (4 modes)					-
Driver enable	X	X	X	X	X	X
LPUART/USART data length	7, 8 and 9 bits					

1. X = supported.

3.39 Low-power universal asynchronous receiver transmitter (LPUART)

The devices embed one low-power UART. The LPUART supports asynchronous serial communication with minimum power consumption. It supports half-duplex single-wire communication and modem operations (CTS/RTS). It allows multiprocessor communication.

The LPUART has a clock domain independent from the CPU clock, and can wakeup the system from Stop mode using baudrates up to 220 Kbaud. The wake up events from Stop mode are programmable and can be:

- Start bit detection
- Any received data frame
- A specific programmed data frame

Only a 32.768 kHz clock (LSE) is needed to allow LPUART communication up to 9600 baud. Therefore, even in Stop mode, the LPUART can wait for an incoming frame while having an extremely low energy consumption. Higher speed clock can be used to reach higher baudrates.

The LPUART interface can be served by the DMA controller.

3.40 Serial peripheral interface (SPI)

Three SPI interfaces allow communication up to slave modes, in half-duplex, full-duplex and simplex modes. The 3-bit prescaler gives eight master mode frequencies and the frame size is configurable from 4 bits to 16 bits. The SPI interfaces support NSS pulse mode, TI mode and hardware CRC calculation.

All SPI interfaces can be served by the DMA controller.

3.41 Serial audio interfaces (SAI)

The devices embed two SAI. Refer to [Table 13: SAI implementation](#) for the features implementation. The SAI bus interface handles communications between the microcontroller and the serial audio protocol.

The SAI peripheral supports:

- Two independent audio sub-blocks which can be transmitters or receivers with their respective FIFO.
- 8-word integrated FIFOs for each audio sub-block.
- Synchronous or asynchronous mode between the audio sub-blocks.
- Master or slave configuration independent for both audio sub-blocks.
- Clock generator for each audio block to target independent audio frequency sampling when both audio sub-blocks are configured in master mode.
- Data size configurable: 8-, 10-, 16-, 20-, 24-, 32-bit.
- Peripheral with large configurability and flexibility allowing to target as example the following audio protocol: I2S, LSB or MSB-justified, PCM/DSP, TDM, AC'97 and SPDIF out.

- Up to 16 slots available with configurable size and with the possibility to select which ones are active in the audio frame.
- Number of bits by frame may be configurable.
- Frame synchronization active level configurable (offset, bit length, level).
- First active bit position in the slot is configurable.
- LSB first or MSB first for data transfer.
- Mute mode.
- Stereo/Mono audio frame capability.
- Communication clock strobing edge configurable (SCK).
- Error flags with associated interrupts if enabled respectively.
 - Overrun and underrun detection.
 - Anticipated frame synchronization signal detection in slave mode.
 - Late frame synchronization signal detection in slave mode.
 - Codec not ready for the AC'97 mode in reception.
- Interruption sources when enabled:
 - Errors.
 - FIFO requests.
- DMA interface with two dedicated channels to handle access to the dedicated integrated FIFO of each SAI audio sub-block.

Table 13. SAI implementation

SAI features ⁽¹⁾	SAI1	SAI2
I2S, LSB or MSB-justified, PCM/DSP, TDM, AC'97	X	X
Mute mode	X	X
Stereo/Mono audio frame capability.	X	X
16 slots	X	X
Data size configurable: 8-, 10-, 16-, 20-, 24-, 32-bit	X	X
FIFO size	X (8 Word)	X (8 Word)
SPDIF	X	X
PDM	X	-

1. X: supported

3.42 Secure digital input/output and MultiMediaCards Interface (SDMMC)

The devices embed two SDMMC host interfaces. Each SDMMC host interface embeds a dedicated DMA controller allowing high-speed transfers between the interface and the other AHB slaves.

The SD/SDIO, MultiMediaCard (MMC) host interface (SDMMC) provides an interface between the AHB bus and SD memory cards, SDIO cards and MMC devices.

The SDMMC features include the following:

- Full compliance with MultiMediaCard System Specification Version 4.51. Card support for three different databus modes: 1-bit (default), 4-bit and 8-bit
- Full compatibility with previous versions of MultiMediaCards (backward compatibility)
- Full compliance with SD Memory Card Specifications Version 4.1. (SDR104 SDMMC_CK speed limited to maximum allowed IO speed, SPI mode and UHS-II mode not supported)
- Full compliance with SDIO Card Specification Version 4.0: card support for two different databus modes: 1-bit (default) and 4-bit. (SDR104 SDMMC_CK speed limited to maximum allowed IO speed, SPI mode and UHS-II mode not supported)
- Data transfer up to 104 Mbyte/s for the 8-bit mode (depending maximum allowed IO speed)
- Data and command output enable signals to control external bidirectional drivers.

3.43 Controller area network (CAN)

The CAN is compliant with the 2.0A and B (active) specifications with a bit rate of up to 1 Mbit/s. They can receive and transmit standard frames with 11-bit identifiers as well as extended frames with 29-bit identifiers. The CAN has three transmit mailboxes, two receive FIFOs with three stages and 28 shared scalable filter banks (all of them can be used even if one CAN is used). 256 bytes of SRAM are allocated.

The CAN peripheral supports:

- CAN protocol version 2.0 A, B Active
- Bit rates of up to 1 Mbit/s
- Transmission
 - Three transmit mailboxes
 - Configurable transmit priority
- Reception
 - Two receive FIFOs with three stages
 - Scalable filter banks: 28 filter banks
 - Identifier list feature
 - Configurable FIFO overrun
- Time-triggered communication option
 - Disable automatic retransmission mode
 - 16-bit free running timer
 - Time Stamp sent in last two data bytes
- Management
 - Maskable interrupts
 - Software-efficient mailbox mapping at a unique address space

3.44 Universal serial bus on-the-go full-speed (OTG_FS)

The devices embed an USB OTG full-speed device/host/OTG peripheral with integrated transceivers. The USB OTG FS peripheral is compliant with the USB 2.0 specification and with the OTG 2.0 specification. It has software-configurable endpoint setting and supports suspend/resume.

The USB OTG controller requires a dedicated 48 MHz clock that can be provided by the internal multispeed oscillator (MSI) automatically trimmed by 32.768 kHz external oscillator (LSE). This allows to use the USB device without external high speed crystal (HSE).

The major features are:

- Combined Rx and Tx FIFO size of 1.25 Kbytes with dynamic FIFO sizing
- Supports the session request protocol (SRP) and host negotiation protocol (HNP)
- One bidirectional control endpoint + 5 IN endpoints + 5 OUT endpoints
- Eight host channels with periodic OUT support
- HNP/SNP/IP inside (no need for any external resistor)
- Software configurable to OTG 1.3 and OTG 2.0 modes of operation
- OTG 2.0 Supports ADP (Attach detection Protocol)
- USB 2.0 LPM (Link Power Management) support
- Battery charging specification revision 1.2 support
- Internal FS OTG PHY support

For OTG/Host modes, a power switch is needed in case bus-powered devices are connected.

The synchronization for this oscillator can also be taken from the USB data stream itself (SOF signalization) which allows crystal less operation.

3.45 Development support

3.45.1 Serial wire JTAG debug port (SWJ-DP)

The Arm® SWJ-DP interface is embedded, and is a combined JTAG and serial wire debug port that enables either a serial wire debug or a JTAG probe to be connected to the target.

Debug is performed using two pins only instead of five required by the JTAG (JTAG pins could be re-used as GPIO with alternate function): the JTAG TMS and TCK pins are shared with SWDIO and SWCLK, respectively, and a specific sequence on the TMS pin is used to switch between JTAG-DP and SW-DP.

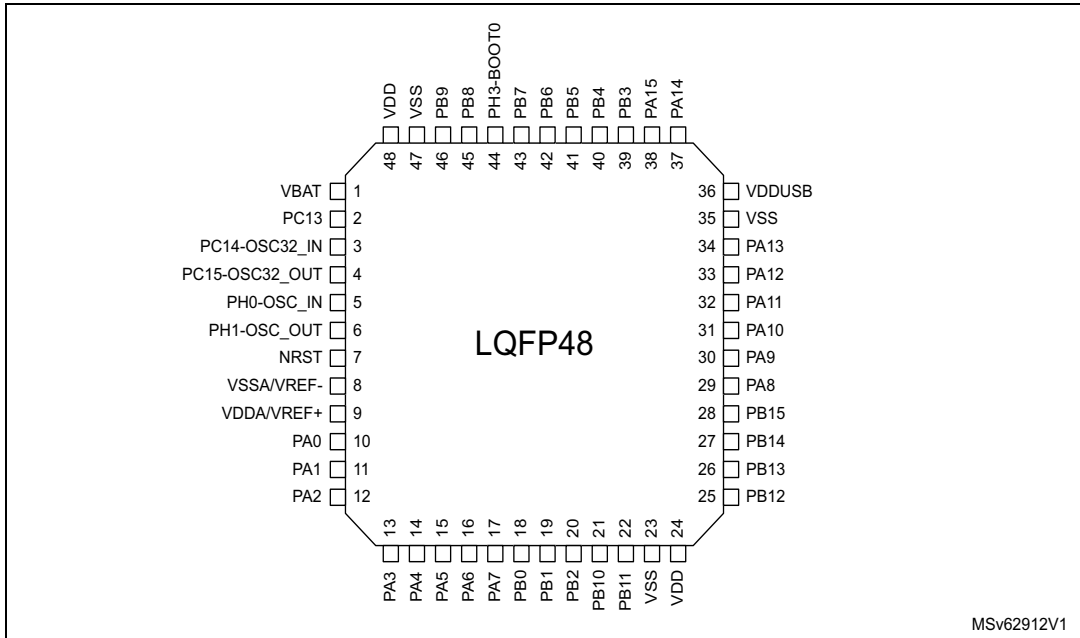
3.45.2 Embedded Trace Macrocell™

The Arm® Embedded Trace Macrocell provides a greater visibility of the instruction and data flow inside the CPU core by streaming compressed data at a very high rate from the devices through a small number of ETM pins to an external hardware trace port analyzer (TPA) device. Real-time instruction and data flow activity being recorded and then formatted for display on the host computer that runs the debugger software. TPA hardware is commercially available from common development tool vendors.

The Embedded Trace Macrocell operates with third party debugger software tools.

4 Pinouts and pin description

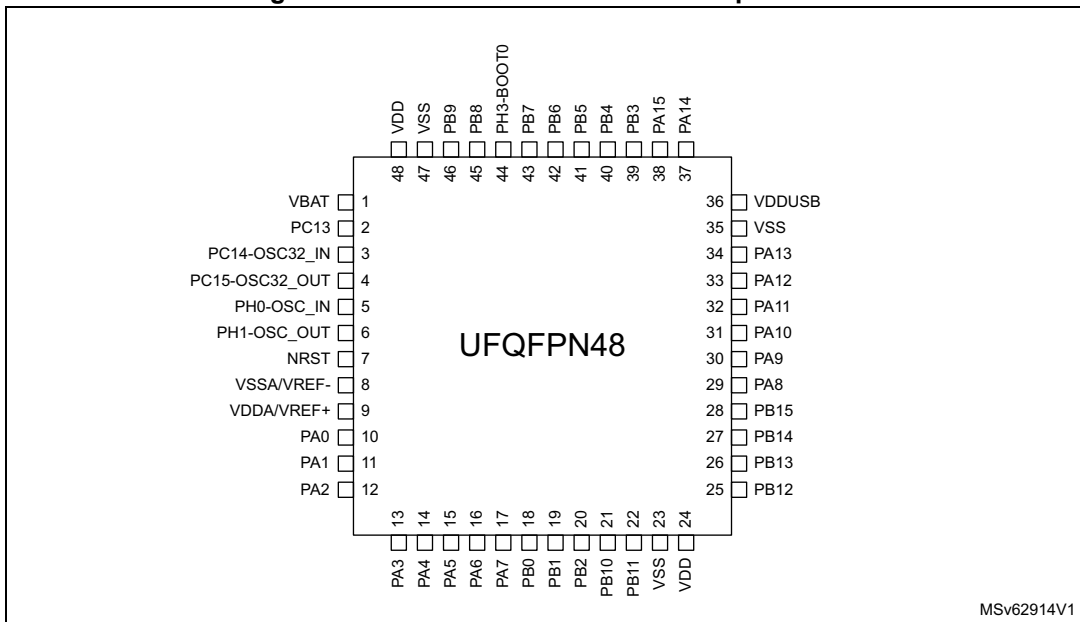
Figure 7. STM32L4Q5Cxxx LQFP48 pinout⁽¹⁾



MSv62912V1

1. The above figure shows the package top view.

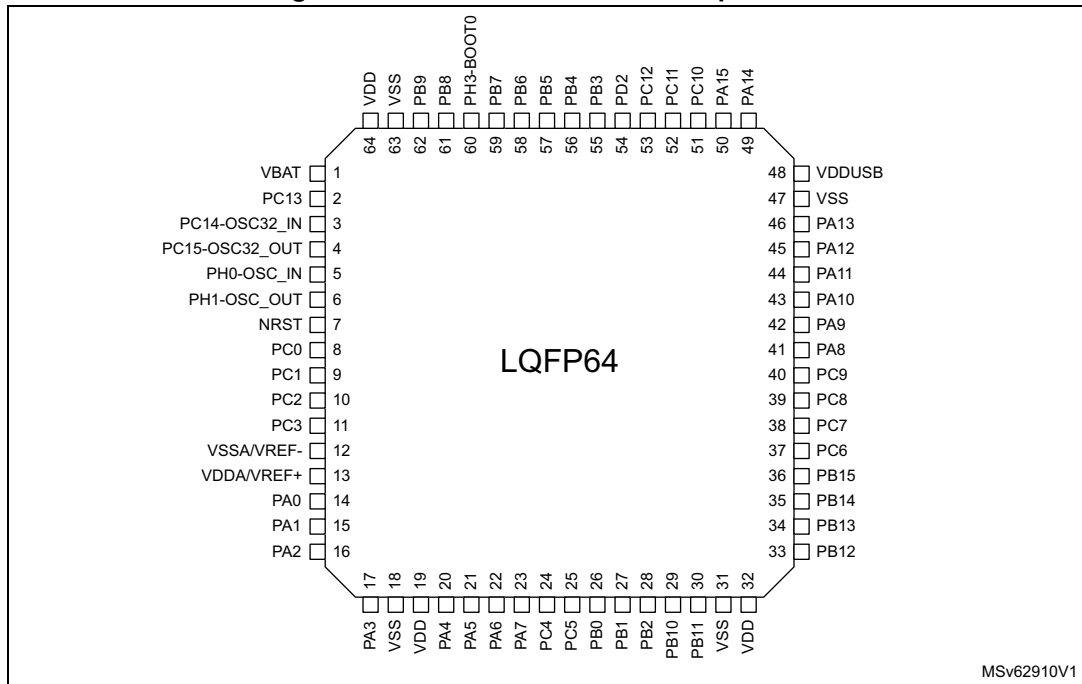
Figure 8. STM32L4Q5Cxxx UFQFPN48 pinout⁽¹⁾



MSv62914V1

1. The above figure shows the package top view.

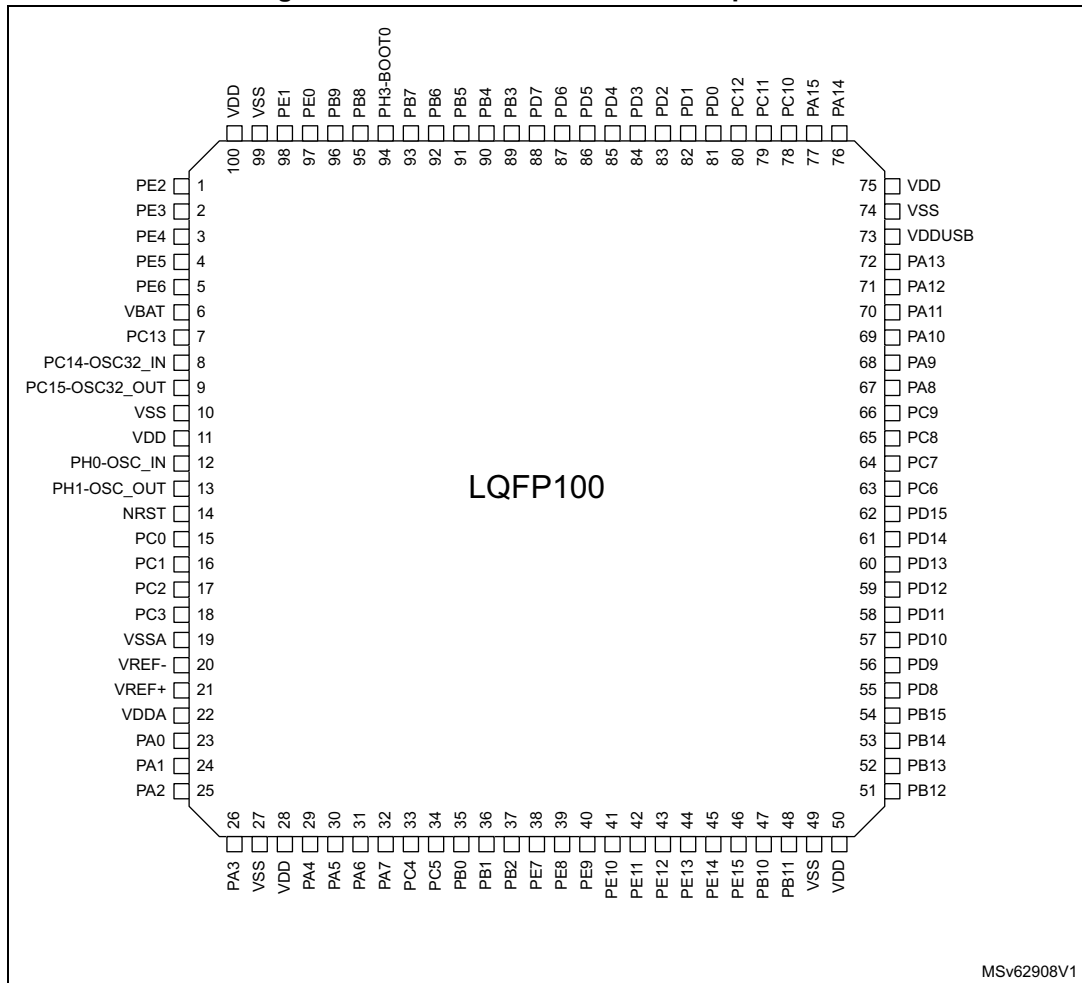
Figure 9. STM32L4Q5Rxxx LQFP64 pinout⁽¹⁾



MSv62910V1

1. The above figure shows the package top view.

Figure 10. STM32L4Q5Vxxx LQFP100 pinout⁽¹⁾



1. The above figure shows the package top view.

Figure 11. STM32L4Q5Vxxx WLCSP100 ballout⁽¹⁾

	1	2	3	4	5	6	7	8	9	10
A	VDDUSB	PA15	PD1	VDD	PG10	VDDIO2	PB6	PB9	VSS	VDD
B	VSS	PA14	PD0	PD4	PG9	PG12	PB5	PB8	PE2	PE3
C	PA12	PA13	PC11	PC12	PD7	PB3	PB4	PE4	PC13	VBAT
D	PA11	PA10	PA9	PC10	PD6	PG11	PB7	PE5	VSS	PC14-OSC32_IN
E	PC8	PC9	PA8	PD2	PD5	PH3-BOOT0	PE6	NRST	VDD	PC15-OSC32_OUT
F	VDD	PC6	PC7	PD15	PB2	PA4	PC3	PC1	PC0	PH0-OSC_IN
G	PD10	PD9	PD14	PE13	PE12	PA5	VREF+	VREF-	PA0	PH1-OSC_OUT
H	PB15	PB14	PD8	PE15	PE10	PC4	PA2	PA1	VSSA	PC2
J	PB12	PB13	PB11	PE14	PE9	PB0	PA7	VDD	PA3	VDDA
K	VDD	VSS	PB10	PE11	PE8	PE7	PB1	PC5	PA6	VSS

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1. The above figure shows the package top view.

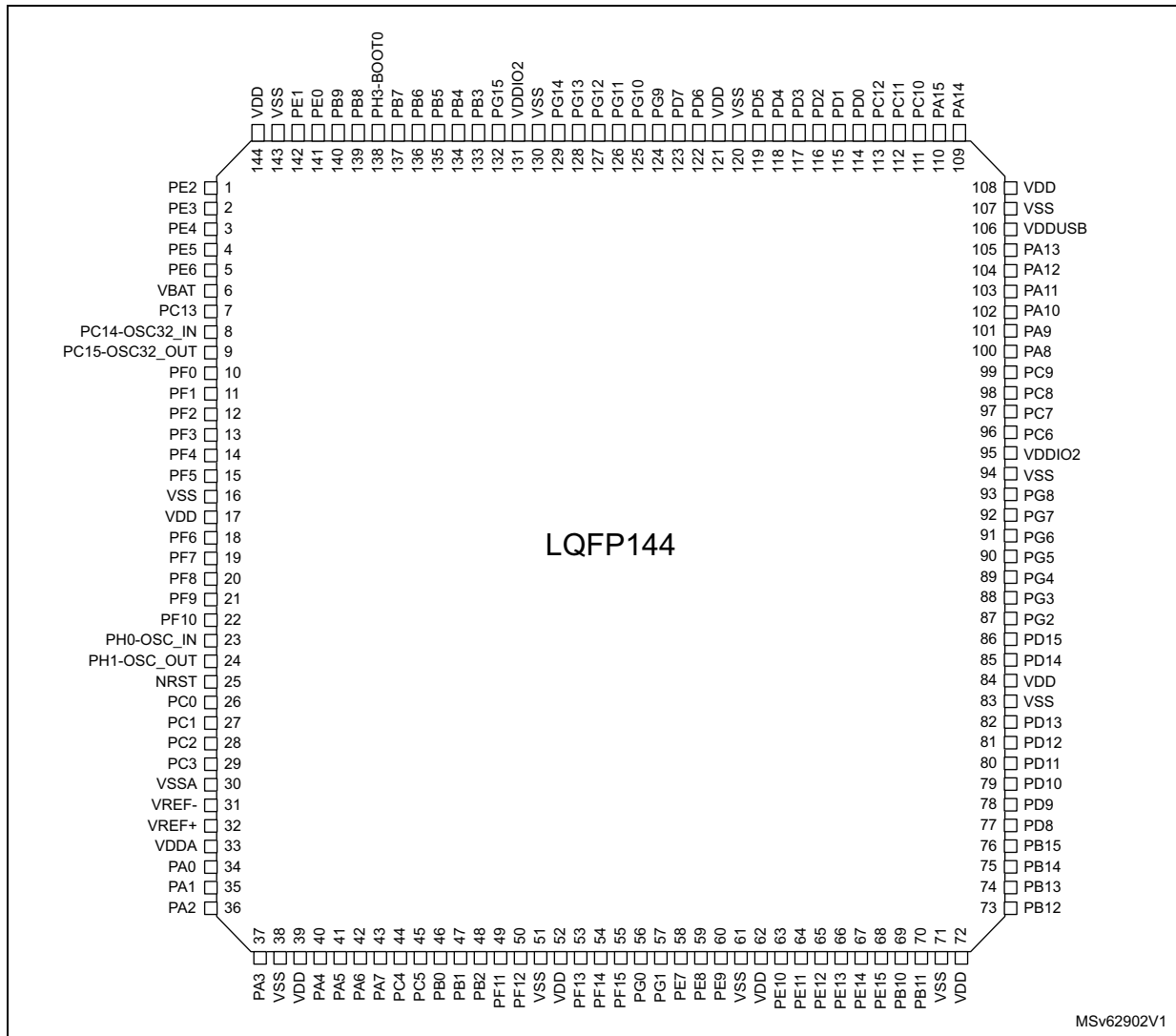
Figure 12. STM32L4Q5Qxxx UFBGA132 ballout⁽¹⁾

	1	2	3	4	5	6	7	8	9	10	11	12				
A	PE3	PE1	PB8	PH3-BOOT0	PD7	PD5	PB4	PB3	PA15	PA14	PA13	PA12				
B	PE4	PE2	PB9	PB7	PB6	PD6	PD4	PD3	PD1	PC12	PC10	PA11				
C	PC13	PE5	PE0	VDD	PB5	PG14	PG13	PD2	PD0	PC11	VDDUSB	PA10				
D	PC14-OSC32_IN	PE6	VSS	PF2	PF1	PF0	PG12	PG10	PG9	PA9	PA8	PC9				
E	PC15-OSC32_OUT	VBAT	VSS	PF3	<table border="1" style="margin: auto;"> <tr> <td>VSS</td> <td>VSS</td> </tr> <tr> <td>VDD</td> <td>VDDIO2</td> </tr> </table>				VSS	VSS	VDD	VDDIO2	PG5	PC8	PC7	PC6
VSS	VSS															
VDD	VDDIO2															
F	PH0-OSC_IN	VSS	PF4	PF5	PG3	PG4	VSS	VSS								
G	PH1-OSC_OUT	VDD	PG11	PG6	PG1	PG2	VDD	VDD								
H	PC0	NRST	VDD	PG7	PG0	PD15	PD14	PD13								
J	VSSA/VREF-	PC1	PC2	PA4	PA7	PG8	PF12	PF14	PF15	PD12	PD11	PD10				
K	PG15	PC3	PA2	PA5	PC4	PF11	PF13	PD9	PD8	PB15	PB14	PB13				
L	VREF+	PA0	PA3	PA6	PC5	PB2	PE8	PE10	PE12	PB10	PB11	PB12				
M	VDDA	PA1	OPAMP1_VI NM	OPAMP2_VI NM	PB0	PB1	PE7	PE9	PE11	PE13	PE14	PE15				

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1. The above figure shows the package top view.

Figure 13. STM32L4Q5Zxxx LQFP144 pinout⁽¹⁾



1. The above figure shows the package top view.

Figure 14. STM32L4Q5Axxx UFBGA169 ballout⁽¹⁾

	1	2	3	4	5	6	7	8	9	10	11	12	13
A	PH10	PH2	VDD	PE0	PB4	PB3	VSS	VDD	PA15	PA14	PA13	PI0	PH14
B	PI9	PI7	VSS	PE1	PB5	VDDIO2	PG9	PD0	PI6	PI2	PI1	PH15	PH12
C	VDD	VSS	PH11	PB8	PB6	PG15	PD4	PD1	PH13	PI3	PI8	VSS	VDD
D	PE4	PE3	PE2	PB9	PB7	PG10	PD5	PD2	PC10	PI4	PH9	PH7	PA12
E	PC13	VBAT	PE6	PE5	PH3-BOOT0	PG11	PD6	PD3	PC11	PI5	PH6	VDDUSB	PA11
F	PC14- OSC32_IN	VSS	PF2	PF1	PF0	PG12	PD7	PC12	PA10	PA9	PC6	VDDIO2	VSS
G	PC15- OSC32_OUT	VDD	PF3	PF4	PF5	PG14	PG13	PA8	PC9	PC8	PG6	PC7	VDD
H	PH0-OSC_IN	VSS	NRST	PF10	PC4	PG1	PE10	PB11	PG8	PG7	PD15	VSS	VDD
J	PH1- OSC_OUT	PC0	PC1	PC2	PC5	PG0	PE9	PE15	PG5	PG4	PG3	PG2	PD10
K	PC3	VSSA/VREF-	PA0	PA5	PB0	PF15	PE8	PE14	PH4	PD14	PD12	PD11	PD13
L	VREF+	VDDA	PA4	PA7	PB1	PF14	PE7	PE13	PH5	PD9	PD8	VDD	VSS
M	OPAMP1_VI NM	PA3	VSS	PA6	PF11	PF13	VSS	PE12	PH10	PH11	VSS	PB15	PB14
N	PA2	PA1	VDD	OPAMP2_VI NM	PB2	PF12	VDD	PE11	PB10	PH8	VDD	PB12	PB13

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1. The above figure shows the package top view.

Table 14. Legend/abbreviations used in the pinout table

Name		Abbreviation	Definition
Pin name		Unless otherwise specified in brackets below the pin name, the pin function during and after reset is the same as the actual pin name	
Pin type	S		Supply pin
	I		Input only pin
	I/O		Input / output pin
I/O structure	FT		5 V tolerant I/O
	TT		3.6 V tolerant I/O
	B		Dedicated BOOT0 pin
	RST		Bidirectional reset pin with embedded weak pull-up resistor
	Option for TT or FT I/Os		
		_f ⁽¹⁾	I/O, Fm+ capable
		_u ⁽²⁾	I/O, with USB function supplied by V _{DDUSB}
		_a ^{(3) (4)}	I/O, with Analog switch function supplied by V _{DDA}
		_s ⁽⁵⁾	I/O supplied only by V _{DDIO2}
Notes		Unless otherwise specified by a note, all I/Os are set as analog inputs during and after reset.	
Pin functions	Alternate functions	Functions selected through GPIOx_AFR registers	
	Additional functions	Functions directly selected/enabled through peripheral registers	

1. The related I/O structures are: FT_f, FT_fa, FT_fl, FT_fla.
2. The related I/O structures are: FT_u, FT_lu.
3. The related I/O structures are: FT_a, FT_la, FT_fa, FT_fla, TT_a, TT_la.
4. The analog switch for the TSC function is supplied by VDD.
5. The related I/O structures are: FT_s, FT_fs.



Table 15. STM32L4Q5xx pin definitions

Pin Number								Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions
UFQFPN48	LQFP48	LQFP64	WLCSP100	LQFP100	UFBGA132	LQFP144	UFBGA169						
-	-	-	-	-	D3	-	M11	VSS	S	-	-	-	-
-	-	-	-	-	C4	-	C1	VDD	S	-	-	-	-
-	-	-	-	-	-	-	C3	PI11	I/O	FT	-	OCTOSPIM_P2_IO0, PSSI_D15, EVENTOUT	-
-	-	-	B9	1	B2	1	D3	PE2	I/O	FT_I	-	TRACECK, TIM3_ETR, SAI1_CK1, TSC_G7_IO1, LCD_R7, FMC_A23, SAI1_MCLK_A, EVENTOUT	-
-	-	-	B10	2	A1	2	D2	PE3	I/O	FT_I	-	TRACED0, TIM3_CH1, OCTOSPIM_P1_DQS, TSC_G7_IO2, LCD_R6, FMC_A19, SAI1_SD_B, EVENTOUT	-
-	-	-	C8	3	B1	3	D1	PE4	I/O	FT	-	TRACED1, TIM3_CH2, SAI1_D2, DFSDM1_DATIN3, TSC_G7_IO3, DCMI_D4/PSSI_D4, LCD_B7, FMC_A20, SAI1_FS_A, EVENTOUT	-

Table 15. STM32L4Q5xx pin definitions (continued)

Pin Number								Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions
UFQFPN48	LQFP48	LQFP64	WLCSP100	LQFP100	UFBGA132	LQFP144	UFBGA169						
-	-	-	D8	4	C2	4	E4	PE5	I/O	FT	-	TRACED2, TIM3_CH3, SAI1_CK2, DFSDM1_CKIN3, TSC_G7_IO4, DCMI_D6/PSSI_D6, LCD_G7, FMC_A21, SAI1_SCK_A, EVENTOUT	-
-	-	-	E7	5	D2	5	E3	PE6	I/O	FT	-	TRACED3, TIM3_CH4, SAI1_D1, DCMI_D7/PSSI_D7, LCD_G6, FMC_A22, SAI1_SD_A, EVENTOUT	RTC_TAMP3,WKUP3
1	1	1	C10	6	E2	6	E2	VBAT	S	-	-	-	-
2	2	2	C9	7	C1	7	E1	PC13	I/O	FT	(1) (2)	EVENTOUT	RTC_TAMP1/RTC_TS/R TC_OUT,WKUP2
3	3	3	D10	8	D1	8	F1	PC14- OSC32_IN (PC14)	I/O	FT	(1) (2)	EVENTOUT	OSC32_IN
4	4	4	E10	9	E1	9	G1	PC15- OSC32_OUT (PC15)	I/O	FT	(1) (2)	EVENTOUT	OSC32_OUT
-	-	-	-	-	D6	10	F5	PF0	I/O	FT_f	-	I2C2_SDA, OCTOSPIM_P2_IO0, FMC_A0, EVENTOUT	-



Table 15. STM32L4Q5xx pin definitions (continued)

Pin Number								Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions
UFQFPN48	LQFP48	LQFP64	WLCSP100	LQFP100	UFBGA132	LQFP144	UFBGA169						
-	-	-	-	-	D5	11	F4	PF1	I/O	FT_f	-	I2C2_SCL, OCTOSPIM_P2_IO1, FMC_A1, EVENTOUT	-
-	-	-	-	-	D4	12	F3	PF2	I/O	FT	-	I2C2_SMBA, OCTOSPIM_P2_IO2, FMC_A2, EVENTOUT	-
-	-	-	-	-	E4	13	G3	PF3	I/O	FT	-	OCTOSPIM_P2_IO3, FMC_A3, EVENTOUT	-
-	-	-	-	-	F3	14	G4	PF4	I/O	FT	-	OCTOSPIM_P2_CLK, FMC_A4, EVENTOUT	-
-	-	-	-	-	F4	15	G5	PF5	I/O	FT	-	OCTOSPIM_P2_NCLK, FMC_A5, EVENTOUT	-
-	-	-	D9	10	F2	16	F2	VSS	S	-	-	-	-
-	-	-	E9	11	G2	17	G2	VDD	S	-	-	-	-
-	-	-	-	-	-	18	-	PF6	I/O	FT	-	TIM5_ETR, TIM5_CH1, OCTOSPIM_P1_IO3, SAI1_SD_B, EVENTOUT	-
-	-	-	-	-	-	19	-	PF7	I/O	FT	-	TIM5_CH2, OCTOSPIM_P1_IO2, SAI1_MCLK_B, EVENTOUT	-
-	-	-	-	-	-	20	-	PF8	I/O	FT	-	TIM5_CH3, OCTOSPIM_P1_IO0, SAI1_SCK_B, EVENTOUT	-

Table 15. STM32L4Q5xx pin definitions (continued)

Pin Number								Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions
UFQFPN48	LQFP48	LQFP64	WLCSP100	LQFP100	UFBGA132	LQFP144	UFBGA169						
-	-	-	-	-	-	21	-	PF9	I/O	FT	-	TIM5_CH4, OCTOSPIM_P1_IO1, SAI1_FS_B, TIM15_CH1, EVENTOUT	-
-	-	-	-	-	-	22	H4	PF10	I/O	FT	-	OCTOSPIM_P1_CLK, PSSI_D15, DFSDM1_CKOUT, DCMI_D11/PSSI_D11, SAI1_D3, TIM15_CH2, EVENTOUT	-
5	5	5	F10	12	F1	23	H1	PH0-OSC_IN (PH0)	I/O	FT	-	EVENTOUT	OSC_IN
6	6	6	G10	13	G1	24	J1	PH1- OSC_OUT (PH1)	I/O	FT	-	EVENTOUT	OSC_OUT
7	7	7	E8	14	H2	25	H3	NRST	I-O	RST	-	-	-
-	-	8	F9	15	H1	26	J2	PC0	I/O	FT_fa	-	LPTIM1_IN1, I2C3_SCL, SDMMC2_CKIN, LPUART1_RX, LCD_DE, SDMMC1_CMD, SAI2_FS_A, LPTIM2_IN1, EVENTOUT	ADC12_IN1



Table 15. STM32L4Q5xx pin definitions (continued)

Pin Number								Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions
UFQFPN48	LQFP48	LQFP64	WLCSP100	LQFP100	UFBGA132	LQFP144	UFBGA169						
-	-	9	F8	16	J2	27	J3	PC1	I/O	FT_la_f	-	TRACED0, LPTIM1_OUT, SPI2_MOSI, I2C3_SDA, LPUART1_TX, OCTOSPIM_P1_IO4, SAI1_SD_A, EVENTOUT	ADC12_IN2
-	-	10	H10	17	J3	28	J4	PC2	I/O	FT_la	-	LPTIM1_IN2, SPI2_MISO, DFSDM1_CKOUT, OCTOSPIM_P1_IO5, LCD_HSYNC, EVENTOUT	ADC12_IN3
-	-	11	F7	18	K2	29	K1	PC3	I/O	FT_la	-	LPTIM1_ETR, SAI1_D1, SPI2_MOSI, OCTOSPIM_P1_IO6, SAI1_SD_A, LPTIM2_ETR, EVENTOUT	ADC12_IN4
-	-	-	H9	19	-	30	-	VSSA	S	-	-	-	-
-	-	-	G8	20	-	31	-	VREF-	S	-	-	-	-
8	8	12	-	-	J1	-	K2	VSSA/VREF-	S	-	-	-	-
-	-	-	G7	21	L1	32	L1	VREF+	S	-	-	-	VREFBUF_OUT
-	-	-	J10	22	M1	33	L2	VDDA	S	-	-	-	-
9	9	13	-	-	-	-	-	VDDA/VREF+	S	-	-	-	-

Table 15. STM32L4Q5xx pin definitions (continued)

Pin Number								Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions
UFQFPN48	LQFP48	LQFP64	WLCSP100	LQFP100	UFBGA132	LQFP144	UFBGA169						
10	10	14	G9	23	L2	34	K3	PA0	I/O	FT_a	-	TIM2_CH1, TIM5_CH1, TIM8_ETR, USART2_CTS_NSS, UART4_TX, SAI1_EXTCLK, TIM2_ETR, EVENTOUT	OPAMP1_VINP, ADC12_IN5, RTC_TAMP2,WKUP1
-	-	-	-	-	M3	-	M1	OPAMP1_VI NM	I	TT	-	-	-
11	11	15	H8	24	M2	35	N2	PA1	I/O	FT_I a	-	TIM2_CH2, TIM5_CH2, I2C1_SMBA, SPI1_SCK, USART2_RTS_DE, UART4_RX, OCTOSPIM_P1_DQS, SDMMC2_CMD, TIM15_CH1N, EVENTOUT	OPAMP1_VINM, ADC12_IN6
12	12	16	H7	25	K3	36	N1	PA2	I/O	FT_I a	-	TIM2_CH3, TIM5_CH3, USART2_TX, LPUART1_TX, OCTOSPIM_P1_NCS, SAI2_EXTCLK, TIM15_CH1, EVENTOUT	ADC12_IN7, WKUP4/LSCO
13	13	17	J9	26	L3	37	M2	PA3	I/O	TT_a	-	TIM2_CH4, TIM5_CH4, SAI1_CK1, USART2_RX, LPUART1_RX, OCTOSPIM_P1_CLK, SAI1_MCLK_A, TIM15_CH2, EVENTOUT	OPAMP1_VOUT, ADC12_IN8



Table 15. STM32L4Q5xx pin definitions (continued)

Pin Number								Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions
UFQFPN48	LQFP48	LQFP64	WLCSP100	LQFP100	UFBGA132	LQFP144	UFBGA169						
-	-	18	K10	27	E3	38	H2	VSS	S	-	-	-	-
-	-	19	J8	28	H3	39	N3	VDD	S	-	-	-	-
14	14	20	F6	29	J4	40	L3	PA4	I/O	TT_a	-	OCTOSPIM_P1_NCS, SPI1_NSS, SPI3_NSS, USART2_CK, DCMI_HSYNC/PSSI_DE, LCD_CLK, SAI1_FS_B, LPTIM2_OUT, EVENTOUT	ADC12_IN9, DAC1_OUT1
15	15	21	G6	30	K4	41	K4	PA5	I/O	TT_a	-	TIM2_CH1, TIM2_ETR, TIM8_CH1N, PSSI_D14, SPI1_SCK, LCD_R7, LPTIM2_ETR, EVENTOUT	ADC12_IN10, DAC1_OUT2
16	16	22	K9	31	L4	42	M4	PA6	I/O	FT_a	-	TIM1_BKIN, TIM3_CH1, TIM8_BKIN, DCMI_PIXCLK/PSSI_PDCK, SPI1_MISO, USART3_CTS_NSS, LPUART1_CTS, OCTOSPIM_P1_IO3, TIM16_CH1, EVENTOUT	OPAMP2_VINP, ADC12_IN11
-	-	-	-	-	M4	-	N4	OPAMP2_VI NM	I	TT	-	-	-

Table 15. STM32L4Q5xx pin definitions (continued)

Pin Number								Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions
UFQFPN48	LQFP48	LQFP64	WLCSP100	LQFP100	UFBGA132	LQFP144	UFBGA169						
17	17	23	J7	32	J5	43	L4	PA7	I/O	FT_f la	-	TIM1_CH1N, TIM3_CH2, TIM8_CH1N, I2C3_SCL, SPI1_MOSI, OCTOSPIM_P1_IO2, TIM17_CH1, EVENTOUT	OPAMP2_VINM, ADC12_IN12
-	-	24	H6	33	K5	44	H5	PC4	I/O	FT_a	-	OCTOSPIM_P2_NCS, USART3_TX, OCTOSPIM_P1_IO7, EVENTOUT	COMP1_INM, ADC12_IN13
-	-	25	K8	34	L5	45	J5	PC5	I/O	FT_a	(3)	SAI1_D3, PSSI_D15, USART3_RX, LCD_CLK, EVENTOUT	COMP1_INP, ADC12_IN14, WKUP5
18	18	26	J6	35	M5	46	K5	PB0	I/O	TT_I a	-	TIM1_CH2N, TIM3_CH3, TIM8_CH2N, SPI1_NSS, USART3_CK, OCTOSPIM_P1_IO1, LCD_B6, COMP1_OUT, SAI1_EXTCLK, EVENTOUT	OPAMP2_VOUT, ADC12_IN15
19	19	27	K7	36	M6	47	L5	PB1	I/O	FT_a	-	TIM1_CH3N, TIM3_CH4, TIM8_CH3N, DFSDM1_DATIN0, USART3_RTS_DE, LPUART1_RTS_DE, OCTOSPIM_P1_IO0, LCD_G6, LPTIM2_IN1, EVENTOUT	COMP1_INM, ADC12_IN16



Table 15. STM32L4Q5xx pin definitions (continued)

Pin Number								Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions
UFQFPN48	LQFP48	LQFP64	WLCSP100	LQFP100	UFBGA132	LQFP144	UFBGA169						
20	20	28	F5	37	L6	48	N5	PB2	I/O	FT_a	-	RTC_OUT2, LPTIM1_OUT, I2C3_SMBA, DFSDM1_CKIN0, OCTOSPIM_P1_DQS, EVENTOUT	COMP1_INP
-	-	-	-	-	K6	49	M5	PF11	I/O	FT	-	OCTOSPIM_P1_NCLK, LCD_DE, DCMI_D12/PSSI_D12, EVENTOUT	-
-	-	-	-	-	J7	50	N6	PF12	I/O	FT	-	OCTOSPIM_P2_DQS, LCD_B0, FMC_A6, EVENTOUT	-
-	-	-	-	-	E3	51	H2	VSS	S	-	-	-	-
-	-	-	-	-	H3	52	N7	VDD	S	-	-	-	-
-	-	-	-	-	K7	53	M6	PF13	I/O	FT	-	I2C4_SMBA, LCD_B1, FMC_A7, EVENTOUT	-
-	-	-	-	-	J8	54	L6	PF14	I/O	FT_f	-	I2C4_SCL, TSC_G8_IO1, LCD_G0, FMC_A8, EVENTOUT	-
-	-	-	-	-	J9	55	K6	PF15	I/O	FT_f	-	I2C4_SDA, TSC_G8_IO2, LCD_G1, FMC_A9, EVENTOUT	-
-	-	-	-	-	H9	56	J6	PG0	I/O	FT	-	OCTOSPIM_P2_IO4, TSC_G8_IO3, FMC_A10, EVENTOUT	-

Table 15. STM32L4Q5xx pin definitions (continued)

Pin Number								Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions
UFQFPN48	LQFP48	LQFP64	WLCSP100	LQFP100	UFBGA132	LQFP144	UFBGA169						
-	-	-	-	-	G9	57	H6	PG1	I/O	FT	-	OCTOSPIM_P2_IO5, TSC_G8_IO4, FMC_A11, EVENTOUT	-
-	-	-	K6	38	M7	58	L7	PE7	I/O	FT	-	TIM1_ETR, DFSDM1_DATIN2, LCD_B6, FMC_D4, SAI1_SD_B, EVENTOUT	-
-	-	-	K5	39	L7	59	K7	PE8	I/O	FT	-	TIM1_CH1N, DFSDM1_CKIN2, LCD_B7, FMC_D5, SAI1_SCK_B, EVENTOUT	-
-	-	-	J5	40	M8	60	J7	PE9	I/O	FT	-	TIM1_CH1, DFSDM1_CKOUT, OCTOSPIM_P1_NCLK, LCD_G2, FMC_D6, SAI1_FS_B, EVENTOUT	-
-	-	-	-	-	F6	61	M7	VSS	S	-	-	-	-
-	-	-	-	-	G6	62	N7	VDD	S	-	-	-	-
-	-	-	H5	41	L8	63	H7	PE10	I/O	FT	-	TIM1_CH2N, TSC_G5_IO1, OCTOSPIM_P1_CLK, LCD_G3, FMC_D7, SAI1_MCLK_B, EVENTOUT	-



Table 15. STM32L4Q5xx pin definitions (continued)

Pin Number								Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions
UFQFPN48	LQFP48	LQFP64	WLCSP100	LQFP100	UFBGA132	LQFP144	UFBGA169						
-	-	-	K4	42	M9	64	N8	PE11	I/O	FT	-	TIM1_CH2, TSC_G5_IO2, OCTOSPIM_P1_NCS, LCD_G4, FMC_D8, EVENTOUT	-
-	-	-	G5	43	L9	65	M8	PE12	I/O	FT	-	TIM1_CH3N, SPI1_NSS, TSC_G5_IO3, OCTOSPIM_P1_IO0, LCD_G5, FMC_D9, EVENTOUT	-
-	-	-	G4	44	M10	66	L8	PE13	I/O	FT	-	TIM1_CH3, SPI1_SCK, TSC_G5_IO4, OCTOSPIM_P1_IO1, LCD_G6, FMC_D10, EVENTOUT	-
-	-	-	J4	45	M11	67	K8	PE14	I/O	FT	-	TIM1_CH4, TIM1_BKIN2, SPI1_MISO, OCTOSPIM_P1_IO2, LCD_G7, FMC_D11, EVENTOUT	-
-	-	-	H4	46	M12	68	J8	PE15	I/O	FT	-	TIM1_BKIN, SPI1_MOSI, OCTOSPIM_P1_IO3, LCD_R2, FMC_D12, EVENTOUT	-

Table 15. STM32L4Q5xx pin definitions (continued)

Pin Number								Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions
UFQFPN48	LQFP48	LQFP64	WLCSP100	LQFP100	UFBGA132	LQFP144	UFBGA169						
21	21	29	K3	47	L10	69	N9	PB10	I/O	FT_f I	-	TIM2_CH3, I2C4_SCL, I2C2_SCL, SPI2_SCK, OCTOSPIM_P1_IO3, USART3_TX, LPUART1_RX, TSC_SYNC, OCTOSPIM_P1_CLK, COMP1_OUT, SAI1_SCK_A, EVENTOUT	-
22	22	30	J3	48	L11	70	H8	PB11	I/O	FT_f I	(3)	TIM2_CH4, I2C4_SDA, I2C2_SDA, USART3_RX, LPUART1_TX, OCTOSPIM_P1_NCS, LCD_VSYNC, COMP2_OUT, EVENTOUT	-
-	-	-	-	-	-	-	K9	PH4	I/O	FT_f	-	I2C2_SCL, OCTOSPIM_P2_DQS, PSSI_D14, EVENTOUT	-
-	-	-	-	-	-	-	L9	PH5	I/O	FT_f	-	I2C2_SDA, DCMI_PIXCLK/PSSI_PDCK , EVENTOUT	-
-	-	-	-	-	-	-	N10	PH8	I/O	FT_f	-	I2C3_SDA, OCTOSPIM_P2_IO3, DCMI_HSYNC/PSSI_DE, EVENTOUT	-



Table 15. STM32L4Q5xx pin definitions (continued)

Pin Number								Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions
UFQFPN48	LQFP48	LQFP64	WLCSP100	LQFP100	UFBGA132	LQFP144	UFBGA169						
-	-	-	-	-	-	-	M9	PH10	I/O	FT	-	TIM5_CH1, OCTOSPIM_P2_IO5, DCMI_D1/PSSI_D1, EVENTOUT	-
-	-	-	-	-	-	-	M10	PH11	I/O	FT	(3)	TIM5_CH2, OCTOSPIM_P2_IO6, DCMI_D2/PSSI_D2, EVENTOUT	-
-	-	-	-	-	F6	-	C2	VSS	S	-	-	-	-
-	-	-	-	-	G6	-	L12	VDD	S	-	-	-	-
23	23	31	K2	49	F12	71	A7	VSS	S	-	-	-	-
24	24	32	K1	50	G12	72	N11	VDD	S	-	-	-	-
25	25	33	J1	51	L12	73	N12	PB12	I/O	FT	-	TIM1_BKIN, I2C2_SMBA, SPI2_NSS, DFSDM1_DATIN1, USART3_CK, LPUART1_RTS_DE, TSC_G1_IO1, OCTOSPIM_P1_NCLK, SDMMC2_CK, SAI2_FS_A, TIM15_BKIN, EVENTOUT	-

Table 15. STM32L4Q5xx pin definitions (continued)

Pin Number								Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions
UFQFPN48	LQFP48	LQFP64	WLCSP100	LQFP100	UFBGA132	LQFP144	UFBGA169						
26	26	34	J2	52	K12	74	N13	PB13	I/O	FT_f I	-	TIM1_CH1N, I2C2_SCL, SPI2_SCK, DFSDM1_CKIN1, USART3_CTS_NSS, LPUART1_CTS, TSC_G1_IO2, OCTOSPIM_P1_IO1, SAI2_SCK_A, TIM15_CH1N, EVENTOUT	-
27	27	35	H2	53	K11	75	M13	PB14	I/O	FT_f I	-	TIM1_CH2N, TIM8_CH2N, I2C2_SDA, SPI2_MISO, DFSDM1_DATIN2, USART3_RTS_DE, TSC_G1_IO3, OCTOSPIM_P1_IO6, SDMMC2_D0, SAI2_MCLK_A, TIM15_CH1, EVENTOUT	-
28	28	36	H1	54	K10	76	M12	PB15	I/O	FT	-	RTC_REFIN, TIM1_CH3N, TIM8_CH3N, SPI2_MOSI, DFSDM1_CKIN2, TSC_G1_IO4, OCTOSPIM_P1_IO7, SDMMC2_D1, SAI2_SD_A, TIM15_CH2, EVENTOUT	-



Table 15. STM32L4Q5xx pin definitions (continued)

Pin Number								Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions
UFQFPN48	LQFP48	LQFP64	WLCSP100	LQFP100	UFBGA132	LQFP144	UFBGA169						
-	-	-	H3	55	K9	77	L11	PD8	I/O	FT	-	USART3_TX, DCMI_HSYNC/PSSI_DE, LCD_R3, FMC_D13, EVENTOUT	-
-	-	-	G2	56	K8	78	L10	PD9	I/O	FT	-	USART3_RX, DCMI_PIXCLK/PSSI_PDCK , LCD_R4, FMC_D14, SAI2_MCLK_A, EVENTOUT	-
-	-	-	G1	57	J12	79	J13	PD10	I/O	FT	-	USART3_CK, TSC_G6_IO1, LCD_R5, FMC_D15, SAI2_SCK_A, EVENTOUT	-
-	-	-	-	-	F11	-	L13	VSS	S	-	-	-	-
-	-	-	-	-	H3	-	H13	VDD	S	-	-	-	-
-	-	-	-	58	J11	80	K12	PD11	I/O	FT	-	I2C4_SMBA, USART3_CTS_NSS, TSC_G6_IO2, LCD_R6, FMC_A16, SAI2_SD_A, LPTIM2_ETR, EVENTOUT	-
-	-	-	-	59	J10	81	K11	PD12	I/O	FT_f I	-	TIM4_CH1, I2C4_SCL, USART3_RTS_DE, TSC_G6_IO3, LCD_R7, FMC_A17, SAI2_FS_A, LPTIM2_IN1, EVENTOUT	-

Table 15. STM32L4Q5xx pin definitions (continued)

Pin Number								Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions
UFQFPN48	LQFP48	LQFP64	WLCSP100	LQFP100	UFBGA132	LQFP144	UFBGA169						
-	-	-	-	60	H12	82	K13	PD13	I/O	FT_f I	-	TIM4_CH2, I2C4_SDA, TSC_G6_IO4, FMC_A18, LPTIM2_OUT, EVENTOUT	-
-	-	-	-	-	F11	83	H12	VSS	S	-	-	-	-
-	-	-	F1	-	H3	84	G13	VDD	S	-	-	-	-
-	-	-	G3	61	H11	85	K10	PD14	I/O	FT	-	TIM4_CH3, LCD_B2, FMC_D0, EVENTOUT	-
-	-	-	F4	62	H10	86	H11	PD15	I/O	FT	-	TIM4_CH4, LCD_B3, FMC_D1, EVENTOUT	-
-	-	-	-	-	G10	87	J12	PG2	I/O	FT_s	-	SPI1_SCK, SDMMC2_D4, FMC_A12, SAI2_SCK_B, EVENTOUT	-
-	-	-	-	-	F9	88	J11	PG3	I/O	FT_s	-	SPI1_MISO, SDMMC2_D5, FMC_A13, SAI2_FS_B, EVENTOUT	-
-	-	-	-	-	F10	89	J10	PG4	I/O	FT_s	-	SPI1_MOSI, SDMMC2_D6, FMC_A14, SAI2_MCLK_B, EVENTOUT	-
-	-	-	-	-	E9	90	J9	PG5	I/O	FT_s	-	SPI1_NSS, LPUART1_CTS, SDMMC2_D7, FMC_A15, SAI2_SD_B, EVENTOUT	-



Table 15. STM32L4Q5xx pin definitions (continued)

Pin Number								Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions
UFQFPN48	LQFP48	LQFP64	WLCSP100	LQFP100	UFBGA132	LQFP144	UFBGA169						
-	-	-	-	-	G4	91	G11	PG6	I/O	FT_s	-	OCTOSPIM_P1_DQS, I2C3_SMBA, LPUART1_RTS_DE, LCD_R1, EVENTOUT	-
-	-	-	-	-	H4	92	H10	PG7	I/O	FT_f s	-	SAI1_CK1, I2C3_SCL, OCTOSPIM_P2_DQS, DFSDM1_CKOUT, LPUART1_TX, FMC_INT, SAI1_MCLK_A, EVENTOUT	-
-	-	-	-	-	J6	93	H9	PG8	I/O	FT_f s	-	I2C3_SDA, LPUART1_RX, EVENTOUT	-
-	-	-	-	-	F7	94	F13	VSS	S	-	-	-	-
-	-	-	-	-	G7	95	F12	VDDIO2	S	-	-	-	-
-	-	-	F1	-	H3	-	G13	VDD_15	-	-	-	-	-
-	-	37	F2	63	E12	96	F11	PC6	I/O	FT	-	TIM3_CH1, TIM8_CH1, DFSDM1_CKIN3, SDMMC2_D6, SDMMC1_D0DIR, TSC_G4_IO1, DCMI_D0/PSSI_D0, LCD_G7, SDMMC1_D6, SAI2_MCLK_A, EVENTOUT	-

Table 15. STM32L4Q5xx pin definitions (continued)

Pin Number								Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions
UFQFPN48	LQFP48	LQFP64	WLCSP100	LQFP100	UFBGA132	LQFP144	UFBGA169						
-	-	38	F3	64	E11	97	G12	PC7	I/O	FT	-	TIM3_CH2, TIM8_CH2, DFSDM1_DATIN3, SDMMC2_D7, SDMMC1_D123DIR, TSC_G4_IO2, DCMI_D1/PSSI_D1, LCD_B6, SDMMC1_D7, SAI2_MCLK_B, EVENTOUT	-
-	-	39	E1	65	E10	98	G10	PC8	I/O	FT	-	TIM3_CH3, TIM8_CH3, TSC_G4_IO3, DCMI_D2/PSSI_D2, SDMMC1_D0, EVENTOUT	-
-	-	40	E2	66	D12	99	G9	PC9	I/O	FT_f I	-	TRACED0, TIM8_BKIN2, TIM3_CH4, TIM8_CH4, DCMI_D3/PSSI_D3, I2C3_SDA, TSC_G4_IO4, OTG_FS_NOE, SDMMC1_D1, SAI2_EXTCLK, EVENTOUT	-
29	29	41	E3	67	D11	100	G8	PA8	I/O	FT_f	-	MCO, TIM1_CH1, SAI1_CK2, USART1_CK, OTG_FS_SOF, LCD_B7, SAI1_SCK_A, LPTIM2_OUT, EVENTOUT	-



Table 15. STM32L4Q5xx pin definitions (continued)

Pin Number								Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions
UFQFPN48	LQFP48	LQFP64	WLCSP100	LQFP100	UFBGA132	LQFP144	UFBGA169						
30	30	42	D3	68	D10	101	F10	PA9	I/O	FT _{lu} ^f	-	TIM1_CH2, SPI2_SCK, DCMI_D0/PSSI_D0, USART1_TX, LCD_G7, SAI1_FS_A, TIM15_BKIN, EVENTOUT	OTG_FS_VBUS
31	31	43	D2	69	C12	102	F9	PA10	I/O	FT _{lu} ^f	-	TIM1_CH3, SAI1_D1, DCMI_D1/PSSI_D1, USART1_RX, OTG_FS_ID, LCD_G6, SAI1_SD_A, TIM17_BKIN, EVENTOUT	-
32	32	44	D1	70	B12	103	E13	PA11	I/O	FT _u	-	TIM1_CH4, TIM1_BKIN2, SPI1_MISO, USART1_CTS_NSS, CAN1_RX, OTG_FS_DM, LCD_DE, EVENTOUT	-
33	33	45	C1	71	A12	104	D13	PA12	I/O	FT _u	-	TIM1_ETR, SPI1_MOSI, OCTOSPIM_P2_NCS, USART1_RTS_DE, CAN1_TX, OTG_FS_DP, LCD_VSYNC, EVENTOUT	-
34	34	46	C2	72	A11	105	A11	PA13 (JTMS/SWDIO)	I/O	FT	(4)	JTMS/SWDIO, IR_OUT, OTG_FS_NOE, SAI1_SD_B, EVENTOUT	-
35	35	47	B1	-	F11	-	C12	VSS	S	-	-	-	-
36	36	48	A1	73	C11	106	E12	VDDUSB	S	-	-	-	-

Table 15. STM32L4Q5xx pin definitions (continued)

Pin Number								Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions
UFQFPN48	LQFP48	LQFP64	WLCSP100	LQFP100	UFBGA132	LQFP144	UFBGA169						
-	-	-	B1	74	F11	107	C12	VSS	S	-	-	-	-
-	-	-	F1	75	G11	108	C13	VDD	S	-	-	-	-
-	-	-	-	-	-	-	E11	PH6	I/O	FT	-	I2C2_SMBA, OCTOSPIM_P2_CLK, DCMI_D8/PSSI_D8, EVENTOUT	-
-	-	-	-	-	-	-	D12	PH7	I/O	FT_f	-	I2C3_SCL, OCTOSPIM_P2_NCLK, DCMI_D9/PSSI_D9, EVENTOUT	-
-	-	-	-	-	-	-	D11	PH9	I/O	FT	-	I2C3_SMBA, OCTOSPIM_P2_IO4, DCMI_D0/PSSI_D0, EVENTOUT	-
-	-	-	-	-	-	-	B13	PH12	I/O	FT	-	TIM5_CH3, OCTOSPIM_P2_IO7, DCMI_D3/PSSI_D3, EVENTOUT	-
-	-	-	-	-	-	-	A13	PH14	I/O	FT	-	TIM8_CH2N, DCMI_D4/PSSI_D4, EVENTOUT	-
-	-	-	-	-	-	-	B12	PH15	I/O	FT	-	TIM8_CH3N, OCTOSPIM_P2_IO6, DCMI_D11/PSSI_D11, EVENTOUT	-



Table 15. STM32L4Q5xx pin definitions (continued)

Pin Number								Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions
UFQFPN48	LQFP48	LQFP64	WLCSP100	LQFP100	UFBGA132	LQFP144	UFBGA169						
-	-	-	-	-	-	-	A12	PI0	I/O	FT	-	TIM5_CH4, OCTOSPIM_P1_IO5, SPI2_NSS, DCMI_D13/PSSI_D13, EVENTOUT	-
-	-	-	-	-	-	-	C11	PI8	I/O	FT	-	OCTOSPIM_P2_NCS, DCMI_D12/PSSI_D12, EVENTOUT	-
-	-	-	-	-	-	-	B11	PI1	I/O	FT	-	SPI2_SCK, DCMI_D8/PSSI_D8, EVENTOUT	-
-	-	-	-	-	-	-	B10	PI2	I/O	FT	-	TIM8_CH4, SPI2_MISO, DCMI_D9/PSSI_D9, EVENTOUT	-
-	-	-	-	-	-	-	C10	PI3	I/O	FT	-	TIM8_ETR, SPI2_MOSI, DCMI_D10/PSSI_D10, EVENTOUT	-
-	-	-	-	-	-	-	D10	PI4	I/O	FT	-	TIM8_BKIN, DCMI_D5/PSSI_D5, EVENTOUT	-
-	-	-	-	-	-	-	E10	PI5	I/O	FT	-	TIM8_CH1, OCTOSPIM_P2_NCS, DCMI_VSYNC/PSSI_RDY, EVENTOUT	-

Table 15. STM32L4Q5xx pin definitions (continued)

Pin Number								Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions
UFQFPN48	LQFP48	LQFP64	WLCSP100	LQFP100	UFBGA132	LQFP144	UFBGA169						
-	-	-	-	-	-	-	C9	PH13	I/O	FT	-	TIM8_CH1N, CAN1_TX, EVENTOUT	-
-	-	-	-	-	-	-	B9	PI6	I/O	FT	-	TIM8_CH2, OCTOSPIM_P2_CLK, DCMI_D6/PSSI_D6, EVENTOUT	-
37	37	49	B2	76	A10	109	A10	PA14 (JTCK/SWCLK)	I/O	FT	(4)	JTCK/SWCLK, LPTIM1_OUT, I2C1_SMBA, I2C4_SMBA, OTG_FS_SOF, SAI1_FS_B, EVENTOUT	-
38	38	50	A2	77	A9	110	A9	PA15 (JTDI)	I/O	FT	(4)	JTDI, TIM2_CH1, TIM2_ETR, USART2_RX, SPI1_NSS, SPI3_NSS, USART3_RTS_DE, UART4_RTS_DE, TSC_G3_IO1, LCD_HSYNC, SAI2_FS_B, EVENTOUT	-
-	-	51	D4	78	B11	111	D9	PC10	I/O	FT	-	TRACED1, DCMI_VSYNC/PSSI_RDY, SPI3_SCK, USART3_TX, UART4_TX, TSC_G3_IO2, DCMI_D8/PSSI_D8, SDMMC1_D2, SAI2_SCK_B, EVENTOUT	-



Table 15. STM32L4Q5xx pin definitions (continued)

Pin Number								Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions
UFQFPN48	LQFP48	LQFP64	WLCSP100	LQFP100	UFBGA132	LQFP144	UFBGA169						
-	-	52	C3	79	C10	112	E9	PC11	I/O	FT	-	DCMI_D2/PSSI_D2, OCTOSPIM_P1_NCS, SPI3_MISO, USART3_RX, UART4_RX, TSC_G3_IO3, DCMI_D4/PSSI_D4, SDMMC1_D3, SAI2_MCLK_B, EVENTOUT	-
-	-	53	C4	80	B10	113	F8	PC12	I/O	FT	-	TRACED3, SPI3_MOSI, USART3_CK, UART5_TX, TSC_G3_IO4, DCMI_D9/PSSI_D9, LCD_R6, SDMMC1_CK, SAI2_SD_B, EVENTOUT	-
-	-	-	B3	81	C9	114	B8	PD0	I/O	FT	-	SPI2_NSS, CAN1_RX, LCD_B4, FMC_D2, EVENTOUT	-
-	-	-	A3	82	B9	115	C8	PD1	I/O	FT	-	SPI2_SCK, CAN1_TX, LCD_B5, FMC_D3, EVENTOUT	-
-	-	54	E4	83	C8	116	D8	PD2	I/O	FT	(3)	TRACED2, TIM3_ETR, USART3_RTS_DE, UART5_RX, TSC_SYNC, DCMI_D11/PSSI_D11, SDMMC1_CMD, EVENTOUT	-

Table 15. STM32L4Q5xx pin definitions (continued)

Pin Number								Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions
UFQFPN48	LQFP48	LQFP64	WLCSP100	LQFP100	UFBGA132	LQFP144	UFBGA169						
-	-	-	-	84	B8	117	E8	PD3	I/O	FT	-	SPI2_SCK, DCMI_D5/PSSI_D5, SPI2_MISO, DFSDM1_DATIN0, USART2_CTS_NSS, OCTOSPIM_P2_NCS, LCD_CLK, FMC_CLK, EVENTOUT	-
-	-	-	B4	85	B7	118	C7	PD4	I/O	FT	-	SPI2_MOSI, DFSDM1_CKIN0, USART2_RTS_DE, SDMMC2_CKIN, OCTOSPIM_P1_IO4, FMC_NOE, EVENTOUT	-
-	-	-	E5	86	A6	119	D7	PD5	I/O	FT	-	USART2_TX, OCTOSPIM_P1_IO5, FMC_NWE, EVENTOUT	-
-	-	-	-	-	F11	120	M3	VSS	S	-	-	-	-
-	-	-	A4	-	G11	121	A8	VDD	S	-	-	-	-



Table 15. STM32L4Q5xx pin definitions (continued)

Pin Number								Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions
UFQFPN48	LQFP48	LQFP64	WLCSP100	LQFP100	UFBGA132	LQFP144	UFBGA169						
-	-	-	D5	87	B6	122	E7	PD6	I/O	FT	-	SAI1_D1, DCMI_D10/PSSI_D10, SPI3_MOSI, DFSDM1_DATIN1, USART2_RX, SDMMC2_CK, OCTOSPIM_P1_IO6, LCD_DE, FMC_NWAIT, SAI1_SD_A, EVENTOUT	-
-	-	-	C5	88	A5	123	F7	PD7	I/O	FT	-	DFSDM1_CKIN1, USART2_CK, SDMMC2_CMD, OCTOSPIM_P1_IO7, FMC_NCE/FMC_NE1, EVENTOUT	-
-	-	-	B5	-	D9	124	B7	PG9	I/O	FT_s	-	OCTOSPIM_P2_IO6, SPI3_SCK, USART1_TX, SDMMC2_D0, FMC_NCE/FMC_NE2, SAI2_SCK_A, TIM15_CH1N, EVENTOUT	-
-	-	-	A5	-	D8	125	D6	PG10	I/O	FT_s	-	LPTIM1_IN1, OCTOSPIM_P2_IO7, SPI3_MISO, USART1_RX, SDMMC2_D1, FMC_NE3, SAI2_FS_A, TIM15_CH1, EVENTOUT	-

Table 15. STM32L4Q5xx pin definitions (continued)

Pin Number								Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions
UFQFPN48	LQFP48	LQFP64	WLCSP100	LQFP100	UFBGA132	LQFP144	UFBGA169						
-	-	-	D6	-	G3	126	E6	PG11	I/O	FT_s	-	LPTIM1_IN2, OCTOSPIM_P1_IO5, SPI3_MOSI, USART1_CTS_NSS, SDMMC2_D2, SAI2_MCLK_A, TIM15_CH2, EVENTOUT	-
-	-	-	B6	-	D7	127	F6	PG12	I/O	FT_s	-	LPTIM1_ETR, OCTOSPIM_P2_NCS, SPI3_NSS, USART1_RTS_DE, SDMMC2_D3, FMC_NE4, SAI2_SD_A, EVENTOUT	-
-	-	-	-	-	C7	128	G7	PG13	I/O	FT_f s	-	I2C1_SDA, USART1_CK, LCD_R0, FMC_A24, EVENTOUT	-
-	-	-	-	-	C6	129	G6	PG14	I/O	FT_f s	(3)	I2C1_SCL, LCD_R1, FMC_A25, EVENTOUT	-
-	-	-	-	-	F7	130	A7	VSS	S	-	-	-	-
-	-	-	A6	-	G7	131	B6	VDDIO2	S	-	-	-	-
-	-	-	-	-	K1	132	C6	PG15	I/O	FT_s	(3)	LPTIM1_OUT, I2C1_SMBA, OCTOSPIM_P2_DQS, DCMI_D13/PSSI_D13, EVENTOUT	-



Table 15. STM32L4Q5xx pin definitions (continued)

Pin Number								Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions
UFQFPN48	LQFP48	LQFP64	WLCSP100	LQFP100	UFBGA132	LQFP144	UFBGA169						
39	39	55	C6	89	A8	133	A6	PB3 (JTDO/TRAC ESWO)	I/O	FT_I a	-	JTDO/TRACESWO, TIM2_CH2, OCTOSPIM_P1_IO4, SPI1_SCK, SPI3_SCK, USART1_RTS_DE, OTG_FS_CRS_SYNC, SDMMC2_D2, SAI1_SCK_B, EVENTOUT	COMP2_INM
40	40	56	C7	90	A7	134	A5	PB4 (NJTRST)	I/O	FT_f a	(4)	NJTRST, TIM3_CH1, OCTOSPIM_P1_IO5, I2C3_SDA, SPI1_MISO, SPI3_MISO, USART1_CTS_NSS, UART5_RTS_DE, TSC_G2_IO1, DCMI_D12/PSSI_D12, SDMMC2_D3, SAI1_MCLK_B, TIM17_BKIN, EVENTOUT	COMP2_INP
41	41	57	B7	91	C5	135	B5	PB5	I/O	FT_I a	-	LPTIM1_IN1, TIM3_CH2, OCTOSPIM_P1_IO0, I2C1_SMB, SPI1_MOSI, SPI3_MOSI, USART1_CK, UART5_CTS, TSC_G2_IO2, DCMI_D10/PSSI_D10, COMP2_OUT, SAI1_SD_B, TIM16_BKIN, EVENTOUT	-

Table 15. STM32L4Q5xx pin definitions (continued)

Pin Number								Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions
UFQFPN48	LQFP48	LQFP64	WLCSP100	LQFP100	UFBGA132	LQFP144	UFBGA169						
42	42	58	A7	92	B5	136	C5	PB6	I/O	FT_fa	-	LPTIM1_ETR, TIM4_CH1, TIM8_BKIN2, I2C1_SCL, I2C4_SCL, USART1_TX, TSC_G2_IO3, DCMI_D5/PSSI_D5, LCD_R6, SAI1_FS_B, TIM16_CH1N, EVENTOUT	COMP2_INP
43	43	59	D7	93	B4	137	D5	PB7	I/O	FT_la	-	LPTIM1_IN2, TIM4_CH2, TIM8_BKIN, I2C1_SDA, I2C4_SDA, USART1_RX, UART4_CTS, TSC_G2_IO4, DCMI_VSYNC/PSSI_RDY, LCD_VSYNC, FMC_NL, TIM17_CH1N, EVENTOUT	COMP2_INM, PVD_IN
44	44	60	E6	94	A4	138	E5	PH3-BOOT0	I/O	FT	-	EVENTOUT	-
45	45	61	B8	95	A3	139	C4	PB8	I/O	FT_l	(3)	TIM4_CH3, SAI1_CK1, I2C1_SCL, DFSDM1_CKOUT, SDMMC2_D4, SDMMC1_CKIN, CAN1_RX, DCMI_D6/PSSI_D6, LCD_DE, SDMMC1_D4, SAI1_MCLK_A, TIM16_CH1, EVENTOUT	-



Table 15. STM32L4Q5xx pin definitions (continued)

Pin Number								Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions
UFQFPN48	LQFP48	LQFP64	WLCSP100	LQFP100	UFBGA132	LQFP144	UFBGA169						
46	46	62	A8	96	B3	140	D4	PB9	I/O	FT_f I	-	IR_OUT, TIM4_CH4, SAI1_D2, I2C1_SDA, SPI2_NSS, SDMMC2_D5, SDMMC1_CD1R, CAN1_TX, DCMI_D7/PSSI_D7, SDMMC1_D5, SAI1_FS_A, TIM17_CH1, EVENTOUT	-
-	-	-	-	97	C3	141	A4	PE0	I/O	FT	-	TIM4_ETR, DCMI_D2/PSSI_D2, LCD_HSYNC, FMC_NBL0, TIM16_CH1, EVENTOUT	-
-	-	-	-	98	A2	142	B4	PE1	I/O	FT	(3)	DCMI_D3/PSSI_D3, LCD_VSYNC, FMC_NBL1, TIM17_CH1, EVENTOUT	-
47	47	63	A9	99	D3	143	B3	VSS	S	-	-	-	-
48	48	64	A10	100	C4	144	A3	VDD	S	-	-	-	-
-	-	-	-	-	C4	-	A3	VDD	S	-	-	-	-
-	-	-	-	-	D3	-	B3	VSS	S	-	-	-	-
-	-	-	-	-	-	-	A2	PH2	I/O	FT	-	OCTOSPIM_P1_IO4, EVENTOUT	-
-	-	-	-	-	-	-	B2	PI7	I/O	FT	-	TIM8_CH3, OCTOSPIM_P2_NCLK, DCMI_D7/PSSI_D7, EVENTOUT	-

Table 15. STM32L4Q5xx pin definitions (continued)

Pin Number								Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions
UFQFPN48	LQFP48	LQFP64	WLCSP100	LQFP100	UFBGA132	LQFP144	UFBGA169						
-	-	-	-	-	-	-	B1	PI9	I/O	FT	-	OCTOSPIM_P2_IO2, CAN1_RX, EVENTOUT	-
-	-	-	-	-	-	-	A1	PI10	I/O	FT	-	OCTOSPIM_P2_IO1, PSSI_D14, EVENTOUT	-

- PC13, PC14 and PC15 are supplied through the power switch. Since the switch only sinks a limited amount of current (3 mA), the use of GPIOs PC13 to PC15 in output mode is limited:
 - The speed should not exceed 2 MHz with a maximum load of 30 pF
 - These GPIOs must not be used as current sources (for example to drive a LED).
- After a Backup domain power-up, PC13, PC14 and PC15 operate as GPIOs. Their function then depends on the content of the RTC registers which are not reset by the system reset. For details on how to manage these GPIOs, refer to the Backup domain and RTC register descriptions in the RM0432 reference manual.
- NC (not-connected) balls must be left unconnected. However, PF8 and PF9 NC' IOs are not bonded. They must be configured by software to output push-pull and forced to 0 in the output data register to avoid extra current consumption in low-power modes.
- After reset, these pins are configured as JTAG/SW debug alternate functions, and the internal pull-up on PA15, PA13, PB4 pins and the internal pull-down on PA14 pin are activated.

Table 16. Alternate function AF0 to AF7⁽¹⁾

Port		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7
		OTG_FS/SYS_A F	TIM1/2/5/8/LPT1 M1	TIM1/2/3/4/5	SPI2/SAI1/I2C4/ USART2/CAN2/ OTG_FS/TIM1/8 /OCTOSPIM_P1/ QUADSPI	I2C1/2/3/4/DCMI	SPI1/2/3/I2C4/D FSDM1/DCMI/O CTOSPIM_P1/2/ QUADSPI	SPI3/I2C3/DFSD M1/COMP1/OCT OSPIM_P1/2/QU ADSPI	USART1/2/3/SD MMC2
Port A	PA0	-	TIM2_CH1	TIM5_CH1	TIM8_ETR	-	-	-	USART2_CTS_ NSS
	PA1	-	TIM2_CH2	TIM5_CH2	-	I2C1_SMBA	SPI1_SCK	-	USART2_RTS_ DE
	PA2	-	TIM2_CH3	TIM5_CH3	-	-	-	-	USART2_TX
	PA3	-	TIM2_CH4	TIM5_CH4	SAI1_CK1	-	-	-	USART2_RX
	PA4	-	-	-	OCTOSPIM_P1_ NCS	-	SPI1_NSS	SPI3_NSS	USART2_CK
	PA5	-	TIM2_CH1	TIM2_ETR	TIM8_CH1N	PSSI_D14	SPI1_SCK	-	-
	PA6	-	TIM1_BKIN	TIM3_CH1	TIM8_BKIN	DCMI_PIXCLK/P SSI_PDCK	SPI1_MISO	-	USART3_CTS_ NSS
	PA7	-	TIM1_CH1N	TIM3_CH2	TIM8_CH1N	I2C3_SCL	SPI1_MOSI	-	-
	PA8	MCO	TIM1_CH1	-	SAI1_CK2	-	-	-	USART1_CK
	PA9	-	TIM1_CH2	-	SPI2_SCK	-	DCMI_D0/PSSI_ D0	-	USART1_TX
	PA10	-	TIM1_CH3	-	SAI1_D1	-	DCMI_D1/PSSI_ D1	-	USART1_RX
	PA11	-	TIM1_CH4	TIM1_BKIN2	-	-	SPI1_MISO	-	USART1_CTS_ NSS
	PA12	-	TIM1_ETR	-	-	-	SPI1_MOSI	OCTOSPIM_P2_ NCS	USART1_RTS_ DE
	PA13	JTMS/SWDIO	IR_OUT	-	-	-	-	-	-
	PA14	JTCK/SWCLK	LPTIM1_OUT	-	-	I2C1_SMBA	I2C4_SMBA	-	-
PA15	JTDI	TIM2_CH1	TIM2_ETR	USART2_RX	-	SPI1_NSS	SPI3_NSS	USART3_RTS_ DE	



Table 16. Alternate function AF0 to AF7⁽¹⁾ (continued)

Port	AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7	
	OTG_FS/SYS_AF	TIM1/2/5/8/LPTIM1	TIM1/2/3/4/5	SPI2/SAI1/I2C4/USART2/CAN2/OTG_FS/TIM1/8/OCTOSPIM_P1/QUADSPI	I2C1/2/3/4/DCMI	SPI1/2/3/I2C4/DFSDM1/DCMI/OCTOSPIM_P1/2/QUADSPI	SPI3/I2C3/DFSDM1/COMP1/OCTOSPIM_P1/2/QUADSPI	USART1/2/3/SDMMC2	
Port B	PB0	-	TIM1_CH2N	TIM3_CH3	TIM8_CH2N	-	SPI1_NSS	-	USART3_CK
	PB1	-	TIM1_CH3N	TIM3_CH4	TIM8_CH3N	-	-	DFSDM1_DATIN0	USART3_RTS_DE
	PB2	RTC_OUT2	LPTIM1_OUT	-	-	I2C3_SMBA	-	DFSDM1_CKIN0	-
	PB3	JTDO/TRACESWO	TIM2_CH2	-	OCTOSPIM_P1_IO4	-	SPI1_SCK	SPI3_SCK	USART1_RTS_DE
	PB4	NJTRST	-	TIM3_CH1	OCTOSPIM_P1_IO5	I2C3_SDA	SPI1_MISO	SPI3_MISO	USART1_CTS_NSS
	PB5	-	LPTIM1_IN1	TIM3_CH2	OCTOSPIM_P1_IO0	I2C1_SMBA	SPI1_MOSI	SPI3_MOSI	USART1_CK
	PB6	-	LPTIM1_ETR	TIM4_CH1	TIM8_BKIN2	I2C1_SCL	I2C4_SCL	-	USART1_TX
	PB7	-	LPTIM1_IN2	TIM4_CH2	TIM8_BKIN	I2C1_SDA	I2C4_SDA	-	USART1_RX
	PB8	-	-	TIM4_CH3	SAI1_CK1	I2C1_SCL	DFSDM1_CKOUT	-	SDMMC2_D4
	PB9	-	IR_OUT	TIM4_CH4	SAI1_D2	I2C1_SDA	SPI2_NSS	-	SDMMC2_D5
	PB10	-	TIM2_CH3	-	I2C4_SCL	I2C2_SCL	SPI2_SCK	OCTOSPIM_P1_IO3	USART3_TX
	PB11	-	TIM2_CH4	-	I2C4_SDA	I2C2_SDA	-	-	USART3_RX
	PB12	-	TIM1_BKIN	-	TIM1_BKIN	I2C2_SMBA	SPI2_NSS	DFSDM1_DATIN1	USART3_CK
	PB13	-	TIM1_CH1N	-	-	I2C2_SCL	SPI2_SCK	DFSDM1_CKIN1	USART3_CTS_NSS
	PB14	-	TIM1_CH2N	-	TIM8_CH2N	I2C2_SDA	SPI2_MISO	DFSDM1_DATIN2	USART3_RTS_DE
PB15	RTC_REFIN	TIM1_CH3N	-	TIM8_CH3N	-	SPI2_MOSI	DFSDM1_CKIN2	-	

Table 16. Alternate function AF0 to AF7⁽¹⁾ (continued)

Port	AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7	
	OTG_FS/SYS_AF	TIM1/2/5/8/LPTIM1	TIM1/2/3/4/5	SPI2/SAI1/I2C4/USART2/CAN2/OTG_FS/TIM1/8/OCTOSPIM_P1/QUADSPI	I2C1/2/3/4/DCMI	SPI1/2/3/I2C4/DFSDM1/DCMI/OCTOSPIM_P1/2/QUADSPI	SPI3/I2C3/DFSDM1/COMP1/OCTOSPIM_P1/2/QUADSPI	USART1/2/3/SDMMC2	
Port C	PC0	-	LPTIM1_IN1	-	-	I2C3_SCL	-	-	SDMMC2_CKIN
	PC1	TRACED0	LPTIM1_OUT	-	SPI2_MOSI	I2C3_SDA	-	-	-
	PC2	-	LPTIM1_IN2	-	-	-	SPI2_MISO	DFSDM1_CKOUT	-
	PC3	-	LPTIM1_ETR	-	SAI1_D1	-	SPI2_MOSI	-	-
	PC4	-	-	-	-	-	OCTOSPIM_P2_NCS	-	USART3_TX
	PC5	-	-	-	SAI1_D3	PSSI_D15	-	-	USART3_RX
	PC6	-	-	TIM3_CH1	TIM8_CH1	-	-	DFSDM1_CKIN3	SDMMC2_D6
	PC7	-	-	TIM3_CH2	TIM8_CH2	-	-	DFSDM1_DATIN3	SDMMC2_D7
	PC8	-	-	TIM3_CH3	TIM8_CH3	-	-	-	-
	PC9	TRACED0	TIM8_BKIN2	TIM3_CH4	TIM8_CH4	DCMI_D3/PSSI_D3	-	I2C3_SDA	-
	PC10	TRACED1	-	-	-	DCMI_VSYNC/PSSI_RDY	-	SPI3_SCK	USART3_TX
	PC11	-	-	-	-	DCMI_D2/PSSI_D2	OCTOSPIM_P1_NCS	SPI3_MISO	USART3_RX
	PC12	TRACED3	-	-	-	-	-	SPI3_MOSI	USART3_CK
	PC13	-	-	-	-	-	-	-	-
	PC14	-	-	-	-	-	-	-	-
PC15	-	-	-	-	-	-	-	-	

Table 16. Alternate function AF0 to AF7⁽¹⁾ (continued)

Port	AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7	
	OTG_FS/SYS_A F	TIM1/2/5/8/LPT1 M1	TIM1/2/3/4/5	SPI2/SAI1/I2C4/ USART2/CAN2/ OTG_FS/TIM1/8 /OCTOSPIM_P1/ QUADSPI	I2C1/2/3/4/DCMI	SPI1/2/3/I2C4/D FSDM1/DCMI/O CTOSPIM_P1/2/ QUADSPI	SPI3/I2C3/DFSD M1/COMP1/OCT OSPIM_P1/2/QU ADSPI	USART1/2/3/SD MMC2	
Port D	PD0	-	-	-	-	-	SPI2_NSS	-	
	PD1	-	-	-	-	-	SPI2_SCK	-	
	PD2	TRACED2	-	TIM3_ETR	-	-	-	USART3_RTS_ DE	
	PD3	-	-	-	SPI2_SCK	DCMI_D5/PSSI_ D5	SPI2_MISO	DFSDM1_DATIN 0	USART2_CTS_ NSS
	PD4	-	-	-	-	-	SPI2_MOSI	DFSDM1_CKIN0	USART2_RTS_ DE
	PD5	-	-	-	-	-	-	-	USART2_TX
	PD6	-	-	-	SAI1_D1	DCMI_D10/PSSI_ D10	SPI3_MOSI	DFSDM1_DATIN 1	USART2_RX
	PD7	-	-	-	-	-	-	DFSDM1_CKIN1	USART2_CK
	PD8	-	-	-	-	-	-	-	USART3_TX
	PD9	-	-	-	-	-	-	-	USART3_RX
	PD10	-	-	-	-	-	-	-	USART3_CK
	PD11	-	-	-	-	I2C4_SMBA	-	-	USART3_CTS_ NSS
	PD12	-	-	TIM4_CH1	-	I2C4_SCL	-	-	USART3_RTS_ DE
	PD13	-	-	TIM4_CH2	-	I2C4_SDA	-	-	-
	PD14	-	-	TIM4_CH3	-	-	-	-	-
PD15	-	-	TIM4_CH4	-	-	-	-	-	

Table 16. Alternate function AF0 to AF7⁽¹⁾ (continued)

Port	AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7
	OTG_FS/SYS_AF	TIM1/2/5/8/LPT1M1	TIM1/2/3/4/5	SPI2/SAI1/I2C4/USART2/CAN2/OTG_FS/TIM1/8/OCTOSPIM_P1/QUADSPI	I2C1/2/3/4/DCMI	SPI1/2/3/I2C4/DFSDM1/DCMI/OCTOSPIM_P1/2/QUADSPI	SPI3/I2C3/DFSDM1/COMP1/OCTOSPIM_P1/2/QUADSPI	USART1/2/3/SDMMC2
Port E	PE0	-	-	TIM4_ETR	-	-	-	-
	PE1	-	-	-	-	-	-	-
	PE2	TRACECK	-	TIM3_ETR	SAI1_CK1	-	-	-
	PE3	TRACED0	-	TIM3_CH1	OCTOSPIM_P1_DQS	-	-	-
	PE4	TRACED1	-	TIM3_CH2	SAI1_D2	-	-	DFSDM1_DATIN3
	PE5	TRACED2	-	TIM3_CH3	SAI1_CK2	-	-	DFSDM1_CKIN3
	PE6	TRACED3	-	TIM3_CH4	SAI1_D1	-	-	-
	PE7	-	TIM1_ETR	-	-	-	-	DFSDM1_DATIN2
	PE8	-	TIM1_CH1N	-	-	-	-	DFSDM1_CKIN2
	PE9	-	TIM1_CH1	-	-	-	-	DFSDM1_CKOUT
	PE10	-	TIM1_CH2N	-	-	-	-	-
	PE11	-	TIM1_CH2	-	-	-	-	-
	PE12	-	TIM1_CH3N	-	-	-	SPI1_NSS	-
	PE13	-	TIM1_CH3	-	-	-	SPI1_SCK	-
	PE14	-	TIM1_CH4	TIM1_BKIN2	TIM1_BKIN2	-	SPI1_MISO	-
	PE15	-	TIM1_BKIN	-	TIM1_BKIN	-	SPI1_MOSI	-



Table 16. Alternate function AF0 to AF7⁽¹⁾ (continued)

Port		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7
		OTG_FS/SYS_A F	TIM1/2/5/8/LPT1 M1	TIM1/2/3/4/5	SPI2/SAI1/I2C4/ USART2/CAN2/ OTG_FS/TIM1/8 /OCTOSPIM_P1/ QUADSPI	I2C1/2/3/4/DCMI	SPI1/2/3/I2C4/D FSDM1/DCMI/O CTOSPIM_P1/2/ QUADSPI	SPI3/I2C3/DFSD M1/COMP1/OCT OSPIM_P1/2/QU ADSPI	USART1/2/3/SD MMC2
Port F	PF0	-	-	-	-	I2C2_SDA	OCTOSPIM_P2_ IO0	-	-
	PF1	-	-	-	-	I2C2_SCL	OCTOSPIM_P2_ IO1	-	-
	PF2	-	-	-	-	I2C2_SMBA	OCTOSPIM_P2_ IO2	-	-
	PF3	-	-	-	-	-	OCTOSPIM_P2_ IO3	-	-
	PF4	-	-	-	-	-	OCTOSPIM_P2_ CLK	-	-
	PF5	-	-	-	-	-	OCTOSPIM_P2_ NCLK	-	-
	PF6	-	TIM5_ETR	TIM5_CH1	-	-	-	-	-
	PF7	-	-	TIM5_CH2	-	-	-	-	-
	PF8	-	-	TIM5_CH3	-	-	-	-	-
	PF9	-	-	TIM5_CH4	-	-	-	-	-
	PF10	-	-	-	OCTOSPIM_P1_ CLK	PSSI_D15	-	DFSDM1_CKOU T	-
	PF11	-	-	-	OCTOSPIM_P1_ NCLK	-	-	-	-
	PF12	-	-	-	-	-	OCTOSPIM_P2_ DQS	-	-
	PF13	-	-	-	-	I2C4_SMBA	-	-	-
	PF14	-	-	-	-	I2C4_SCL	-	-	-
PF15	-	-	-	-	I2C4_SDA	-	-	-	

Table 16. Alternate function AF0 to AF7⁽¹⁾ (continued)

Port		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7
		OTG_FS/SYS_AF	TIM1/2/5/8/LPTIM1	TIM1/2/3/4/5	SPI2/SAI1/I2C4/USART2/CAN2/OTG_FS/TIM1/8/OCTOSPIM_P1/QUADSPI	I2C1/2/3/4/DCMI	SPI1/2/3/I2C4/DFSDM1/DCMI/OCTOSPIM_P1/2/QUADSPI	SPI3/I2C3/DFSDM1/COMP1/OCTOSPIM_P1/2/QUADSPI	USART1/2/3/SDMMC2
Port G	PG0	-	-	-	-	-	OCTOSPIM_P2_IO4	-	-
	PG1	-	-	-	-	-	OCTOSPIM_P2_IO5	-	-
	PG2	-	-	-	-	-	SPI1_SCK	-	-
	PG3	-	-	-	-	-	SPI1_MISO	-	-
	PG4	-	-	-	-	-	SPI1_MOSI	-	-
	PG5	-	-	-	-	-	SPI1_NSS	-	-
	PG6	-	-	-	OCTOSPIM_P1_DQS	I2C3_SMBA	-	-	-
	PG7	-	-	-	SAI1_CK1	I2C3_SCL	OCTOSPIM_P2_DQS	DFSDM1_CKOUT	-
	PG8	-	-	-	-	I2C3_SDA	-	-	-
	PG9	-	-	-	-	-	OCTOSPIM_P2_IO6	SPI3_SCK	USART1_TX
	PG10	-	LPTIM1_IN1	-	-	-	OCTOSPIM_P2_IO7	SPI3_MISO	USART1_RX
	PG11	-	LPTIM1_IN2	-	OCTOSPIM_P1_IO5	-	-	SPI3_MOSI	USART1_CTS_NSS
	PG12	-	LPTIM1_ETR	-	-	-	OCTOSPIM_P2_NCS	SPI3_NSS	USART1_RTS_DE
	PG13	-	-	-	-	I2C1_SDA	-	-	USART1_CK
	PG14	-	-	-	-	I2C1_SCL	-	-	-
PG15	-	LPTIM1_OUT	-	-	I2C1_SMBA	OCTOSPIM_P2_DQS	-	-	

Table 16. Alternate function AF0 to AF7⁽¹⁾ (continued)

Port	AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7
	OTG_FS/SYS_A F	TIM1/2/5/8/LPT1 M1	TIM1/2/3/4/5	SPI2/SAI1/I2C4/ USART2/CAN2/ OTG_FS/TIM1/8 /OCTOSPIM_P1/ QUADSPI	I2C1/2/3/4/DCMI	SPI1/2/3/I2C4/D FSDM1/DCMI/O CTOSPIM_P1/2/ QUADSPI	SPI3/I2C3/DFSD M1/COMP1/OCT OSPIM_P1/2/QU ADSPI	USART1/2/3/SD MMC2
Port H	PH0	-	-	-	-	-	-	-
	PH1	-	-	-	-	-	-	-
	PH2	-	-	-	OCTOSPIM_P1_ IO4	-	-	-
	PH3	-	-	-	-	-	-	-
	PH4	-	-	-	-	I2C2_SCL	OCTOSPIM_P2_ DQS	-
	PH5	-	-	-	-	I2C2_SDA	-	-
	PH6	-	-	-	-	I2C2_SMBA	OCTOSPIM_P2_ CLK	-
	PH7	-	-	-	-	I2C3_SCL	OCTOSPIM_P2_ NCLK	-
	PH8	-	-	-	-	I2C3_SDA	OCTOSPIM_P2_ IO3	-
	PH9	-	-	-	-	I2C3_SMBA	OCTOSPIM_P2_ IO4	-
	PH10	-	-	TIM5_CH1	-	-	OCTOSPIM_P2_ IO5	-
	PH11	-	-	TIM5_CH2	-	-	OCTOSPIM_P2_ IO6	-
	PH12	-	-	TIM5_CH3	-	-	OCTOSPIM_P2_ IO7	-
	PH13	-	-	-	TIM8_CH1N	-	-	-
	PH14	-	-	-	TIM8_CH2N	-	-	-
PH15	-	-	-	TIM8_CH3N	-	OCTOSPIM_P2_ IO6	-	

Table 16. Alternate function AF0 to AF7⁽¹⁾ (continued)

Port		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7
		OTG_FS/SYS_A F	TIM1/2/5/8/LPT1 M1	TIM1/2/3/4/5	SPI2/SAI1/I2C4/ USART2/CAN2/ OTG_FS/TIM1/8 /OCTOSPIM_P1/ QUADSPI	I2C1/2/3/4/DCMI	SPI1/2/3/I2C4/D FSDM1/DCMI/O CTOSPIM_P1/2/ QUADSPI	SPI3/I2C3/DFSD M1/COMP1/OCT OSPIM_P1/2/QU ADSPI	USART1/2/3/SD MMC2
Port I	PI0	-	-	TIM5_CH4	OCTOSPIM_P1_ IO5	-	SPI2_NSS	-	-
	PI1	-	-	-	-	-	SPI2_SCK	-	-
	PI2	-	-	-	TIM8_CH4	-	SPI2_MISO	-	-
	PI3	-	-	-	TIM8_ETR	-	SPI2_MOSI	-	-
	PI4	-	-	-	TIM8_BKIN	-	-	-	-
	PI5	-	-	-	TIM8_CH1	-	OCTOSPIM_P2_ NCS	-	-
	PI6	-	-	-	TIM8_CH2	-	OCTOSPIM_P2_ CLK	-	-
	PI7	-	-	-	TIM8_CH3	-	OCTOSPIM_P2_ NCLK	-	-
	PI8	-	-	-	-	-	OCTOSPIM_P2_ NCS	-	-
	PI9	-	-	-	-	-	OCTOSPIM_P2_ IO2	-	-
	PI10	-	-	-	-	-	OCTOSPIM_P2_ IO1	-	-
	PI11	-	-	-	-	-	OCTOSPIM_P2_ IO0	-	-

1. Refer to next table for AF8 to AF15.

Table 17. Alternate function AF8 to AF15⁽¹⁾

Port	AF8	AF9	AF10	AF11	AF12	AF13	AF14	AF15	
	UART4/5/LPUART1/CAN2/SDMMC1/2	CAN1/TSC/LCD	CAN2/OTG_FS/DCMI/OCTOSPIM_P1/2/QUADSPI	SDMMC2/LCD/DSI/LCD	SDMMC1/2/COMP1/2/TIM1/8/FMC/SWPM1	SAI1/2/TIM8	TIM2/8/15/16/17/LPTIM2	EVENTOUT	
Port A	PA0	UART4_TX	-	-	-	-	SAI1_EXTCLK	TIM2_ETR	EVENTOUT
	PA1	UART4_RX	-	OCTOSPIM_P1_DQS	-	SDMMC2_CMD	-	TIM15_CH1N	EVENTOUT
	PA2	LPUART1_TX	-	OCTOSPIM_P1_NCS	-	-	SAI2_EXTCLK	TIM15_CH1	EVENTOUT
	PA3	LPUART1_RX	-	OCTOSPIM_P1_CLK	-	-	SAI1_MCLK_A	TIM15_CH2	EVENTOUT
	PA4	-	-	DCMI_HSYNC/PSSI_DE	LCD_CLK	-	SAI1_FS_B	LPTIM2_OUT	EVENTOUT
	PA5	-	-	-	LCD_R7	-	-	LPTIM2_ETR	EVENTOUT
	PA6	LPUART1_CTS	-	OCTOSPIM_P1_IO3	-	TIM1_BKIN	TIM8_BKIN	TIM16_CH1	EVENTOUT
	PA7	-	-	OCTOSPIM_P1_IO2	-	-	-	TIM17_CH1	EVENTOUT
	PA8	-	-	OTG_FS_SOF	LCD_B7	-	SAI1_SCK_A	LPTIM2_OUT	EVENTOUT
	PA9	-	-	-	LCD_G7	-	SAI1_FS_A	TIM15_BKIN	EVENTOUT
	PA10	-	-	OTG_FS_ID	LCD_G6	-	SAI1_SD_A	TIM17_BKIN	EVENTOUT
	PA11	-	CAN1_RX	OTG_FS_DM	LCD_DE	TIM1_BKIN2	-	-	EVENTOUT
	PA12	-	CAN1_TX	OTG_FS_DP	LCD_VSYNC	-	-	-	EVENTOUT
	PA13	-	-	OTG_FS_NOE	-	-	SAI1_SD_B	-	EVENTOUT
	PA14	-	-	OTG_FS_SOF	-	-	SAI1_FS_B	-	EVENTOUT
PA15	UART4_RTS_DE	TSC_G3_IO1	-	LCD_HSYNC	-	SAI2_FS_B	-	EVENTOUT	

Table 17. Alternate function AF8 to AF15⁽¹⁾ (continued)

Port	AF8	AF9	AF10	AF11	AF12	AF13	AF14	AF15	
	UART4/5/LPUART1/CAN2/SDMMC1/2	CAN1/TSC/LCD	CAN2/OTG_FS/DCMI/OCTOSPIM_P1/2/QUADSPI	SDMMC2/LCD/DSI/LCD	SDMMC1/2/COMP1/2/TIM1/8/FMC/SWP/MI1	SAI1/2/TIM8	TIM2/8/15/16/17/LPTIM2	EVENTOUT	
Port B	PB0	-	-	OCTOSPIM_P1_IO1	LCD_B6	COMP1_OUT	SAI1_EXTCLK	-	EVENTOUT
	PB1	LPUART1_RTS_DE	-	OCTOSPIM_P1_IO0	LCD_G6	-	-	LPTIM2_IN1	EVENTOUT
	PB2	-	-	OCTOSPIM_P1_DQS	-	-	-	-	EVENTOUT
	PB3	-	-	OTG_FS_CRCS_SYNC	-	SDMMC2_D2	SAI1_SCK_B	-	EVENTOUT
	PB4	UART5_RTS_DE	TSC_G2_IO1	DCMI_D12/PSSI_D12	-	SDMMC2_D3	SAI1_MCLK_B	TIM17_BKIN	EVENTOUT
	PB5	UART5_CTS	TSC_G2_IO2	DCMI_D10/PSSI_D10	-	COMP2_OUT	SAI1_SD_B	TIM16_BKIN	EVENTOUT
	PB6	-	TSC_G2_IO3	DCMI_D5/PSSI_D5	LCD_R6	TIM8_BKIN2	SAI1_FS_B	TIM16_CH1N	EVENTOUT
	PB7	UART4_CTS	TSC_G2_IO4	DCMI_VSYNC/PSSI_RDY	LCD_VSYNC	FMC_NL	TIM8_BKIN	TIM17_CH1N	EVENTOUT
	PB8	SDMMC1_CKIN	CAN1_RX	DCMI_D6/PSSI_D6	LCD_DE	SDMMC1_D4	SAI1_MCLK_A	TIM16_CH1	EVENTOUT
	PB9	SDMMC1_CDIR	CAN1_TX	DCMI_D7/PSSI_D7	-	SDMMC1_D5	SAI1_FS_A	TIM17_CH1	EVENTOUT
	PB10	LPUART1_RX	TSC_SYNC	OCTOSPIM_P1_CLK	-	COMP1_OUT	SAI1_SCK_A	-	EVENTOUT
	PB11	LPUART1_TX	-	OCTOSPIM_P1_NCS	LCD_VSYNC	COMP2_OUT	-	-	EVENTOUT
	PB12	LPUART1_RTS_DE	TSC_G1_IO1	OCTOSPIM_P1_NCLK	-	SDMMC2_CK	SAI2_FS_A	TIM15_BKIN	EVENTOUT
	PB13	LPUART1_CTS	TSC_G1_IO2	OCTOSPIM_P1_IO1	-	-	SAI2_SCK_A	TIM15_CH1N	EVENTOUT
	PB14	-	TSC_G1_IO3	OCTOSPIM_P1_IO6	SDMMC2_D0	-	SAI2_MCLK_A	TIM15_CH1	EVENTOUT
PB15	-	TSC_G1_IO4	OCTOSPIM_P1_IO7	SDMMC2_D1	-	SAI2_SD_A	TIM15_CH2	EVENTOUT	



Table 17. Alternate function AF8 to AF15⁽¹⁾ (continued)

Port	AF8	AF9	AF10	AF11	AF12	AF13	AF14	AF15	
	UART4/5/LPUART1/CAN2/SDMMC1/2	CAN1/TSC/LCD	CAN2/OTG_FS/DCMI/OCTOSPIM_P1/2/QUADSPI	SDMMC2/LCD/DSI/LCD	SDMMC1/2/COMP1/2/TIM1/8/FMC/SWP1	SAI1/2/TIM8	TIM2/8/15/16/17/LPTIM2	EVENTOUT	
Port C	PC0	LPUART1_RX	-	-	LCD_DE	SDMMC1_CMD	SAI2_FS_A	LPTIM2_IN1	EVENTOUT
	PC1	LPUART1_TX	-	OCTOSPIM_P1_IO4	-	-	SAI1_SD_A	-	EVENTOUT
	PC2	-	-	OCTOSPIM_P1_IO5	LCD_HSYNC	-	-	-	EVENTOUT
	PC3	-	-	OCTOSPIM_P1_IO6	-	-	SAI1_SD_A	LPTIM2_ETR	EVENTOUT
	PC4	-	-	OCTOSPIM_P1_IO7	-	-	-	-	EVENTOUT
	PC5	-	-	-	LCD_CLK	-	-	-	EVENTOUT
	PC6	SDMMC1_D0DIR	TSC_G4_IO1	DCMI_D0/PSSI_D0	LCD_G7	SDMMC1_D6	SAI2_MCLK_A	-	EVENTOUT
	PC7	SDMMC1_D123DIR	TSC_G4_IO2	DCMI_D1/PSSI_D1	LCD_B6	SDMMC1_D7	SAI2_MCLK_B	-	EVENTOUT
	PC8	-	TSC_G4_IO3	DCMI_D2/PSSI_D2	-	SDMMC1_D0	-	-	EVENTOUT
	PC9	-	TSC_G4_IO4	OTG_FS_NOE	-	SDMMC1_D1	SAI2_EXTCLK	TIM8_BKIN2	EVENTOUT
	PC10	UART4_TX	TSC_G3_IO2	DCMI_D8/PSSI_D8	-	SDMMC1_D2	SAI2_SCK_B	-	EVENTOUT
	PC11	UART4_RX	TSC_G3_IO3	DCMI_D4/PSSI_D4	-	SDMMC1_D3	SAI2_MCLK_B	-	EVENTOUT
	PC12	UART5_TX	TSC_G3_IO4	DCMI_D9/PSSI_D9	LCD_R6	SDMMC1_CK	SAI2_SD_B	-	EVENTOUT
	PC13	-	-	-	-	-	-	-	EVENTOUT
	PC14	-	-	-	-	-	-	-	EVENTOUT
PC15	-	-	-	-	-	-	-	EVENTOUT	

Table 17. Alternate function AF8 to AF15⁽¹⁾ (continued)

Port	AF8	AF9	AF10	AF11	AF12	AF13	AF14	AF15	
	UART4/5/LPUART1/CAN2/SDMMC1/2	CAN1/TSC/LCD	CAN2/OTG_FS/DCMI/OCTOSPIM_P1/2/QUADSPI	SDMMC2/LCD/DSI/LCD	SDMMC1/2/COMP1/2/TIM1/8/FMC/SWP/MI1	SAI1/2/TIM8	TIM2/8/15/16/17/LPTIM2	EVENTOUT	
Port D	PD0	-	CAN1_RX	-	LCD_B4	FMC_D2	-	-	EVENTOUT
	PD1	-	CAN1_TX	-	LCD_B5	FMC_D3	-	-	EVENTOUT
	PD2	UART5_RX	TSC_SYNC	DCMI_D11/PSSI_D11	-	SDMMC1_CMD	-	-	EVENTOUT
	PD3	-	-	OCTOSPIM_P2_NCS	LCD_CLK	FMC_CLK	-	-	EVENTOUT
	PD4	SDMMC2_CKIN	-	OCTOSPIM_P1_IO4	-	FMC_NOE	-	-	EVENTOUT
	PD5	-	-	OCTOSPIM_P1_IO5	-	FMC_NWE	-	-	EVENTOUT
	PD6	SDMMC2_CK	-	OCTOSPIM_P1_IO6	LCD_DE	FMC_NWAIT	SAI1_SD_A	-	EVENTOUT
	PD7	SDMMC2_CMD	-	OCTOSPIM_P1_IO7	-	FMC_NCE/FMC_NE1	-	-	EVENTOUT
	PD8	-	-	DCMI_HSYNC/PSSI_DE	LCD_R3	FMC_D13	-	-	EVENTOUT
	PD9	-	-	DCMI_PIXCLK/PSSI_PDCK	LCD_R4	FMC_D14	SAI2_MCLK_A	-	EVENTOUT
	PD10	-	TSC_G6_IO1	-	LCD_R5	FMC_D15	SAI2_SCK_A	-	EVENTOUT
	PD11	-	TSC_G6_IO2	-	LCD_R6	FMC_A16	SAI2_SD_A	LPTIM2_ETR	EVENTOUT
	PD12	-	TSC_G6_IO3	-	LCD_R7	FMC_A17	SAI2_FS_A	LPTIM2_IN1	EVENTOUT
	PD13	-	TSC_G6_IO4	-	-	FMC_A18	-	LPTIM2_OUT	EVENTOUT
	PD14	-	-	-	LCD_B2	FMC_D0	-	-	EVENTOUT
	PD15	-	-	-	LCD_B3	FMC_D1	-	-	EVENTOUT



Table 17. Alternate function AF8 to AF15⁽¹⁾ (continued)

Port	AF8	AF9	AF10	AF11	AF12	AF13	AF14	AF15	
	UART4/5/LPUART1/CAN2/SDMMC1/2	CAN1/TSC/LCD	CAN2/OTG_FS/DCMI/OCTOSPIM_P1/2/QUADSPI	SDMMC2/LCD/DSI/LCD	SDMMC1/2/COMP1/2/TIM1/8/FMC/SWP/MI1	SAI1/2/TIM8	TIM2/8/15/16/17/LPTIM2	EVENTOUT	
Port E	PE0	-	-	DCMI_D2/PSSI_D2	LCD_HSYNC	FMC_NBL0	-	TIM16_CH1	EVENTOUT
	PE1	-	-	DCMI_D3/PSSI_D3	LCD_VSYNC	FMC_NBL1	-	TIM17_CH1	EVENTOUT
	PE2	-	TSC_G7_IO1	-	LCD_R7	FMC_A23	SAI1_MCLK_A	-	EVENTOUT
	PE3	-	TSC_G7_IO2	-	LCD_R6	FMC_A19	SAI1_SD_B	-	EVENTOUT
	PE4	-	TSC_G7_IO3	DCMI_D4/PSSI_D4	LCD_B7	FMC_A20	SAI1_FS_A	-	EVENTOUT
	PE5	-	TSC_G7_IO4	DCMI_D6/PSSI_D6	LCD_G7	FMC_A21	SAI1_SCK_A	-	EVENTOUT
	PE6	-	-	DCMI_D7/PSSI_D7	LCD_G6	FMC_A22	SAI1_SD_A	-	EVENTOUT
	PE7	-	-	-	LCD_B6	FMC_D4	SAI1_SD_B	-	EVENTOUT
	PE8	-	-	-	LCD_B7	FMC_D5	SAI1_SCK_B	-	EVENTOUT
	PE9	-	-	OCTOSPIM_P1_NCLK	LCD_G2	FMC_D6	SAI1_FS_B	-	EVENTOUT
	PE10	-	TSC_G5_IO1	OCTOSPIM_P1_CLK	LCD_G3	FMC_D7	SAI1_MCLK_B	-	EVENTOUT
	PE11	-	TSC_G5_IO2	OCTOSPIM_P1_NCS	LCD_G4	FMC_D8	-	-	EVENTOUT
	PE12	-	TSC_G5_IO3	OCTOSPIM_P1_IO0	LCD_G5	FMC_D9	-	-	EVENTOUT
	PE13	-	TSC_G5_IO4	OCTOSPIM_P1_IO1	LCD_G6	FMC_D10	-	-	EVENTOUT
	PE14	-	-	OCTOSPIM_P1_IO2	LCD_G7	FMC_D11	-	-	EVENTOUT
	PE15	-	-	OCTOSPIM_P1_IO3	LCD_R2	FMC_D12	-	-	EVENTOUT

Table 17. Alternate function AF8 to AF15⁽¹⁾ (continued)

Port	AF8	AF9	AF10	AF11	AF12	AF13	AF14	AF15	
	UART4/5/LPUART1/CAN2/SDMMC1/2	CAN1/TSC/LCD	CAN2/OTG_FS/DCMI/OCTOSPIM_P1/2/QUADSPI	SDMMC2/LCD/DSI/LCD	SDMMC1/2/COMP1/2/TIM1/8/FMC/SWP/MI1	SAI1/2/TIM8	TIM2/8/15/16/17/LPTIM2	EVENTOUT	
Port F	PF0	-	-	-	-	FMC_A0	-	-	EVENTOUT
	PF1	-	-	-	-	FMC_A1	-	-	EVENTOUT
	PF2	-	-	-	-	FMC_A2	-	-	EVENTOUT
	PF3	-	-	-	-	FMC_A3	-	-	EVENTOUT
	PF4	-	-	-	-	FMC_A4	-	-	EVENTOUT
	PF5	-	-	-	-	FMC_A5	-	-	EVENTOUT
	PF6	-	-	OCTOSPIM_P1_IO3	-	-	SAI1_SD_B	-	EVENTOUT
	PF7	-	-	OCTOSPIM_P1_IO2	-	-	SAI1_MCLK_B	-	EVENTOUT
	PF8	-	-	OCTOSPIM_P1_IO0	-	-	SAI1_SCK_B	-	EVENTOUT
	PF9	-	-	OCTOSPIM_P1_IO1	-	-	SAI1_FS_B	TIM15_CH1	EVENTOUT
	PF10	-	-	DCMI_D11/PSSI_D11	-	-	SAI1_D3	TIM15_CH2	EVENTOUT
	PF11	-	LCD_DE	DCMI_D12/PSSI_D12	-	-	-	-	EVENTOUT
	PF12	-	-	-	LCD_B0	FMC_A6	-	-	EVENTOUT
	PF13	-	-	-	LCD_B1	FMC_A7	-	-	EVENTOUT
	PF14	-	TSC_G8_IO1	-	LCD_G0	FMC_A8	-	-	EVENTOUT
	PF15	-	TSC_G8_IO2	-	LCD_G1	FMC_A9	-	-	EVENTOUT



Table 17. Alternate function AF8 to AF15⁽¹⁾ (continued)

Port	AF8	AF9	AF10	AF11	AF12	AF13	AF14	AF15	
	UART4/5/LPUART1/CAN2/SDMMC1/2	CAN1/TSC/LCD	CAN2/OTG_FS/DCMI/OCTOSPIM_P1/2/QUADSPI	SDMMC2/LCD/DSI/LCD	SDMMC1/2/COMP1/2/TIM1/8/FMC/SWP/MI1	SAI1/2/TIM8	TIM2/8/15/16/17/LPTIM2	EVENTOUT	
Port G	PG0	-	TSC_G8_IO3	-	-	FMC_A10	-	-	EVENTOUT
	PG1	-	TSC_G8_IO4	-	-	FMC_A11	-	-	EVENTOUT
	PG2	-	-	-	SDMMC2_D4	FMC_A12	SAI2_SCK_B	-	EVENTOUT
	PG3	-	-	-	SDMMC2_D5	FMC_A13	SAI2_FS_B	-	EVENTOUT
	PG4	-	-	-	SDMMC2_D6	FMC_A14	SAI2_MCLK_B	-	EVENTOUT
	PG5	LPUART1_CTS	-	-	SDMMC2_D7	FMC_A15	SAI2_SD_B	-	EVENTOUT
	PG6	LPUART1_RTS_DE	LCD_R1	-	-	-	-	-	EVENTOUT
	PG7	LPUART1_TX	-	-	-	FMC_INT	SAI1_MCLK_A	-	EVENTOUT
	PG8	LPUART1_RX	-	-	-	-	-	-	EVENTOUT
	PG9	-	-	-	SDMMC2_D0	FMC_NCE/FMC_NE2	SAI2_SCK_A	TIM15_CH1N	EVENTOUT
	PG10	-	-	-	SDMMC2_D1	FMC_NE3	SAI2_FS_A	TIM15_CH1	EVENTOUT
	PG11	-	-	-	SDMMC2_D2	-	SAI2_MCLK_A	TIM15_CH2	EVENTOUT
	PG12	-	-	-	SDMMC2_D3	FMC_NE4	SAI2_SD_A	-	EVENTOUT
	PG13	-	-	-	LCD_R0	FMC_A24	-	-	EVENTOUT
	PG14	-	-	-	LCD_R1	FMC_A25	-	-	EVENTOUT
	PG15	-	-	DCMI_D13/PSSI_D13	-	-	-	-	EVENTOUT

Table 17. Alternate function AF8 to AF15⁽¹⁾ (continued)

Port	AF8	AF9	AF10	AF11	AF12	AF13	AF14	AF15
	UART4/5/LPUART1/CAN2/SDMMC1/2	CAN1/TSC/LCD	CAN2/OTG_FS/DCMI/OCTOSPIM_P1/2/QUADSPI	SDMMC2/LCD/DSI/LCD	SDMMC1/2/COMP1/2/TIM1/8/FMC/SWPMI1	SAI1/2/TIM8	TIM2/8/15/16/17/LPTIM2	EVENTOUT
Port H	PH0	-	-	-	-	-	-	EVENTOUT
	PH1	-	-	-	-	-	-	EVENTOUT
	PH2	-	-	-	-	-	-	EVENTOUT
	PH3	-	-	-	-	-	-	EVENTOUT
	PH4	-	-	PSSI_D14	-	-	-	EVENTOUT
	PH5	-	-	DCMI_PIXCLK/PSSI_PDCK	-	-	-	EVENTOUT
	PH6	-	-	DCMI_D8/PSSI_D8	-	-	-	EVENTOUT
	PH7	-	-	DCMI_D9/PSSI_D9	-	-	-	EVENTOUT
	PH8	-	-	DCMI_HSYNC/PSSI_DE	-	-	-	EVENTOUT
	PH9	-	-	DCMI_D0/PSSI_D0	-	-	-	EVENTOUT
	PH10	-	-	DCMI_D1/PSSI_D1	-	-	-	EVENTOUT
	PH11	-	-	DCMI_D2/PSSI_D2	-	-	-	EVENTOUT
	PH12	-	-	DCMI_D3/PSSI_D3	-	-	-	EVENTOUT
	PH13	-	CAN1_TX	-	-	-	-	EVENTOUT
	PH14	-	-	DCMI_D4/PSSI_D4	-	-	-	EVENTOUT
	PH15	-	-	DCMI_D11/PSSI_D11	-	-	-	EVENTOUT

Table 17. Alternate function AF8 to AF15⁽¹⁾ (continued)

Port	AF8	AF9	AF10	AF11	AF12	AF13	AF14	AF15	
	UART4/5/LPUART1/CAN2/SDMMC1/2	CAN1/TSC/LCD	CAN2/OTG_FS/DCMI/OCTOSPIM_P1/2/QUADSPI	SDMMC2/LCD/DSI/LCD	SDMMC1/2/COMP1/2/TIM1/8/FMC/SWPMI1	SAI1/2/TIM8	TIM2/8/15/16/17/LPTIM2	EVENTOUT	
Port I	PI0	-	-	DCMI_D13/PSSI_D13	-	-	-	-	EVENTOUT
	PI1	-	-	DCMI_D8/PSSI_D8	-	-	-	-	EVENTOUT
	PI2	-	-	DCMI_D9/PSSI_D9	-	-	-	-	EVENTOUT
	PI3	-	-	DCMI_D10/PSSI_D10	-	-	-	-	EVENTOUT
	PI4	-	-	DCMI_D5/PSSI_D5	-	-	-	-	EVENTOUT
	PI5	-	-	DCMI_VSYNC/PSSI_RDY	-	-	-	-	EVENTOUT
	PI6	-	-	DCMI_D6/PSSI_D6	-	-	-	-	EVENTOUT
	PI7	-	-	DCMI_D7/PSSI_D7	-	-	-	-	EVENTOUT
	PI8	-	-	DCMI_D12/PSSI_D12	-	-	-	-	EVENTOUT
	PI9	-	CAN1_RX	-	-	-	-	-	EVENTOUT
	PI10	-	-	PSSI_D14	-	-	-	-	EVENTOUT
	PI11	-	-	PSSI_D15	-	-	-	-	EVENTOUT

1. Refer to previous table for AF0 to AF7.

5 Memory mapping

For memory map and peripheral register boundary addresses refer to the corresponding section of reference manual RM0432.

6 Electrical characteristics

6.1 Parameter conditions

Unless otherwise specified, all voltages are referenced to V_{SS} .

6.1.1 Minimum and maximum values

Unless otherwise specified, the minimum and maximum values are guaranteed in the worst conditions of ambient temperature, supply voltage and frequencies by tests in production on 100% of the devices with an ambient temperature at $T_A = 25\text{ }^\circ\text{C}$ and $T_A = T_{A\text{max}}$ (given by the selected temperature range).

Data based on characterization results, design simulation and/or technology characteristics are indicated in the table footnotes and are not tested in production. Based on characterization, the minimum and maximum values refer to sample tests and represent the mean value plus or minus three times the standard deviation (mean $\pm 3\sigma$).

6.1.2 Typical values

Unless otherwise specified, typical data are based on $T_A = 25\text{ }^\circ\text{C}$, $V_{DD} = V_{DDA} = 3\text{ V}$. They are given only as design guidelines and are not tested.

Typical ADC accuracy values are determined by characterization of a batch of samples from a standard diffusion lot over the full temperature range, where 95% of the devices have an error less than or equal to the value indicated (mean $\pm 2\sigma$).

6.1.3 Typical curves

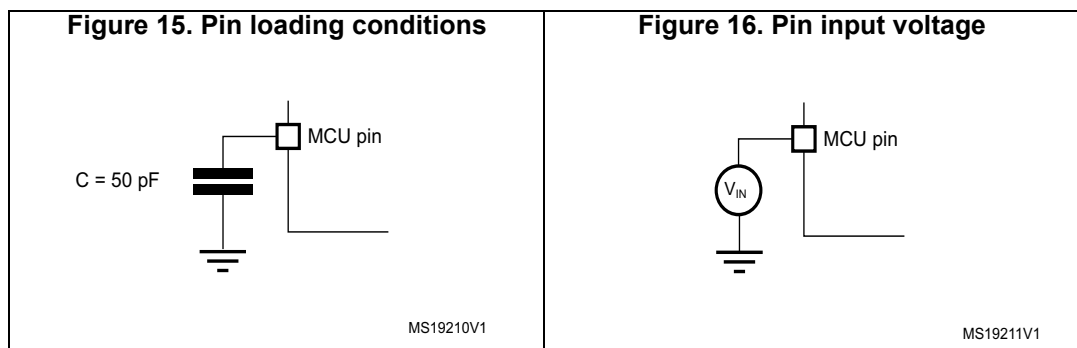
Unless otherwise specified, all typical curves are given only as design guidelines and are not tested.

6.1.4 Loading capacitor

The loading conditions used for pin parameter measurement are shown in [Figure 15](#).

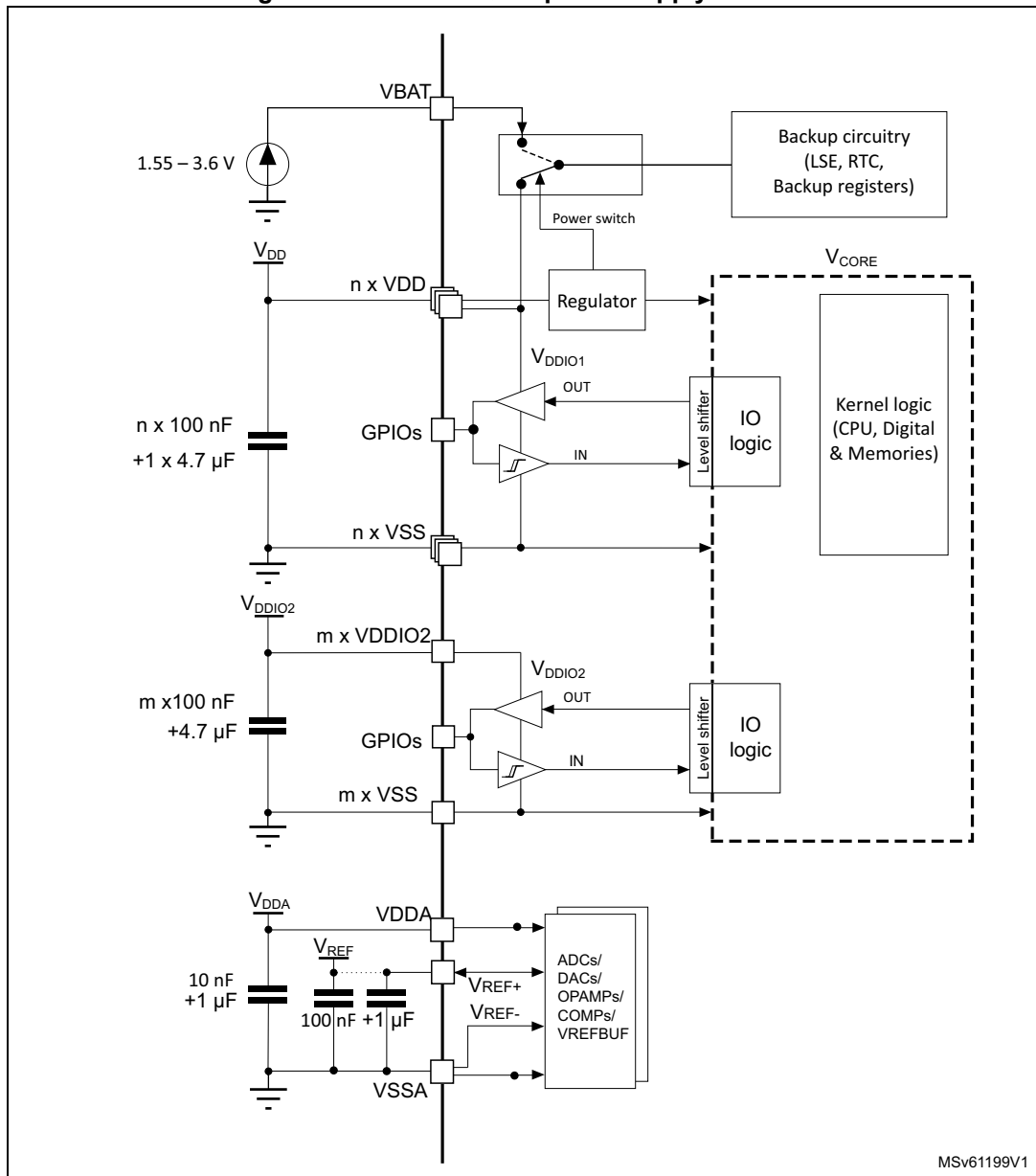
6.1.5 Pin input voltage

The input voltage measurement on a pin of the device is described in [Figure 16](#).



6.1.6 Power supply scheme

Figure 17. STM32L4Q5xx power supply scheme

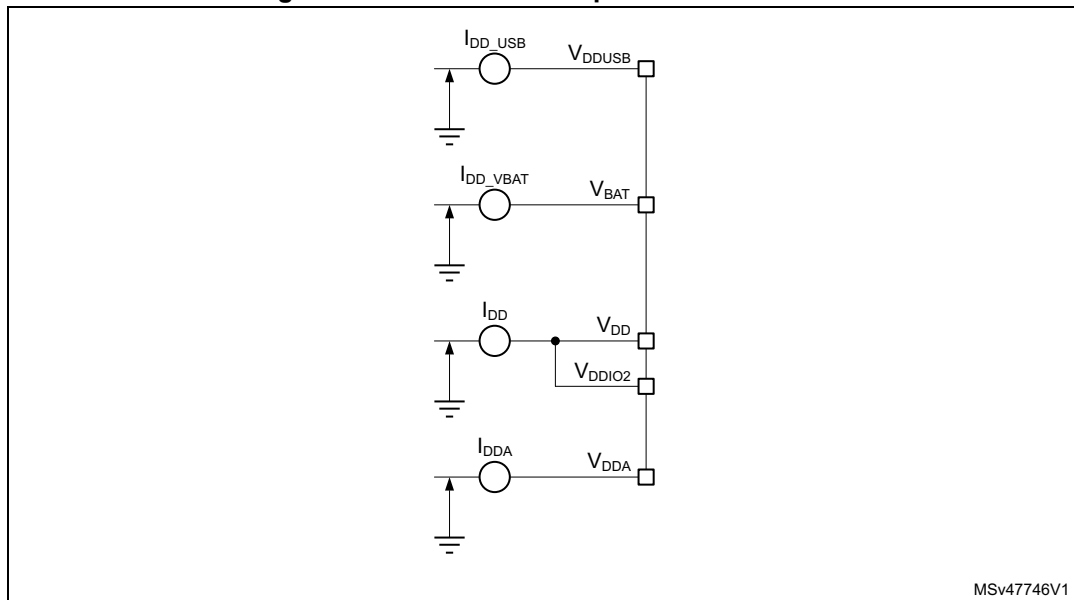


Caution: Each power supply pair (V_{DD}/V_{SS} , V_{DDA}/V_{SSA} etc.) must be decoupled with filtering ceramic capacitors as shown above. These capacitors must be placed as close as possible to, or below, the appropriate pins on the underside of the PCB to ensure the good functionality of the device.

6.1.7 Current consumption measurement

The I_{DD_ALL} parameters given in [Table 25](#) to [Table 32](#) represent the total MCU consumption including the current supplying V_{DD} , V_{DDIO2} , V_{DDA} , V_{DDUSB} and V_{BAT} .

Figure 18. Current consumption measurement



6.2 Absolute maximum ratings

Stresses above the absolute maximum ratings listed in [Table 18: Voltage characteristics](#), [Table 19: Current characteristics](#) and [Table 20: Thermal characteristics](#) may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these conditions is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability. Device mission profile (application conditions) is compliant with JEDEC JESD47 qualification standard, extended mission profiles are available on demand.

Table 18. Voltage characteristics⁽¹⁾

Symbol	Ratings	Min	Max	Unit
$V_{DDX} - V_{SS}$	External main supply voltage (including V_{DD} , V_{DDA} , V_{DDIO2} , V_{DDUSB} , V_{BAT} , V_{REF+})	-0.3	4.0	V
$V_{IN}^{(2)}$	Input voltage on FT_xxx pins	$V_{SS}-0.3$	$\min(V_{DD}, V_{DDA}, V_{DDIO2}, V_{DDUSB}) + 4.0^{(3)(4)}$	V
	Input voltage on TT_xx pins	$V_{SS}-0.3$	4.0	
	Input voltage on any other pins	$V_{SS}-0.3$	4.0	

Table 18. Voltage characteristics⁽¹⁾ (continued)

Symbol	Ratings	Min	Max	Unit
$ \Delta V_{DDx} $	Variations between different V_{DDx} power pins of the same domain	-	50	mV
$ V_{SSx}-V_{SS} $	Variations between all the different ground pins ⁽⁵⁾	-	50	
$V_{REF+} - V_{DDA}$	Allowed voltage difference for $V_{REF+} > V_{DDA}$	-	0.4	V

1. All main power (V_{DD} , V_{DDA} , V_{DDIO2} , V_{DDUSB} , V_{BAT}) and ground (V_{SS} , V_{SSA}) pins must always be connected to the external power supply, in the permitted range.
2. V_{IN} maximum must always be respected. Refer to [Table 19: Current characteristics](#) for the maximum allowed injected current values.
3. This formula has to be applied only on the power supplies related to the IO structure described in the pin definition table.
4. To sustain a voltage higher than 4 V the internal pull-up/pull-down resistors must be disabled.
5. Include VREF- pin.

Table 19. Current characteristics

Symbol	Ratings	Max	Unit
$\Sigma I_{V_{DD}}$	Total current into sum of all V_{DD} power lines (source) ⁽¹⁾	200	mA
$\Sigma I_{V_{SS}}$	Total current out of sum of all V_{SS} ground lines (sink) ⁽¹⁾	200	
$I_{V_{DD}(PIN)}$	Maximum current into each V_{DD} power pin (source) ⁽¹⁾	100	
$I_{V_{SS}(PIN)}$	Maximum current out of each V_{SS} ground pin (sink) ⁽¹⁾	100	
$I_{IO(PIN)}$	Output current sunk by any I/O and control pin except FT_f	20	
	Output current sunk by any FT_f pin	20	
	Output current sourced by any I/O and control pin	20	
$\Sigma I_{IO(PIN)}$	Total output current sunk by sum of all I/Os and control pins ⁽²⁾	100	
	Total output current sourced by sum of all I/Os and control pins ⁽²⁾	100	
$I_{INJ(PIN)}^{(3)}$	Injected current on FT_xxx, TT_xx, RST and B pins, except PA4, PA5	-5/+0 ⁽⁴⁾	
	Injected current on PA4, PA5	-5/0	
$\Sigma I_{INJ(PIN)} $	Total injected current (sum of all I/Os and control pins) ⁽⁵⁾	25	

1. All main power (V_{DD} , V_{DDA} , V_{DDIO2} , V_{DDUSB} , V_{BAT}) and ground (V_{SS} , V_{SSA}) pins must always be connected to the external power supplies, in the permitted range.
2. This current consumption must be correctly distributed over all I/Os and control pins. The total output current must not be sunk/sourced between two consecutive power supply pins referring to high pin count QFP packages.
3. Positive injection (when $V_{IN} > V_{DDIOx}$) is not possible on these I/Os and does not occur for input voltages lower than the specified maximum value.
4. A negative injection is induced by $V_{IN} < V_{SS}$. $I_{INJ(PIN)}$ must never be exceeded. Refer also to [Table 18: Voltage characteristics](#) for the minimum allowed input voltage values.
5. When several inputs are submitted to a current injection, the maximum $\Sigma |I_{INJ(PIN)}|$ is the absolute sum of the negative injected currents (instantaneous values).

Table 20. Thermal characteristics

Symbol	Ratings	Value	Unit
T _{STG}	Storage temperature range	-65 to +150	°C
T _J	Maximum junction temperature	150	°C

6.3 Operating conditions

6.3.1 General operating conditions

Table 21. General operating conditions

Symbol	Parameter	Conditions	Min	Max	Unit	
f _{HCLK}	Internal AHB clock frequency	-	0	120	MHz	
f _{PCLK1}	Internal APB1 clock frequency	-	0	120		
f _{PCLK2}	Internal APB2 clock frequency	-	0	120		
V _{DD}	Standard operating voltage	-	1.71 ⁽¹⁾	3.6	V	
V _{DDIO2}	PG[15:2] I/Os supply voltage	At least one I/O in PG[15:2] used	1.08	3.6	V	
		PG[15:2] not used	0	3.6		
V _{DDA}	Analog supply voltage	ADC or COMP used	1.62	3.6		
		DAC or OPAMP used	1.8			
		VREFBUF used	2.4			
		ADC, DAC, OPAMP, COMP, VREFBUF not used	0			
V _{BAT}	Backup operating voltage	-	1.55	3.6		
V _{DDUSB}	USB supply voltage	USB used	3.0	3.6		V
		USB not used	0	3.6		
V _{IN}	I/O input voltage	TT_xx I/O	-0.3	V _{DDIOx} +0.3	V	
		All I/O except TT_xx	-0.3	MIN(MIN(V _{DD} , V _{DDA} , V _{DDIO2} , V _{DDUSB})+3.6 V, 5.5 V) ⁽²⁾⁽³⁾		
P _D	Power dissipation at T _A = 85 °C for suffix 6 ⁽⁴⁾	LQFP144	See Section 7.9: Thermal characteristics for application appropriate thermal resistance and package. Power dissipation is then calculated according ambient temperature (T _A) and maximum junction temperature (T _J) and selected thermal resistance.		mW	
		LQFP100				
		UFBGA169				
		UFBGA132				
		WLCSP100				
		LQFP48				
		LQFP64				
		UFQFPN48				

Table 21. General operating conditions (continued)

Symbol	Parameter	Conditions	Min	Max	Unit
P _D	Power dissipation at T _A = 125 °C for suffix 3 ⁽⁴⁾	LQFP144	See Section 7.9: Thermal characteristics for application appropriate thermal resistance and package. Power dissipation is then calculated according ambient temperature (T _A) and maximum junction temperature (T _J) and selected thermal resistance.		mW
		LQFP100			
		UFBGA169			
		UFBGA132			
		WLCSP100			
		LQFP48			
		LGFP64			
		UFQFPN48			
T _A	Ambient temperature for the suffix 6 version	Maximum power dissipation	-40	85	°C
		Low-power dissipation ⁽⁵⁾	-40	105	
	Ambient temperature for the suffix 3 version	Maximum power dissipation	-40	125	
		Low-power dissipation ⁽⁵⁾	-40	130	
T _J	Junction temperature range	Suffix 6 version	-40	105	°C
		Suffix 3 version	-40	130	

1. When RESET is released functionality is guaranteed down to V_{BOR0} Min.
2. This formula has to be applied only on the power supplies related to the IO structure described by the pin definition table. Maximum I/O input voltage is the smallest value between MIN(V_{DD}, V_{DDA}, V_{DDIO2}, V_{DDUSB})+3.6 V and 5.5V.
3. For operation with voltage higher than Min (V_{DD}, V_{DDA}, V_{DDIO2}, V_{DDUSB}) +0.3 V, the internal Pull-up and Pull-Down resistors must be disabled.
4. If T_A is lower, higher P_D values are allowed as long as T_J does not exceed T_{Jmax} (see [Section 7.9: Thermal characteristics](#)).
5. In low-power dissipation state, T_A can be extended to this range as long as T_J does not exceed T_{Jmax} (see [Section 7.9: Thermal characteristics](#)).

6.3.2 Operating conditions at power-up / power-down

The parameters given in [Table 22](#) are derived from tests performed under the ambient temperature condition summarized in [Table 21](#).

Table 22. Operating conditions at power-up / power-down

Symbol	Parameter	Conditions	Min	Max	Unit
t_{VDD}	V_{DD} rise time rate	-	0	∞	$\mu\text{s/V}$
	V_{DD} fall time rate	ENULP=0	10	∞	
		ENULP=1	100	∞	ms/V
t_{VDDA}	V_{DDA} rise time rate	-	0	∞	$\mu\text{s/V}$
	V_{DDA} fall time rate		10	∞	
t_{VDDUSB}	V_{DDUSB} rise time rate	-	0	∞	$\mu\text{s/V}$
	V_{DDUSB} fall time rate		10	∞	
t_{VDDIO2}	V_{DDIO2} rise time rate	-	0	∞	$\mu\text{s/V}$
	V_{DDIO2} fall time rate		10	∞	

6.3.3 Embedded reset and power control block characteristics

The parameters given in [Table 23](#) are derived from tests performed under the ambient temperature conditions summarized in [Table 21: General operating conditions](#).

Table 23. Embedded reset and power control block characteristics

Symbol	Parameter	Conditions ⁽¹⁾	Min	Typ	Max	Unit
$t_{RSTTEMPO}$ ⁽²⁾	Reset temporization after BOR0 is detected	V_{DD} rising	-	250	400	μs
V_{BOR0} ⁽²⁾	Brown-out reset threshold 0	Rising edge	1.62	1.66	1.7	V
		Falling edge	1.6	1.64	1.69	
V_{BOR1}	Brown-out reset threshold 1	Rising edge	2.06	2.1	2.14	V
		Falling edge	1.96	2	2.04	
V_{BOR2}	Brown-out reset threshold 2	Rising edge	2.26	2.31	2.35	V
		Falling edge	2.16	2.20	2.24	
V_{BOR3}	Brown-out reset threshold 3	Rising edge	2.56	2.61	2.66	V
		Falling edge	2.47	2.52	2.57	
V_{BOR4}	Brown-out reset threshold 4	Rising edge	2.85	2.90	2.95	V
		Falling edge	2.76	2.81	2.86	
V_{PVD0}	Programmable voltage detector threshold 0	Rising edge	2.1	2.15	2.19	V
		Falling edge	2	2.05	2.1	
V_{PVD1}	PVD threshold 1	Rising edge	2.26	2.31	2.36	V
		Falling edge	2.15	2.20	2.25	

Table 23. Embedded reset and power control block characteristics (continued)

Symbol	Parameter	Conditions ⁽¹⁾	Min	Typ	Max	Unit
V _{PVD2}	PVD threshold 2	Rising edge	2.41	2.46	2.51	V
		Falling edge	2.31	2.36	2.41	
V _{PVD3}	PVD threshold 3	Rising edge	2.56	2.61	2.66	V
		Falling edge	2.47	2.52	2.57	
V _{PVD4}	PVD threshold 4	Rising edge	2.69	2.74	2.79	V
		Falling edge	2.59	2.64	2.69	
V _{PVD5}	PVD threshold 5	Rising edge	2.85	2.91	2.96	V
		Falling edge	2.75	2.81	2.86	
V _{PVD6}	PVD threshold 6	Rising edge	2.92	2.98	3.04	V
		Falling edge	2.84	2.90	2.96	
V _{hyst_BORH0}	Hysteresis voltage of BORH0	Hysteresis in continuous mode	-	20	-	mV
		Hysteresis in other mode	-	30	-	
V _{hyst_BOR_PVD}	Hysteresis voltage of BORH (except BORH0) and PVD	-	-	100	-	mV
I _{DD} (BOR_PVD) ⁽²⁾	BOR ⁽³⁾ (except BOR0) and PVD consumption from V _{DD} with ENULP=0	-	-	1.1	1.6	μA
	BOR ⁽³⁾ (except BOR0) and PVD consumption from V _{DD} with ENULP=1	-	-	55	1000	nA
V _{PVM1}	V _{DDUSB} peripheral voltage monitoring	-	1.18	1.22	1.26	V
V _{PVM3}	V _{DDA} peripheral voltage monitoring	Rising edge	1.61	1.65	1.69	V
		Falling edge	1.6	1.64	1.68	
V _{PVM4}	V _{DDA} peripheral voltage monitoring	Rising edge	1.78	1.82	1.86	V
		Falling edge	1.77	1.81	1.85	
V _{hyst_PVM3}	PVM3 hysteresis	-	-	10	-	mV
V _{hyst_PVM4}	PVM4 hysteresis	-	-	10	-	mV
I _{DD} (PVM1/PVM2) ⁽²⁾	PVM1 and PVM2 consumption from V _{DD}	-	-	0.2	-	μA
I _{DD} (PVM3/PVM4) ⁽²⁾	PVM3 and PVM4 consumption from V _{DD}	-	-	2	-	μA

1. Continuous mode means Run/Sleep modes, or temperature sensor enable in Low-power run/Low-power sleep modes.
2. Guaranteed by design.
3. BOR0 is enabled in all modes (except shutdown) and its consumption is therefore included in the supply current characteristics tables.

6.3.4 Embedded voltage reference

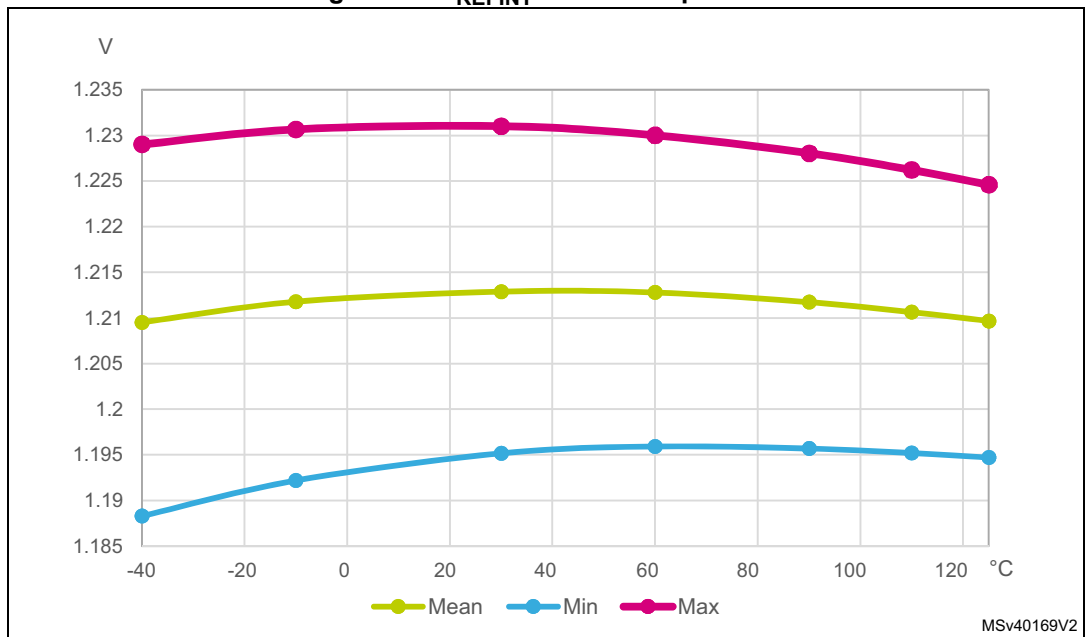
The parameters given in [Table 24](#) are derived from tests performed under the ambient temperature and supply voltage conditions summarized in [Table 21: General operating conditions](#).

Table 24. Embedded internal voltage reference

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
V_{REFINT}	Internal reference voltage	$-40\text{ °C} < T_J < +130\text{ °C}$	1.182	1.212	1.232	V
$t_{S_vrefint}^{(1)}$	ADC sampling time when reading the internal reference voltage	-	4 ⁽²⁾	-	-	μs
$t_{start_vrefint}$	Start time of reference voltage buffer when ADC is enable	-	-	8	12 ⁽²⁾	μs
$I_{DD}(V_{REFINTBUF})$	V_{REFINT} buffer consumption from V_{DD} when converted by ADC	-	-	12.5	20 ⁽²⁾	μA
ΔV_{REFINT}	Internal reference voltage spread over the temperature range	$V_{DD} = 3\text{ V}$	-	5	7.5 ⁽²⁾	mV
T_{Coeff}	Average temperature coefficient	$-40\text{ °C} < T_J < +130\text{ °C}$	-	30	50 ⁽²⁾	ppm/°C
A_{Coeff}	Long term stability	1000 hours, $T = 25\text{ °C}$	-	300	1000 ⁽²⁾	ppm
$V_{DDCoeff}$	Average voltage coefficient	$3.0\text{ V} < V_{DD} < 3.6\text{ V}$	-	250	1200 ⁽²⁾	ppm/V
V_{REFINT_DIV1}	1/4 reference voltage	-	24	25	26	% V_{REFINT}
V_{REFINT_DIV2}	1/2 reference voltage		49	50	51	
V_{REFINT_DIV3}	3/4 reference voltage		74	75	76	

1. The shortest sampling time can be determined in the application by multiple iterations.
2. Guaranteed by design.

Figure 19. V_{REFINT} versus temperature



6.3.5 Supply current characteristics

The current consumption is a function of several parameters and factors such as the operating voltage, ambient temperature, I/O pin loading, device software configuration, operating frequencies, I/O pin switching rate, program location in memory and executed binary code

The current consumption is measured as described in .

Typical and maximum current consumption

The MCU is placed under the following conditions:

- All I/O pins are in analog input mode
- All peripherals are disabled except when explicitly mentioned
- The Flash memory access time is adjusted with the minimum wait states number, depending on the f_{HCLK} frequency (refer to the table “Number of wait states according to CPU clock (HCLK) frequency” available in the RM0432 reference manual).
- When the peripherals are enabled $f_{PCLK} = f_{HCLK}$
- The voltage scaling Range 1 is adjusted to f_{HCLK} frequency as follows:
 - Voltage Range 1 Boost mode for $80 \text{ MHz} < f_{HCLK} \leq 120 \text{ MHz}$
 - Voltage Range 1 Normal mode for $26 \text{ MHz} < f_{HCLK} \leq 80 \text{ MHz}$

The parameters given in [Table 25](#) to [Table 32](#) are derived from tests performed under ambient temperature and supply voltage conditions summarized in [Table 21: General operating conditions](#).



Table 25. Current consumption in Run and Low-power run modes, code with data processing running from Flash in single Bank, ART enable (Cache ON Prefetch OFF)

Symbol	Parameter	Conditions		fHCLK	TYP					MAX ⁽¹⁾					Unit
		-	Voltage scaling		25°C	55°C	85°C	105°C	125°C	25°C	55°C	85°C	105°C	125°C	
IDD(Run)	Supply current in Run mode	fHCLK = fHSE up to 48 MHz included, bypass mode PLL ON above 48 MHz all peripherals disable	Range 2	26 MHz	3.15	3.50	4.50	5.95	8.50	3.90	4.8	6.8	11	17	mA
				16 MHz	2.05	2.40	3.35	4.80	7.35	2.60	3.4	5.4	8.7	15	
				8 MHz	1.15	1.50	2.50	3.85	6.40	1.60	2.3	4.3	7.6	14	
				4 MHz	0.69	1.00	2.00	3.40	5.95	1.00	1.8	3.8	7.1	13	
				2 MHz	0.46	0.79	1.75	3.15	5.70	0.70	1.5	3.5	6.8	13	
				1 MHz	0.35	0.68	1.65	3.05	5.60	0.60	1.4	3.4	6.6	13	
			100 KHz	0.25	0.58	1.55	2.95	5.50	0.40	1.2	3.2	6.5	13		
			Range 1 Boost mode	120 MHz	17.50	18.00	19.50	21.50	25.00	21	23	26	30	38	
			Range 1 Normal mode	80 MHz	11.00	11.50	13.00	15.00	18.00	13	14	17	21	28	
				72 MHz	10.00	10.50	12.00	13.50	17.00	12	13	16	20	27	
				64 MHz	9.05	9.55	11.00	12.50	16.00	11	12	14	18	26	
				48 MHz	6.80	7.30	8.55	10.50	13.50	8.30	9.3	12	16	23	
				32 MHz	4.65	5.10	6.35	8.10	11.00	5.70	6.7	9.2	14	21	
24 MHz	3.55	4.00		5.25	7.00	10.00	4.40	5.4	7.9	12	19				
16 MHz	2.45	2.90	4.10	5.85	8.90	3.10	4.1	6.6	11	18					
IDD (LPRun)	Supply current in Low-power run mode	fHCLK = fMSI all peripherals disable	2 MHz	420	825	1900	3500	6250	690	1600	4000	7700	14000	µA	
			1 MHz	295	680	1800	3350	6100	490	1500	3900	7500	14000		
			400 KHz	205	610	1700	3250	6050	430	1400	3800	7500	14000		
			100 KHz	190	565	1650	3250	6000	380	1400	3700	7400	14000		

1. Guaranteed by characterization results, unless otherwise specified.

Table 26. Current consumption in Run and Low-power run modes, code with data processing running from Flash in dual bank, ART enable (Cache ON Prefetch OFF)

Symbol	Parameter	Conditions		fHCLK	TYP					MAX ⁽¹⁾					Unit
		-	Voltage scaling		25°C	55°C	85°C	105°C	125°C	25°C	55°C	85°C	105°C	125°C	
IDD (Run)	Supply current in Run mode	fHCLK = fHSE up to 48MHz included, bypass mode PLL ON above 48 MHz all peripherals disable	Range 2	26 MHz	3.30	3.65	4.65	6.10	8.75	4.2	5	7.1	11	17	mA
				16 MHz	2.10	2.45	3.45	4.90	7.50	2.7	3.6	5.6	8.9	15	
				8 MHz	1.20	1.50	2.50	3.95	6.55	1.6	2.4	4.4	7.7	14	
				4 MHz	0.71	1.05	2.05	3.45	6.05	1	1.8	3.8	7.1	14	
				2 MHz	0.47	0.80	1.80	3.20	5.80	0.7	1.5	3.5	6.8	13	
				1 MHz	0.36	0.68	1.65	3.10	5.65	0.6	1.4	3.4	6.7	13	
				100 KHz	0.25	0.58	1.55	2.95	5.55	0.4	1.2	3.2	6.5	13	
			Range 1 Boost Mode	120 MHz	16.00	16.50	18.00	20.00	23.50	19	21	24	28	36	
			Range 1 Normal Mode	80 MHz	11.50	12.50	13.50	15.50	18.50	14	15	18	22	29	
				72 MHz	10.50	11.00	12.50	14.50	17.50	13	14	17	21	28	
				64 MHz	9.50	10.00	11.50	13.00	16.50	12	13	15	19	26	
				48 MHz	7.10	7.60	8.90	10.50	14.00	8.7	9.9	13	17	24	
				32 MHz	4.85	5.30	6.60	8.35	11.50	6.1	7.1	9.6	14	21	
				24 MHz	3.70	4.15	5.45	7.15	10.50	4.7	5.7	8.2	13	20	
16 MHz	2.60	3.00		4.25	6.00	9.10	3.3	4.3	6.8	11	18				
IDD (LPRun)	Supply current in Low-power run mode	fHCLK = fMSI all peripherals disable	2 MHz	420	840	1950	3550	6350	660	1700	4100	7700	15000	µA	
			1 MHz	305	700	1800	3400	6200	540	1500	3900	7600	14000		
			400 KHz	225	610	1700	3300	6100	410	1400	3800	7500	14000		
			100 KHz	185	570	1700	3250	6050	370	1400	3700	7500	14000		



1. Guaranteed by characterization results, unless otherwise specified.

Table 27. Current consumption in Run and Low-power run modes, code with data processing running from Flash in single bank, ART disable

Symbol	Parameter	Conditions		fHCLK	TYP					MAX ⁽¹⁾					Unit	
		-	Voltage scaling		25°C	55°C	85°C	105°C	125°C	25°C	55°C	85°C	105°C	125°C		
IDD(Run)	Supply current in Run mode	fHCLK = fHSE up to 48MHz included, bypass mode PLL ON above 48 MHz all peripherals disable	Range 2	26 MHz	3.60	3.95	4.95	6.40	9.00	4.6	5.5	7.5	11	17	mA	
				16 MHz	2.40	2.75	3.75	5.15	7.70	3.1	4	6	9.3	16		
				8 MHz	1.35	1.70	2.70	4.10	6.60	1.8	2.6	4.6	7.9	14		
				4 MHz	0.79	1.10	2.10	3.50	6.05	1.2	1.9	3.9	7.2	14		
				2 MHz	0.51	0.84	1.80	3.20	5.75	0.77	1.6	3.6	6.8	13		
				1 MHz	0.38	0.70	1.65	3.05	5.70	0.6	1.4	3.4	6.7	13		
				100 KHz	0.25	0.58	1.55	2.95	5.50	0.44	1.2	3.2	6.5	13		
			Range 1 Boost Mode	120 MHz	17.50	18.00	19.50	21.50	25.00	21	23	26	30	38		
				Range 1 Normal Mode	80 MHz	12.50	13.00	14.00	16.00	19.00	15	17	19	23		30
					72 MHz	11.00	11.50	13.00	15.00	18.00	14	15	18	22		29
					64 MHz	10.00	10.50	12.00	13.50	17.00	12	14	16	20		28
					48 MHz	7.90	8.40	9.70	11.50	14.50	9.8	12	14	18		25
					32 MHz	5.60	6.10	7.35	9.10	12.00	7	8.2	11	15		22
					24 MHz	4.25	4.75	5.10	7.75	11.00	5.4	6.5	9	13		20
16 MHz	3.00	3.45	4.70		6.45	9.50	3.9	4.9	7.4	12	19					
IDD (LPRun)	Supply current in Low-power run mode	fHCLK = fMSI all peripherals disable	2 MHz	495	890	2000	3550	6300	810	1700	4100	7800	15000	µA		
			1 MHz	315	720	1800	3400	6150	560	1500	3900	7600	14000			
			400 KHz	235	615	1700	3250	6050	420	1400	3800	7500	14000			
			100 KHz	175	575	1700	3200	6000	400	1400	3700	7400	14000			



1. Guaranteed by characterization results, unless otherwise specified.

Table 28. Current consumption in Run and Low-power run modes, code with data processing running from Flash in dual bank, ART disable

Symbol	Parameter	Conditions		fHCLK	TYP					MAX ⁽¹⁾					Unit
		-	Voltage scaling		25°C	55°C	85°C	105°C	125°C	25°C	55°C	85°C	105°C	125°C	
IDD (Run)	Supply current in Run mode	fHCLK = fHSE up to 48MHz included, bypass mode PLL ON above 48 MHz all peripherals disable	Range 2	26 MHz	3.70	4.05	5.05	6.50	9.15	4.7	5.6	7.6	11	17	mA
				16 MHz	2.50	2.85	3.85	5.25	7.85	3.2	4.1	6.1	9.4	16	
				8 MHz	1.10	1.75	2.75	4.15	6.75	1.7	2.7	4.7	8	14	
				4 MHz	0.82	1.15	2.15	3.55	6.15	1.2	2	4	7.3	14	
				2 MHz	0.53	0.86	1.85	3.20	5.85	0.8	1.6	3.6	6.9	13	
				1 MHz	0.38	0.71	1.70	3.05	5.70	0.6	1.4	3.4	6.7	13	
				100 KHz	0.25	0.58	1.55	2.95	5.55	0.4	1.2	3.2	6.5	13	
			Range 1 Boost Mode	120 MHz	16.00	16.50	18.00	20.00	23.50	19	21	24	28	36	
			Range 1 Normal Mode	80 MHz	12.00	12.50	14.00	16.00	19.00	15	16	19	23	30	
				72 MHz	11.00	11.50	13.00	14.50	18.00	13	15	18	22	29	
				64 MHz	9.65	10.00	11.50	13.50	16.50	12	13	16	20	27	
				48 MHz	8.10	8.65	9.95	12.00	15.00	11	12	15	19	26	
				32 MHz	5.80	6.30	7.60	9.35	12.50	7.3	8.5	12	16	23	
				24 MHz	4.45	4.90	6.15	7.95	11.00	5.6	6.7	9.3	14	21	
16 MHz	3.15	3.65	4.85	6.65	9.75	4.1	5.2	7.7	12	19					



**Table 28. Current consumption in Run and Low-power run modes,
code with data processing running from Flash in dual bank, ART disable (continued)**

Symbol	Parameter	Conditions		fHCLK	TYP					MAX ⁽¹⁾					Unit
		-	Voltage scaling		25°C	55°C	85°C	105°C	125°C	25°C	55°C	85°C	105°C	125°C	
IDD (LPRun)	Supply current in Low-power run mode	fHCLK = fMSI all peripherals disable		2 MHz	485	900	2000	3600	6400	800	1800	4200	7800	15000	μA
				1 MHz	355	730	1850	3400	6200	580	1600	4000	7600	14000	
				400 KHz	240	630	1750	3300	6100	420	1400	3800	7500	14000	
				100 KHz	175	580	1700	3250	6050	370	1400	3700	7500	14000	

1. Guaranteed by characterization results, unless otherwise specified.

Table 29. Current consumption in Run and Low-power run modes, code with data processing running from SRAM1

Symbol	Parameter	Conditions		fHCLK	TYP					MAX ⁽¹⁾					Unit
		-	Voltage scaling		25°C	55°C	85°C	105°C	125°C	25°C	55°C	85°C	105°C	125°C	
IDD(Run)	Supply current in Run mode	fHCLK = fHSE up to 48MHz included, bypass mode PLL ON above 48 MHz all peripherals disable	Range 2	26 MHz	3.00	3.35	4.35	5.80	8.40	4.7	5.6	7.6	11	17	mA
				16 MHz	1.95	2.25	3.25	4.70	7.30	3.2	4.1	6.1	9.4	16	
				8 MHz	1.10	1.40	2.40	3.85	6.45	1.7	2.7	4.7	8	14	
				4 MHz	0.66	0.99	1.95	3.40	5.95	1.2	2	4	7.3	14	
				2 MHz	0.45	0.78	1.75	3.20	5.75	0.79	1.6	3.6	6.9	13	
				1 MHz	0.34	0.67	1.65	3.05	5.65	0.61	1.4	3.4	6.7	13	
				100 KHz	0.25	0.58	1.55	2.95	5.55	0.44	1.2	3.2	6.5	13	
			Range 1 Boost Mode	120 MHz	17.00	17.50	19.00	21.00	24.50	19	21	24	28	36	
			Range 1 Normal Mode	80 MHz	10.50	11.00	12.50	14.50	17.50	15	16	19	23	30	
				72 MHz	9.55	10.00	11.50	13.00	16.50	13	15	18	22	29	
				64 MHz	8.55	9.05	10.50	12.00	15.50	12	13	16	20	27	
				48 MHz	6.45	6.90	8.20	10.00	13.00	11	12	15	19	26	
				32 MHz	4.40	4.85	6.10	7.90	11.00	7.3	8.5	12	16	23	
				24 MHz	3.35	3.80	5.10	6.85	9.95	5.6	6.7	9.3	14	21	
16 MHz	2.35	2.80		4.05	5.75	8.90	4.1	5.2	7.7	12	19				
IDD (LPRun)	Supply current in Low-power run mode	fHCLK = fMSI all peripherals disable FLASH in power-down	2 MHz	415	805	1900	3500	6300	800	1800	4200	7800	15000	μA	
			1 MHz	300	660	1800	3350	6150	580	1600	4000	7600	14000		
			400 KHz	205	595	1700	3250	6100	420	1400	3800	7500	14000		
			100 KHz	160	560	1650	3250	6050	370	1400	3700	7500	14000		

1. Guaranteed by characterization results, unless otherwise specified.



Table 30. Typical current consumption in Run and Low-power run modes, with different codes running from Flash, ART enable (Cache ON Prefetch OFF)

Symbol	Parameter	Conditions		Code	TYP	TYP	Unit	TYP	TYP	Unit
		-	Voltage scaling		Single Bank Mode	Dual Bank Mode		Single Bank Mode	Dual Bank Mode	
					25°C	25°C		25°C	25°C	
IDD (Run)	Supply current in Run mode	fHCLK=fHSE up to 48 MHz included, bypass mode PLL ON above 48 MHz all peripherals disable	Range2 fHCLK=26MHz	Reduced code ⁽¹⁾	3.15	3.30	mA	121	127	μA/MHz
				Coremark	3.40	3.50		131	135	
				Dhrystone2.1	3.80	3.70		146	142	
				Fibonacci	3.45	3.65		133	140	
				While ⁽¹⁾	2.90	2.90		112	112	
			Range 1 Normal Mode fHCLK= 80 MHz	Reduced code ⁽¹⁾	11.00	11.50	mA	138	144	μA/MHz
				Coremark	12.50	13.00		156	163	
				Dhrystone2.1	13.50	13.00		169	163	
				Fibonacci	12.50	13.00		156	163	
				While ⁽¹⁾	10.50	10.50		131	131	
			Range 1 Boost Mode fHCLK= 120 MHz	Reduced code ⁽¹⁾	17.50	16.00	mA	146	133	μA/MHz
				Coremark	19.50	20.00		163	167	
				Dhrystone2.1	20.00	20.50		167	171	
				Fibonacci	19.50	21.00		163	175	
				While ⁽¹⁾	16.50	16.50		138	138	
IDD (LPRun)	Supply current in Low-power run	fHCLK = fMSI = 2MHz all peripherals disable	Reduced code ⁽¹⁾	420	420	μA	210	210	μA/MHz	
			Coremark	435	475		218	238		
			Dhrystone2.1	485	505		243	253		
			Fibonacci	475	475		238	238		
			While ⁽¹⁾	375	515		188	258		

1. Reduced code used for characterization results provided in [Table 25](#), [Table 27](#), [Table 29](#).

Table 31. Typical current consumption in Run and Low-power run modes, with different codes running from Flash, ART disable

Symbol	Parameter	Conditions		Code	TYP Single Bank Mode	TYP Dual Bank Mode	Unit	TYP Single Bank Mode	TYP Dual Bank Mode	Unit
		-	Voltage scaling		25°C	25°C		25°C	25°C	
IDD (Run)	Supply current in Run mode	fHCLK=fHSE up to 48 MHz included, bypass mode PLL ON above 48 MHz all peripherals disable	Range2 fHCLK=26MHz	Reduced code ⁽¹⁾	3.60	3.70	mA	138	142	μA/MHz
				Coremark	3.75	3.45		144	133	
				Dhrystone2.1	3.80	3.50		146	135	
				Fibonacci	3.70	3.20		142	123	
				While ⁽¹⁾	2.90	2.90		112	112	
			Range 1 Normal Mode fHCLK= 80 MHz	Reduced code ⁽¹⁾	12.50	12.00	mA	156	150	μA/MHz
				Coremark	12.50	11.00		156	138	
				Dhrystone2.1	12.50	11.50		156	144	
				Fibonacci	12.00	10.50		150	131	
				While ⁽¹⁾	10.50	10.50		131	131	
			Range 1 Boost Mode fHCLK= 120 MHz	Reduced code ⁽¹⁾	17.50	16.00	mA	146	133	μA/MHz
				Coremark	17.00	15.00		142	125	
				Dhrystone2.1	17.00	15.00		142	125	
				Fibonacci	17.00	14.00		142	117	
				While ⁽¹⁾	16.50	16.50		138	138	



Table 31. Typical current consumption in Run and Low-power run modes, with different codes running from Flash, ART disable (continued)

Symbol	Parameter	Conditions		Code	TYP Single Bank Mode	TYP Dual Bank Mode	Unit	TYP Single Bank Mode	TYP Dual Bank Mode	Unit
		-	Voltage scaling		25°C	25°C		25°C	25°C	
IDD (LPRun)	Supply current in Low-power run	fHCLK = fMSI = 2MHz all peripherals disable		Reduced code ⁽¹⁾	495	485	µA	248	243	µA/MHz
				Coremark	540	550		270	275	
				Dhrystone2.1	600	565		300	283	
				Fibonacci	550	545		275	273	
				While ⁽¹⁾	365	720		183	360	

1. Reduced code used for characterization results provided in [Table 25](#), [Table 27](#), [Table 29](#).

Table 32. Typical current consumption in Run and Low-power run modes, with different codes running from SRAM1

Symbol	Parameter	Conditions		Code	TYP	Unit	TYP	Unit
		-	Voltage scaling		25°C		25°C	
IDD (Run)	Supply current in Run mode	fHCLK=fHSE up to 48 MHz included, bypass mode PLL ON above 48 MHz all peripherals disable	Range2 fHCLK=26 MHz	Reduced code ⁽¹⁾	3.00	mA	115	μA/MHz
				Coremark	3.00		115	
				Dhystone2.1	3.00		115	
				Fibonacci	2.95		113	
				While ⁽¹⁾	2.85		110	
			Range 1 Normal Mode fHCLK= 80 MHz	Reduced code ⁽¹⁾	10.50	mA	131	μA/MHz
				Coremark	9.90		124	
				Dhystone2.1	10.50		131	
				Fibonacci	10.50		131	
				While ⁽¹⁾	10.50		131	
			Range 1 Boost Mode fHCLK= 120 MHz	Reduced code ⁽¹⁾	17.00	mA	142	μA/MHz
				Coremark	15.50		129	
				Dhystone2.1	17.00		142	
				Fibonacci	16.50		138	
				While ⁽¹⁾	16.50		138	
IDD(LPRun)	Supply current in Low-power run	fHCLK = fMSI = 2MHz all peripherals disable	Reduced code ⁽¹⁾	415	μA	208	μA/MHz	
			Coremark	300		150		
			Dhystone2.1	425		213		
			Fibonacci	410		205		
			While ⁽¹⁾	420		210		

1. Reduced code used for characterization results provided in [Table 25](#), [Table 27](#), [Table 29](#).



Table 33. Current consumption in Sleep and Low-power sleep mode, Flash ON

Symbol	Parameter	Conditions		fHCLK	TYP					MAX ⁽¹⁾					Unit
		-	Voltage scaling		25°C	55°C	85°C	105°C	125°C	25°C	55°C	85°C	105°C	125°C	
IDD (Sleep)	Supply current in Sleep mode	fHCLK = fHSE up to 48MHz included, bypass mode PLL ON above 48 MHz all peripherals disable	Range 2	26 MHz	0.98	1.30	2.30	3.75	6.35	1.4	2.2	4.2	7.5	14	mA
				16 MHz	0.70	1.05	2.00	3.45	6.05	1	1.8	3.8	7.1	14	
				8 MHz	0.47	0.80	1.75	3.20	5.80	0.72	1.5	3.5	6.8	13	
				4 MHz	0.35	0.69	1.65	3.10	5.65	0.57	1.4	3.4	6.7	13	
				2 MHz	0.30	0.63	1.60	3.00	5.60	0.5	1.3	3.3	6.6	13	
				1 MHz	0.27	0.60	1.55	3.00	5.55	0.46	1.3	3.3	6.5	13	
			100 KHz	0.24	0.57	1.55	2.95	5.55	0.43	1.2	3.2	6.5	13		
			Range 1 Boost Mode	120 MHz	4.70	5.20	6.55	8.50	12.00	4.8	6	8.7	13	21	
			Range 1 Normal Mode	80 MHz	3.05	3.50	4.75	6.50	9.60	3.3	4.3	6.8	11	18	
				72 MHz	2.80	3.25	4.45	6.20	9.35	3	4	6.5	11	18	
				64 MHz	2.50	2.95	4.20	5.95	9.05	2.7	3.8	6.3	11	18	
				48 MHz	1.90	2.35	3.60	5.35	8.45	2.6	3.5	6	10	18	
				32 MHz	1.40	1.80	3.05	4.80	7.90	1.9	2.9	5.4	9.3	17	
24 MHz	1.10	1.55		2.75	4.50	7.60	1.6	2.5	5	9	16				
16 MHz	0.84	1.25	2.50	4.20	7.30	1.2	2.2	4.7	8.6	16					
IDD (LPSleep)	Supply current in Low-power sleep mode	fHCLK = fMSI all peripherals disable	2 MHz	235	620	1750	3300	6100	430	1400	3800	7500	14000	μA	
			1 MHz	205	590	1700	3300	6050	390	1400	3700	7500	14000		
			400 KHz	190	575	1700	3250	6050	390	1300	3700	7500	14000		
			100 KHz	170	560	1650	3250	6050	380	1300	3700	7500	14000		

1. Guaranteed by characterization results, unless otherwise specified.

Table 34. Current consumption in Low-power sleep mode, Flash in power-down

Symbol	Parameter	Conditions		fHCLK	TYP					MAX ⁽¹⁾					Unit
		-	Voltage scaling		25°C	55°C	85°C	105°C	125°C	25°C	55°C	85°C	105°C	125°C	
IDD (LPSleep)	Supply current in Low-power sleep mode	fHCLK = fMSI all peripherals disable		2 MHz	225	605	1700	3300	6100	430	1400	3700	7400	14000	μA
				1 MHz	185	585	1700	3250	6050	370	1300	3700	7400	14000	
				400 KHz	165	560	1650	3250	6050	350	1300	3700	7400	14000	
				100 KHz	155	550	1650	3250	6050	340	1300	3700	7400	14000	

1. Guaranteed by characterization results, unless otherwise specified.

Table 35. Current consumption in Stop 2 mode, SRAM3 disabled

Symbol	Parameter	Conditions		TYP					MAX ⁽¹⁾					Unit		
		-	VDD	25°C	55°C	85°C	105°C	125°C	25°C	55°C	85°C	105°C	125°C			
IDD (Stop 2)	Supply current in Stop 2 mode, RTC disabled	RTC disabled		1.8 V	2.90	11.0	42.0	96.5	210	7.7	30	120	255	457	μA	
				2.4 V	2.90	11.0	42.5	97.5	215	8	31	120	256	467		
				3 V	2.95	11.0	43.0	98.5	215	8	31	120	258	478		
				3.6 V	3.00	11.5	43.5	100.0	220	8.3	32	120	261	557		
		RTC disabled ENULP=1		1.8 V	2.80	-	-	-	-	-	-	-	-	-		-
				2.4 V	2.85	-	-	-	-	-	-	-	-	-		-
				3 V	2.95	-	-	-	-	-	-	-	-	-		-
				3.6 V	2.95	-	-	-	-	-	-	-	-	-		-



Table 35. Current consumption in Stop 2 mode, SRAM3 disabled (continued)

Symbol	Parameter	Conditions		TYP					MAX ⁽¹⁾					Unit		
		-	VDD	25°C	55°C	85°C	105°C	125°C	25°C	55°C	85°C	105°C	125°C			
IDD (Stop 2 with RTC)	Supply current in Stop 2 mode, RTC enabled	RTC clocked by LSI	1.8 V	3.30	11.5	42.5	97.0	210	8.3	31	120	255	461	μA		
			2.4 V	3.40	11.5	43.0	98.0	215	8.5	32	120	256	468			
			3 V	3.60	12.0	43.5	99.0	215	8.6	32	120	258	480			
			3.6 V	3.90	12.0	44.5	100.0	225	9	33	120	262	559			
		RTC clocked by LSI ENULP=1 LPCAL=1 LSIPREDIV = 1	1.8 V	2.95	-	-	-	-	-	-	-	-	-		-	
			2.4 V	2.95	-	-	-	-	-	-	-	-	-		-	
			3 V	3.05	-	-	-	-	-	-	-	-	-		-	
		RTC clocked by LSE bypassed at 32768 Hz ENULP=1 LPCAL=1 LSESYSDIS =1	1.8 V	2.90	-	-	-	-	-	-	-	-	-		-	
			2.4 V	2.95	-	-	-	-	-	-	-	-	-		-	
			3 V	3.00	-	-	-	-	-	-	-	-	-		-	
		RTC clocked by LSE quartz in low drive mode	1.8 V	2.90	-	-	-	-	-	-	-	-	-		-	
			2.4 V	2.95	-	-	-	-	-	-	-	-	-		-	
			3 V	3.00	-	-	-	-	-	-	-	-	-		-	
		RTC clocked by LSE quartz in low drive mode ENULP = 1 LPCAL = 1 LSESYSDIS =1	1.8 V	2.95	-	-	-	-	-	-	-	-	-		-	
			2.4 V	2.95	-	-	-	-	-	-	-	-	-		-	
			3 V	3.00	-	-	-	-	-	-	-	-	-		-	
					3.6 V	3.15	-	-	-	-	-	-	-		-	-
					1.8 V	3.45	12.0	44.0	100.0	-	-	-	-		-	-
					2.4 V	3.65	12.0	45.0	100.0	-	-	-	-		-	-
					3 V	3.95	12.5	45.5	105.0	-	-	-	-		-	
3.6 V	4.30				13.0	46.5	105.0	-	-	-	-	-				
1.8 V	2.95				11.0	42.5	96.5	-	-	-	-	-				
			2.4 V	2.95	11.0	43.0	97.5	-	-	-	-	-				
			3 V	3.10	11.5	43.5	98.5	-	-	-	-	-				
			3.6 V	3.25	12.0	45.0	100.0	-	-	-	-	-				
IDD (wakeup from Stop 2)	Supply current during wakeup from Stop 2	Wakeup clock is MSI = 48 MHz, voltage Range 1 ⁽²⁾	3 V	2.65	-	-	-	-	-	-	-	-	mA			
		Wakeup clock is MSI = 4 MHz, voltage Range 2 ⁽²⁾	3 V	0.954	-	-	-	-	-	-	-	-				
		Wakeup clock is HSI = 16 MHz, voltage Range 1 ⁽²⁾	3 V	1.6	-	-	-	-	-	-	-	-				

1. Guaranteed by characterization results, unless otherwise specified.

2. Wakeup with code execution from Flash. Average value given for a typical wakeup time as specified in [Table 43: Low-power mode wakeup timings](#).

Table 36. Current consumption in Stop 2 mode, SRAM3 enabled

Symbol	Parameter	Conditions		TYP					MAX ⁽¹⁾					Unit	
		-	VDD	25°C	55°C	85°C	105°C	125°C	25°C	55°C	85°C	105°C	125°C		
IDD (Stop 2)	Supply current in Stop 2 mode, RTC disabled	RTC disabled	1.8 V	3.55	14.0	55.0	125	280	-	-	-	-	-	µA	
			2.4 V	3.60	14.0	55.5	130	280	-	-	-	-	-		
			3 V	3.65	14.0	56.0	130	285	-	-	-	-	-		
			3.6 V	3.70	14.5	56.5	130	290	-	-	-	-	-		
		RTC disabled ENULP=1	1.8 V	3.45	-	-	-	-	-	-	-	-	-		-
			2.4 V	3.45	-	-	-	-	-	-	-	-	-		-
			3 V	3.55	-	-	-	-	-	-	-	-	-		-
			3.6 V	3.65	-	-	-	-	-	-	-	-	-		-



Table 36. Current consumption in Stop 2 mode, SRAM3 enabled (continued)

Symbol	Parameter	Conditions		TYP					MAX ⁽¹⁾					Unit	
		-	VDD	25°C	55°C	85°C	105°C	125°C	25°C	55°C	85°C	105°C	125°C		
IDD (Stop 2 with RTC)	Supply current in Stop 2 mode, RTC enabled	RTC clocked by LSI	1.8 V	3.95	14.5	55.5	125	280	11	46	155	323	590	μA	
			2.4 V	4.15	14.5	56.0	130	280	12	46	155	235	595		
			3 V	4.20	15.0	56.5	130	285	12	47	158	237	603		
			3.6 V	4.60	15.0	57.5	130	290	12	47	158	331	663		
		RTC clocked by LSI ENULP=1 LPCAL=1 LSIPREDIV=1	1.8 V	3.55	-	-	-	-	-	-	-	-	-		-
			2.4 V	3.55	-	-	-	-	-	-	-	-	-		-
			3 V	3.70	-	-	-	-	-	-	-	-	-		-
			3.6 V	3.80	-	-	-	-	-	-	-	-	-		-
		RTC clocked by LSE bypassed at 32768 Hz ENULP=1 LPCAL=1 LSESYSDIS=1	1.8 V	3.60	-	-	-	-	-	-	-	-	-		-
			2.4 V	3.60	-	-	-	-	-	-	-	-	-		-
			3 V	3.80	-	-	-	-	-	-	-	-	-		-
			3.6 V	3.85	-	-	-	-	-	-	-	-	-		-
		RTC clocked by LSE quartz in low drive mode	1.8 V	4.20	15.0	57.5	135	-	-	-	-	-	-		-
			2.4 V	4.30	15.0	58.5	135	-	-	-	-	-	-		-
			3 V	4.55	15.5	59.0	135	-	-	-	-	-	-		-
			3.6 V	4.85	16.0	60.0	140	-	-	-	-	-	-		-
		RTC clocked by LSE bypassed at 32768 Hz ENULP=1 LPCAL=1 LSESYSDIS =1	1.8 V	3.50	14.0	55.5	125	-	-	-	-	-	-		-
			2.4 V	3.65	14.5	56.0	130	-	-	-	-	-	-		-
			3 V	3.70	14.5	56.5	130	-	-	-	-	-	-		-
			3.6 V	3.95	15.0	57.5	130	-	-	-	-	-	-		-
IDD (wakeup from Stop 2)	Supply current during wakeup from Stop 2 mode	Wakeup clock is MSI = 48 MHz, voltage Range 1 ⁽²⁾	3 V	2.63	-	-	-	-	-	-	-	-	mA		
		Wakeup clock is MSI = 4 MHz, voltage Range 2 ⁽²⁾	3 V	0.91	-	-	-	-	-	-	-	-			
		Wakeup clock is HSI = 16 MHz, voltage Range 1 ⁽²⁾	3 V	1.53	-	-	-	-	-	-	-	-			



1. Guaranteed by characterization results, unless otherwise specified.
2. Wakeup with code execution from Flash. Average value given for a typical wakeup time as specified in [Table 43: Low-power mode wakeup timings](#).

Table 37. Current consumption in Stop 1 mode

Symbol	Parameter	Conditions		TYP					MAX ⁽¹⁾					Unit
		-	VDD	25°C	55°C	85°C	105°C	125°C	25°C	55°C	85°C	105°C	125°C	
IDD (Stop 1)	Supply current in Stop 1 mode, RTC disabled	-	1.8 V	115	400	1250	2450	4600	280	1100	3300	6500	13000	μA
			2.4 V	115	405	1250	2500	4650	280	1100	3300	6500	13000	
			3 V	115	405	1250	2500	4650	280	1100	3300	6500	13000	
			3.6 V	115	405	1250	2500	4700	280	1100	3300	6600	13000	
IDD (Stop 1 with RTC)	Supply current in STOP 1 mode, RTC enabled	RTC clocked by LSI	1.8 V	115	400	1250	2450	4600	280	1100	3300	6500	13000	
			2.4 V	115	405	1250	2500	4650	280	1100	3300	6500	13000	
			3 V	115	405	1250	2500	4650	280	1100	3300	6600	13000	
			3.6 V	115	410	1300	2500	4700	280	1100	3300	6600	13000	
		RTC clocked by LSE bypassed at 32768 Hz	1.8 V	115	405	1250	2450	4600	300	1100	3500	6900	13000	
			2.4 V	115	405	1250	2500	4650	300	1100	3500	6900	13000	
			3 V	115	405	1250	2500	4650	320	1100	3500	6900	13000	
			3.6 V	115	405	1250	2500	4700	320	1100	3500	6900	13000	
		RTC clocked by LSE quartz ⁽²⁾ in low drive mode	1.8 V	115	415	1300	2550	-	300	1100	3400	6800	-	
			2.4 V	115	420	1300	2550	-	300	1100	3400	6800	-	
			3 V	120	420	1300	2550	-	300	1100	3400	6800	-	
			3.6 V	120	420	1300	2600	-	300	1100	3400	6800	-	
IDD (wakeup from Stop 1)	Supply current during wakeup from Stop 1 mode	Wakeup clock is MSI = 48 MHz, voltage Range 1 ⁽³⁾	3 V	2.57	-	-	-	-	-	-	-	-	mA	
		Wakeup clock is MSI = 4 MHz, voltage Range 2 ⁽³⁾	3 V	0.67	-	-	-	-	-	-	-	-		
		Wakeup clock is HSI = 16 MHz, voltage Range 1 ⁽³⁾	3 V	1.35	-	-	-	-	-	-	-	-		

1. Guaranteed by characterization results, unless otherwise specified.
2. Based on characterization done with a 32.768 kHz crystal (MC306-G-06Q-32.768, manufacturer JFVNY) with two 6.8 pF loading capacitors.
3. Wakeup with code execution from Flash. Average value given for a typical wakeup time as specified in [Table 43: Low-power mode wakeup timings](#)

Table 38. Current consumption in Stop 0 mode

Symbol	Parameter	Conditions		TYP					MAX ⁽¹⁾					Unit
		-	V _{DD}	25°C	55°C	85°C	105°C	125°C	25°C	55°C	85°C	105°C	125°C	
I _{DD} (Stop 0)	Supply current in Stop 0 mode, RTC disabled	-	1.8 V	275	695	1850	3550	6450	560	1600	4500	8700	16000	μA
			2.4 V	275	695	1850	3550	6450	560	1600	4500	8700	17000	
			3 V	275	695	1850	3500	6450	570	1600	4500	8800	17000	
			3.6 V	280	695	1900	3550	6500	570	1600	4500	8800	17000	

1. Guaranteed by characterization results, unless otherwise specified.

Table 39. Current consumption in Standby mode

Symbol	Parameter	Conditions		TYP					MAX ⁽¹⁾					Unit	
		-	V _{DD}	25°C	55°C	85°C	105°C	125°C	25°C	55°C	85°C	105°C	125°C		
I _{DD} (Standby)	Supply current in Standby mode (backup registers retained), RTC disabled	No independent watchdog	1.8 V	110	325	1650	4650	13000	178.2	760	3372	8655	22689	nA	
			2.4 V	120	380	1900	5350	15000	231.5	950	3933	9811	25497		
			3 V	135	455	2250	6450	17500	269.7	1100	4546	11031	28515		
			3.6 V	175	560	2750	7700	21000	360.6	1380	5249	12553	32039		
		No independent watchdog ENULP=1	1.8 V	42	-	-	-	-	-	-	-	-	-		-
			2.4 V	60	-	-	-	-	-	-	-	-	-		-
			3 V	86	-	-	-	-	-	-	-	-	-		-
			3.6 V	135	-	-	-	-	-	-	-	-	-		-
		With independent watchdog	1.8 V	340	-	-	-	-	-	-	-	-	-		-
			2.4 V	415	-	-	-	-	-	-	-	-	-		-
			3 V	500	-	-	-	-	-	-	-	-	-		-
			3.6 V	605	-	-	-	-	-	-	-	-	-		-
		With independent watchdog ENULP=1	1.8 V	270	-	-	-	-	-	-	-	-	-		-
			2.4 V	350	-	-	-	-	-	-	-	-	-		-
			3 V	445	-	-	-	-	-	-	-	-	-		-
			3.6 V	565	-	-	-	-	-	-	-	-	-		-



Table 39. Current consumption in Standby mode (continued)

Symbol	Parameter	Conditions		TYP					MAX ⁽¹⁾					Unit	
		-	V _{DD}	25°C	55°C	85°C	105°C	125°C	25°C	55°C	85°C	105°C	125°C		
IDD (Standby with RTC)	Supply current in Standby mode (backup registers retained), RTC enabled	RTC clocked by LSI, no independent watchdog	1.8 V	500	715	2000	4900	13000	579.5	1400	3744	9000	22911	nA	
			2.4 V	635	890	2350	5750	15000	764.9	1700	4430	10264	25849		
			3 V	790	1100	2850	6850	17500	942.5	1900	5145	11649	28975		
			3.6 V	980	1350	3550	8400	21500	1173	2400	6040	13305	32681		
		RTC clocked by LSI, with independent watchdog	1.8 V	550	-	-	-	-	-	-	-	-	-		-
			2.4 V	705	-	-	-	-	-	-	-	-	-		-
			3 V	885	-	-	-	-	-	-	-	-	-		-
			3.6 V	1100	-	-	-	-	-	-	-	-	-		-
		RTC clocked by LSI, with independent watchdog ENULP = 1 LPCAL = 1 LSIPREDIV = 1	1.8 V	125	-	-	-	-	-	-	-	-	-		-
			2.4 V	165	-	-	-	-	-	-	-	-	-		-
			3 V	205	-	-	-	-	-	-	-	-	-		-
			3.6 V	265	-	-	-	-	-	-	-	-	-		-

Table 39. Current consumption in Standby mode (continued)

Symbol	Parameter	Conditions		TYP					MAX ⁽¹⁾					Unit	
		-	V _{DD}	25°C	55°C	85°C	105°C	125°C	25°C	55°C	85°C	105°C	125°C		
I _{DD} (Standby with RTC)	Supply current in Standby mode (backup registers retained), RTC enabled	RTC clocked by LSE bypassed at 32768 Hz ENULP = 1 LPCAL = 1 LSESYSDIS = 1	1.8 V	84	370	-	-	-	-	-	-	-	-	nA	
			2.4 V	135	515	-	-	-	-	-	-	-	-		
			3 V	195	690	-	-	-	-	-	-	-	-		
			3.6 V	285	940	-	-	-	-	-	-	-	-		
		RTC clocked by LSE quartz ⁽²⁾ in Low-drive mode	1.8 V	560	810	2200	5350	13500	-	-	-	-	-		
			2.4 V	755	1050	2650	6300	15500	-	-	-	-	-		
			3 V	1000	1350	3200	7550	18500	-	-	-	-	-		
			3.6 V	1300	1750	4050	9250	22000	-	-	-	-	-		
		RTC clocked by LSE quartz ⁽²⁾ in Low-drive mode ENULP=1 LPCAL = 1 LSESYSDIS = 1	1.8 V	190	470	-	-	-	-	-	-	-	-		-
			2.4 V	230	595	-	-	-	-	-	-	-	-		-
			3 V	280	765	-	-	-	-	-	-	-	-		-
			3.6 V	380	1000	-	-	-	-	-	-	-	-		-
I _{DD} (SRAM2 4 KBytes) ⁽³⁾	Supply current in Standby mode when 4 Kbytes of SRAM2 are retained	-	1.8 V	200	-	-	-	-	-	-	-	-	-		
			2.4 V	210	-	-	-	-	-	-	-	-	-		
			3 V	230	-	-	-	-	-	-	-	-	-		
			3.6 V	265	-	-	-	-	-	-	-	-	-		
I _{DD} (SRAM2 64 Kbytes) ⁽³⁾	Supply current in Standby mode when the full 64 Kbytes of SRAM2 are retained	-	1.8 V	529	-	-	-	-	-	-	-	-	-		
			2.4 V	540	-	-	-	-	-	-	-	-	-		
			3 V	556	-	-	-	-	-	-	-	-	-		
			3.6 V	600	-	-	-	-	-	-	-	-	-		
I _{DD} (wakeup from Standby)	Supply current during wakeup from Standby mode	Wakeup clock is MSI = 4 MHz ⁽⁴⁾	3 V	1.52	-	-	-	-	-	-	-	-	mA		

1. Guaranteed by characterization results, unless otherwise specified.



- Based on characterization done with a 32.768 kHz crystal (MC306-G-06Q-32.768, manufacturer JFVNY) with two 6.8 pF loading capacitors.
- The supply current in Standby with SRAM2 mode is: $I_{DD_ALL}(\text{Standby}) + I_{DD_ALL}(\text{SRAM2})$. The supply current in Standby with RTC with SRAM2 mode is: $I_{DD_ALL}(\text{Standby} + \text{RTC}) + I_{DD_ALL}(\text{SRAM2})$.
- Wakeup with code execution from Flash. Average value given for a typical wakeup time as specified in [Table 43: Low-power mode wakeup timings](#).

Table 40. Current consumption in Shutdown mode

Symbol	Parameter	Conditions	TYP					MAX ⁽¹⁾					Unit	
			V _{DD}	25°C	55°C	85°C	105°C	125°C	25°C	55°C	85°C	105°C		125°C
IDD (Shutdown)	Supply current in Shutdown mode (backup registers retained) RTC disabled	-	1.8 V	22	155	985	3100	9400	80	358.3	2190	6643	18540	nA
			2.4 V	31	190	1150	3600	11000	126	445.2	2533	7530	20760	
			3 V	44	240	1450	4400	13000	152	527.2	2900	8517	23240	
			3.6 V	74	340	1900	5650	16500	199	653.7	3410	9798	26260	
IDD (Shutdown with RTC)	Supply current in Shutdown mode (backup registers retained) RTC enabled	RTC clocked by LSE bypassed at 32768 Hz	1.8 V	365	505	1350	3450	9800	-	-	-	-	-	
			2.4 V	565	745	1750	4200	11500	-	-	-	-	-	
			3 V	815	1050	2250	5250	14000	-	-	-	-	-	
			3.6 V	1100	1400	2950	6700	17500	-	-	-	-	-	
		RTC clocked by LSE quartz ⁽²⁾ in Low-drive mode	1.8 V	475	640	1550	3800	10000	-	-	-	-	-	
			2.4 V	665	860	1900	4550	11500	-	-	-	-	-	
			3 V	905	1150	2450	5650	14000	-	-	-	-	-	
			3.6 V	1200	1500	3200	7200	17500	-	-	-	-	-	
		RTC clocked by LSE quartz ⁽²⁾ in Low-drive mode LPCAL = 1 LSESYSDIS = 1	1.8 V	180	-	-	-	-	-	-	-	-	-	
			2.4 V	190	-	-	-	-	-	-	-	-	-	
			3 V	229	-	-	-	-	-	-	-	-	-	
			3.6 V	293	-	-	-	-	-	-	-	-	-	
IDD (wakeup from Shutdown)	Supply current during wakeup from Shutdown mode	Wakeup clock is MSI = 4 MHz ⁽³⁾	3 V	0.538	-	-	-	-	-	-	-	-	mA	

1. Guaranteed by characterization results, unless otherwise specified.

2. Based on characterization done with a 32.768 kHz crystal (MC306-G-06Q-32.768, manufacturer JFVNY) with two 6.8 pF loading capacitors.

3. Wakeup with code execution from Flash. Average value given for a typical wakeup time as specified in [Table 43: Low-power mode wakeup timings](#).

Table 41. Current consumption in VBAT mode

Symbol	Parameter	Conditions		TYP					MAX					Unit	
		-	VBAT	25°C	55°C	85°C	105°C	125°C	25°C	55°C	85°C	105°C	125°C		
IDD(VBAT)	Backup domain supply current	RTC disabled	1.8 V	3	24	155	475	1350	-	-	-	-	-	nA	
			2.4 V	3	28	180	540	1500	-	-	-	-	-		
			3 V	4	36	220	665	1850	-	-	-	-	-		
			3.6 V	12	78	465	1450	4050	-	-	-	-	-		
		RTC enabled and clocked by LSE bypassed at 32768 Hz	1.8 V	345	375	525	865	-	-	-	-	-	-		-
			2.4 V	540	585	760	1150	-	-	-	-	-	-		-
			3 V	780	835	1050	1550	-	-	-	-	-	-		-
			3.6 V	1050	1100	1450	2400	-	-	-	-	-	-		-
		RTC enabled and clocked by LSE quartz ⁽¹⁾	1.8 V	450	-	-	-	-	-	-	-	-	-		-
			2.4 V	635	-	-	-	-	-	-	-	-	-		-
			3 V	865	-	-	-	-	-	-	-	-	-		-
			3.6 V	1150	-	-	-	-	-	-	-	-	-		-
		RTC enabled and clocked by LSE quartz ⁽¹⁾ LPCAL = 1 LSESYSDIS =1	1.8 V	150	-	-	-	-	-	-	-	-	-		-
			2.4 V	165	-	-	-	-	-	-	-	-	-		-
			3 V	195	-	-	-	-	-	-	-	-	-		-
			3.6 V	240	-	-	-	-	-	-	-	-	-		-

1. Based on characterization done with a 32.768 kHz crystal (MC306-G-06Q-32.768, manufacturer JFVNY) with two 6.8 pF loading capacitors.

I/O system current consumption

The current consumption of the I/O system has two components: static and dynamic.

I/O static current consumption

All the I/Os used as inputs with pull-up generate current consumption when the pin is externally held low. The value of this current consumption can be simply computed by using the pull-up/pull-down resistors values given in [Table 62: I/O static characteristics](#).

For the output pins, any external pull-down or external load must also be considered to estimate the current consumption.

Additional I/O current consumption is due to I/Os configured as inputs if an intermediate voltage level is externally applied. This current consumption is caused by the input Schmitt trigger circuits used to discriminate the input value. Unless this specific configuration is required by the application, this supply current consumption can be avoided by configuring these I/Os in analog mode. This is notably the case of ADC input pins which should be configured as analog inputs.

Caution: Any floating input pin can also settle to an intermediate voltage level or switch inadvertently, as a result of external electromagnetic noise. To avoid current consumption related to floating pins, they must either be configured in analog mode, or forced internally to a definite digital value. This can be done either by using pull-up/down resistors or by configuring the pins in output mode.

I/O dynamic current consumption

In addition to the internal peripheral current consumption measured previously (see), the I/Os used by an application also contribute to the current consumption. When an I/O pin switches, it uses the current from the I/O supply voltage to supply the I/O pin circuitry and to charge/discharge the capacitive load (internal or external) connected to the pin:

$$I_{SW} = V_{DDIOx} \times f_{SW} \times C$$

where

I_{SW} is the current sunk by a switching I/O to charge/discharge the capacitive load

V_{DDIOx} is the I/O supply voltage

f_{SW} is the I/O switching frequency

C is the total capacitance seen by the I/O pin: $C = C_{INT} + C_{EXT} + C_S$

C_S is the PCB board capacitance including the pad pin.

The test pin is configured in push-pull output mode and is toggled by software at a fixed frequency.

On-chip peripheral current consumption

The current consumption of the on-chip peripherals is given in [Table 42](#). The MCU is placed under the following conditions:

- All I/O pins are in Analog mode
- The given value is calculated by measuring the difference of the current consumptions:
 - when the peripheral is clocked on
 - when the peripheral is clocked off
- Ambient operating temperature and supply voltage conditions summarized in [Table 18: Voltage characteristics](#)
- The power consumption of the digital part of the on-chip peripherals is given in [Table 42](#). The power consumption of the analog part of the peripherals (where applicable) is indicated in each related section of the datasheet.

Table 42. STM32L4Q5xx peripheral current consumption

Peripheral		Range 1 Boost Mode	Range 1 Normal Mode	Range 2	Low-power run and sleep	Unit
AHB	Bus matrix	3.361	3.144	2.626	2.772	μA/MHz
	ADC independent clock domain	0.254	0.241	0.125	0.750	
	ADC AHB clock domain	4.917	4.591	4.013	4.627	
	AES	3.000	2.750	2.150	3.000	
	CRC	0.534	0.498	0.426	0.405	
	DMA1	2.556	2.386	1.945	2.203	
	DMA2	2.644	2.463	2.030	2.179	
	DMA2D	3.322	3.099	2.525	2.853	
	DMAMUX	5.850	5.447	4.477	5.217	
	DCMI	8.498	7.948	6.563	7.503	
	PSSI	8.498	7.948	6.563	7.503	
	FLASH	5.836	5.442	4.476	5.101	
	FSMC	10.215	9.485	7.807	9.043	
	GPIOA	0.158	0.141	0.142	0.153	
	GPIOB	0.173	0.150	0.160	0.108	
	GPIOC	0.189	0.179	0.164	0.090	
	GIOD	0.181	0.171	0.159	0.134	
	GPIOE	0.157	0.143	0.146	0.151	
	GPIOF	0.178	0.151	0.153	0.340	
	GPIOG	0.171	0.164	0.164	0.162	
GPIOH	0.202	0.193	0.176	0.157		

Table 42. STM32L4Q5xx peripheral current consumption (continued)

Peripheral		Range 1 Boost Mode	Range 1 Normal Mode	Range 2	Low- power run and sleep	Unit
AHB (Cont.)	GPIOI	0.155	0.136	0.134	0.095	μA/MHz
	HASH1	1.808	1.682	1.400	1.530	
	OTG_FS independent clock domain	26.984	29.739	NA	NA	
	OTG_FS AHB clock domain	18.554	17.288	NA	NA	
	OSPIM independent clock domain	0.033	0.013	0.013	0.248	
	OSPIM AHB clock domain	0.120	0.115	0.099	0.087	
	OSPI1 independent clock domain	0.781	0.713	0.601	0.661	
	OSPI1 AHB clock domain	0.330	0.309	0.248	0.223	
	OSPI2 independent clock domain	1.043	0.995	0.799	0.761	
	OSPI2 AHB clock domain	0.334	0.295	0.249	0.218	
	PKA	4.418	4.134	3.443	3.989	
	RNG independent clock domain	1.591	1.601	NA	NA	
	RNG AHB clock domain	1.278	1.203	NA	NA	
	SDMMC1 independent clock domain	18.306	17.327	NA	NA	
	SDMMC1 AHB clock domain	17.582	16.375	NA	NA	
	SDMMC2 independent clock domain	18.176	17.210	NA	NA	
	SDMMC2 AHB clock domain	17.517	16.300	NA	NA	
	SRAM1	1.117	1.016	0.852	0.910	
	SRAM2	3.976	3.709	3.113	3.292	
	SRAM3	1.087	1.026	0.863	1.129	
TSC	1.678	1.580	1.296	1.346		
All AHB peripherals	114.236	106.366	87.640	98.866		

Table 42. STM32L4Q5xx peripheral current consumption (continued)

Peripheral		Range 1 Boost Mode	Range 1 Normal Mode	Range 2	Low- power run and sleep	Unit
APB1	AHB to APB1 bridge	0.217	0.242	0.214	0.460	μA/MHz
	CAN1	5.457	5.076	4.164	4.852	
	CRS	0.342	0.329	0.272	0.332	
	DAC1	3.125	2.925	2.415	2.801	
	I2C1 independent clock domain	3.750	3.400	2.900	2.500	
	I2C1 APB clock domain	1.400	1.400	1.250	2.000	
	I2C2 independent clock domain	3.500	3.400	2.500	3.500	
	I2C2 APB clock domain	1.400	1.250	1.250	1.000	
	I2C3 independent clock domain	3.250	3.150	2.900	3.000	
	I2C3 APB clock domain	1.150	1.000	0.835	1.000	
	I2C4 independent clock domain	3.500	3.250	2.750	3.000	
	I2C4 APB clock domain	1.350	1.250	1.000	1.500	
	LPUART1 independent clock domain	3.150	3.000	2.450	3.000	
	LPUART1 APB clock domain	1.650	1.500	1.300	1.500	
	LPTIM1 independent clock domain	3.368	3.165	2.645	3.101	
	LPTIM1 APB clock domain	1.041	0.957	0.811	0.900	
	LPTIM2 independent clock domain	3.144	2.951	2.424	2.778	
	LPTIM2 APB clock domain	0.883	0.852	0.702	0.824	
	OPAMP	0.326	0.299	0.274	0.950	
	PWR	0.675	0.620	0.524	0.629	
RTCAPB	2.126	1.968	1.629	1.953		
SPI2	2.062	1.934	1.602	1.936		

Table 42. STM32L4Q5xx peripheral current consumption (continued)

Peripheral		Range 1 Boost Mode	Range 1 Normal Mode	Range 2	Low- power run and sleep	Unit
APB1 (Cont.)	SPI3	2.022	1.891	1.561	1.925	μA/MHz
	TIM2	6.192	5.769	4.755	5.751	
	TIM3	5.414	5.051	4.178	5.031	
	TIM4	5.104	4.748	3.901	4.575	
	TIM5	6.204	5.759	4.769	5.413	
	TIM6	1.111	1.026	0.854	0.911	
	TIM7	1.146	1.070	0.882	1.055	
	USART2 independent clock domain	5.350	5.000	4.150	4.500	
	USART2 APB clock domain	3.000	2.750	2.500	2.500	
	USART3 independent clock domain	6.350	6.000	5.000	5.500	
	USART3 APB clock domain	2.600	2.400	2.100	2.500	
	UART4 independent clock domain	5.150	4.900	3.750	4.500	
	UART4 APB clock domain	2.500	2.250	2.100	2.500	
	UART5 independent clock domain	5.400	5.000	4.150	5.000	
	UART5 APB clock domain	2.400	2.250	2.100	2.000	
	WWDG	0.438	0.410	0.352	0.494	
All APB1 on	107.630	100.588	83.653	94.342		

Table 42. STM32L4Q5xx peripheral current consumption (continued)

Peripheral		Range 1 Boost Mode	Range 1 Normal Mode	Range 2	Low- power run and sleep	Unit
APB2	AHB to APB2 bridge	0.136	0.157	0.143	0.565	μA/MHz
	DFSDM	5.432	5.045	4.229	5.355	
	FW	0.388	0.352	0.296	0.286	
	LTDC independent clock domain	39.542	38.367	42.248	NA	
	LTDC APB clock domain	10.490	9.803	8.138	NA	
	SAI1 independent clock domain	3.236	3.022	2.487	2.310	
	SAI1 APB clock domain	2.543	2.376	1.991	2.218	
	SAI2 independent clock domain	3.314	3.080	2.529	2.312	
	SAI2 APB clock domain	2.504	2.338	1.942	2.193	
	SPI1	2.159	2.025	1.677	1.879	
	SYSCFG/VREFBUF/COMP	0.698	0.662	0.546	0.370	
	TIM1	8.846	8.259	6.781	7.915	
	TIM8	9.126	8.479	6.980	7.966	
	TIM15	4.544	4.226	3.510	3.928	
	TIM16	3.158	2.935	2.448	2.825	
	TIM17	3.180	2.950	2.452	2.959	
	USART1 independent clock domain	6.500	6.150	5.250	6.000	
	USART1 APB clock domain	2.900	2.750	2.250	2.000	
All APB2 on	62.690	58.533	48.424	54.434		
ALL		281.81	262.45	216.946	245.38	

6.3.6 Wakeup time from low-power modes and voltage scaling transition times

The wakeup times given in [Table 43](#) are the latency between the event and the execution of the first user instruction.

The device goes in low-power mode after the WFE (wait for event) instruction.

Table 43. Low-power mode wakeup timings⁽¹⁾

Symbol	Parameter	Conditions		Typ	Max	Unit
$t_{WUSLEEP}$	Wakeup time from Sleep mode to Run mode	-		6	7.8	Nb of CPU cycles
$t_{WULPSLEEP}$	Wakeup time from Lowpower sleep mode to Lowpower run mode	Wakeup in Flash with Flash in power-down during low-power sleep mode (SLEEP_PD=1 in FLASH_ACR) and with clock MSI = 2 MHz		7	9	
$t_{WUSTOPO}$	Wake up time from Stop 0 mode to Run mode in Flash	range1	Wakeup clock MSI = 48 MHz	9	9.7	μs
			Wakeup clock HSI16 = 16 MHz	8.3	8.9	
		range2	Wakeup clock MSI = 24 MHz	18.6	19.5	
			Wakeup clock HSI16 = 16 MHz	17.5	18.2	
	Wake up time from Stop 0 mode to Run mode in SRAM1	range1	Wakeup clock MSI = 48 MHz	1.8	2.5	
			Wakeup clock HSI16 = 16 MHz	2.5	2.8	
		range2	Wakeup clock MSI = 24 MHz	2.4	3.1	
			Wakeup clock HSI16 = 16 MHz	2.5	2.9	
			Wakeup clock MSI = 4 MHz	9.2	11.4	

Table 43. Low-power mode wakeup timings⁽¹⁾ (continued)

Symbol	Parameter	Conditions		Typ	Max	Unit
t _{WU} STOP1	Wake up time from Stop 1 mode to Run in Flash	range1	Wakeup clock MSI = 48 MHz	12.6	13.8	μs
			Wakeup clock HSI16 = 16 MHz	12.0	13.0	
		range2	Wakeup clock MSI = 24 MHz	22.1	23.4	
			Wakeup clock HSI16 = 16 MHz	21.2	22.3	
			Wakeup clock MSI = 4 MHz	25.3	26.7	
		Wake up time from Stop 1 mode to Run mode in SRAM1	range1	Wakeup clock MSI = 48 MHz	5.3	
	Wakeup clock HSI16 = 16 MHz			6.2	7.3	
	range2		Wakeup clock MSI = 24 MHz	5.8	6.8	
			Wakeup clock HSI16 = 16 MHz	6.2	7.3	
	Wake up time from Stop 1 mode to Lowpower run mode in Flash	Regulator in lowpower mode (LPR=1 in PWR_CR1)	Wakeup clock MSI = 2 MHz	29	33	
				15.9	19.9	
	Wake up time from Stop 1 mode to Lowpower run mode in SRAM1					
t _{WU} STOP2	Wake up time from Stop 2 mode to Run mode in Flash	range1	Wakeup clock MSI = 48 MHz	13.0	13.9	
			Wakeup clock HSI16 = 16 MHz	12.5	13.4	
		range2	Wakeup clock MSI = 24 MHz	22.6	23.6	
			Wakeup clock HSI16 = 16 MHz	21.7	22.6	
			Wakeup clock MSI = 4 MHz	25.9	27.2	
		Wake up time from Stop 2 mode to Run mode in SRAM1	range1	Wakeup clock MSI = 48 MHz	5.8	6.6
	Wakeup clock HSI16 = 16 MHz			6.8	7.6	
	range2		Wakeup clock MSI = 24 MHz	6.3	7.1	
			Wakeup clock HSI16 = 16 MHz	6.8	7.6	
			Wakeup clock MSI = 4 MHz	11.2	12.4	

Table 43. Low-power mode wakeup timings⁽¹⁾ (continued)

Symbol	Parameter	Conditions	Typ	Max	Unit	
t _{WUSTBY}	Wakeup time from Standby mode to Run mode	range1	Wakeup clock MSI = 8 MHz	36.7	40	
			Wakeup clock MSI = 4 MHz	36.8	40.1	
t _{WUSTBY} SRAM2	Wakeup time from Standby with SRAM2 to Run mode	range1	Wakeup clock MSI = 8 MHz	36.9	40.2	
			Wakeup clock MSI = 4 MHz	36.9	40.3	
t _{WUSHDN}	Wakeup time from Shutdown mode to Run mode	range1	Wakeup clock MSI = 4 MHz	262.6	323.5	

1. Guaranteed by characterization results.

Table 44. Regulator modes transition times⁽¹⁾

Symbol	Parameter	Conditions	Typ	Max	Unit
t _{WULPRUN}	Wakeup time from Low- power run mode to Run mode ⁽²⁾	Code run with MSI 2 MHz	5	7	µs
t _{VOST}	Regulator transition time from Range 2 to Range 1 or Range 1 to Range 2 ⁽³⁾	Code run with MSI 24 MHz	20	40	

1. Guaranteed by characterization results.
2. Time until REGLPF flag is cleared in PWR_SR2.
3. Time until VOSF flag is cleared in PWR_SR2.

Table 45. Wakeup time using USART/LPUART⁽¹⁾

Symbol	Parameter	Conditions	Typ	Max	Unit
t _{WUUSART} t _{WULPUART}	Wakeup time needed to calculate the maximum USART/LPUART baudrate allowing to wakeup up from stop mode when USART/LPUART clock source is HSI	Stop mode 0	-	1.7	µs
		Stop mode 1/2	-	8.5	

1. Guaranteed by characterization results.

6.3.7 External clock source characteristics

High-speed external user clock generated from an external source

In bypass mode the HSE oscillator is switched off and the input pin is a standard GPIO.

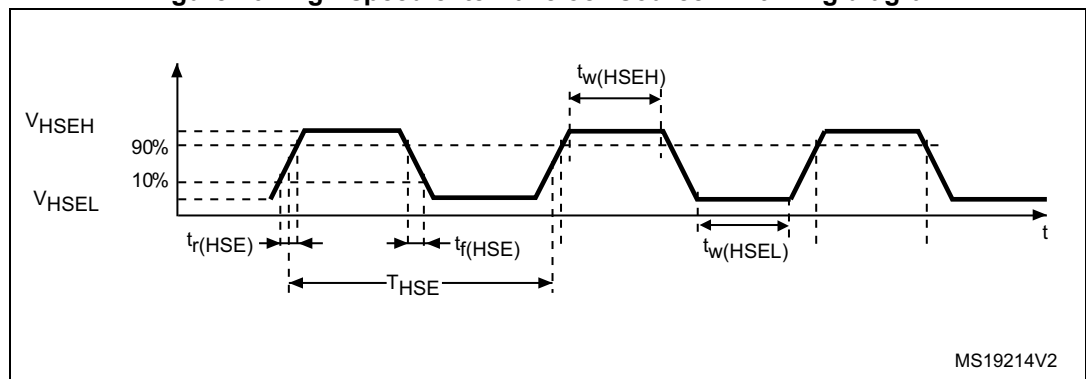
The external clock signal has to respect the I/O characteristics in [Section 6.3.14](#). However, the recommended clock input waveform is shown in [Figure 20: High-speed external clock source AC timing diagram](#).

Table 46. High-speed external user clock characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
f_{HSE_ext}	User external clock source frequency	Voltage scaling Range 1	-	8	48	MHz
		Voltage scaling Range 2	-	8	26	
V_{HSEH}	OSC_IN input pin high level voltage	-	$0.7 V_{DDIOx}$	-	V_{DDIOx}	V
V_{HSEL}	OSC_IN input pin low level voltage	-	V_{SS}	-	$0.3 V_{DDIOx}$	
$t_{w(HSEH)}$ $t_{w(HSEL)}$	OSC_IN high or low time	Voltage scaling Range 1	7	-	-	ns
		Voltage scaling Range 2	18	-	-	

1. Guaranteed by design.

Figure 20. High-speed external clock source AC timing diagram



Low-speed external user clock generated from an external source

In bypass mode the LSE oscillator is switched off and the input pin is a standard GPIO.

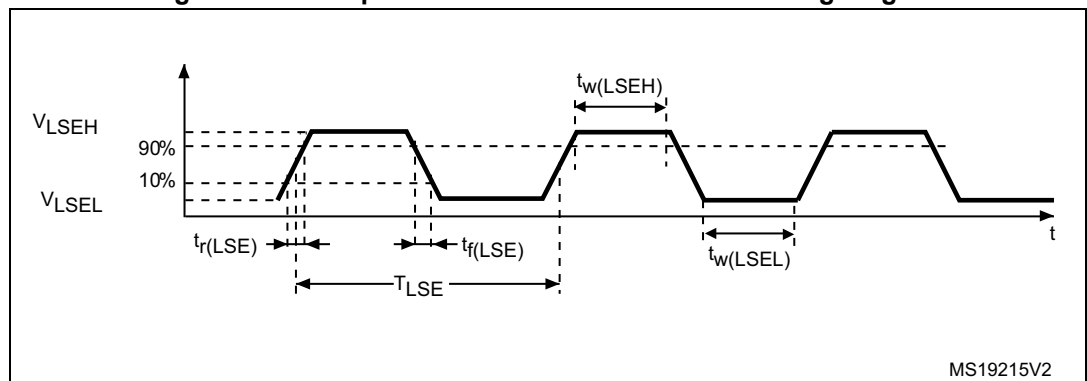
The external clock signal has to respect the I/O characteristics in [Section 6.3.14](#). However, the recommended clock input waveform is shown in [Figure 21](#).

Table 47. Low-speed external user clock characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
f_{LSE_ext}	User external clock source frequency	-	-	32.768	1000	kHz
V_{LSEH}	OSC32_IN input pin high level voltage	-	$0.7 V_{DDIOx}$	-	V_{DDIOx}	V
V_{LSEL}	OSC32_IN input pin low level voltage	-	V_{SS}	-	$0.3 V_{DDIOx}$	
$t_{w(LSEH)}$ $t_{w(LSEL)}$	OSC32_IN high or low time	-	250	-	-	ns

1. Guaranteed by design.

Figure 21. Low-speed external clock source AC timing diagram



MS19215V2

High-speed external clock generated from a crystal/ceramic resonator

The high-speed external (HSE) clock can be supplied with a 4 to 48 MHz crystal/ceramic resonator oscillator. All the information given in this paragraph are based on design simulation results obtained with typical external components specified in [Table 48](#). In the application, the resonator and the load capacitors have to be placed as close as possible to the oscillator pins in order to minimize output distortion and startup stabilization time. Refer to the crystal resonator manufacturer for more details on the resonator characteristics (frequency, package, accuracy).

Table 48. HSE oscillator characteristics⁽¹⁾

Symbol	Parameter	Conditions ⁽²⁾	Min	Typ	Max	Unit
f_{OSC_IN}	Oscillator frequency	-	4	8	48	MHz
R_F	Feedback resistor	-	-	200	-	k Ω
$I_{DD(HSE)}$	HSE current consumption	During startup ⁽³⁾	-	-	5.5	mA
		$V_{DD} = 3\text{ V}$, $R_m = 30\ \Omega$, $CL = 10\text{ pF}@8\text{ MHz}$	-	0.44	-	
		$V_{DD} = 3\text{ V}$, $R_m = 45\ \Omega$, $CL = 10\text{ pF}@8\text{ MHz}$	-	0.45	-	
		$V_{DD} = 3\text{ V}$, $R_m = 30\ \Omega$, $CL = 5\text{ pF}@48\text{ MHz}$	-	0.68	-	
		$V_{DD} = 3\text{ V}$, $R_m = 30\ \Omega$, $CL = 10\text{ pF}@48\text{ MHz}$	-	0.94	-	
		$V_{DD} = 3\text{ V}$, $R_m = 30\ \Omega$, $CL = 20\text{ pF}@48\text{ MHz}$	-	1.77	-	
G_m	Maximum critical crystal transconductance	Startup	-	-	1.5	mA/V
$t_{SU(HSE)}^{(4)}$	Startup time	V_{DD} is stabilized	-	2	-	ms

1. Guaranteed by design.
2. Resonator characteristics given by the crystal/ceramic resonator manufacturer.
3. This consumption level occurs during the first 2/3 of the $t_{SU(HSE)}$ startup time
4. $t_{SU(HSE)}$ is the startup time measured from the moment it is enabled (by software) to a stabilized 8 MHz oscillation is reached. This value is measured for a standard crystal resonator and it can vary significantly with the crystal manufacturer

For C_{L1} and C_{L2} , it is recommended to use high-quality external ceramic capacitors in the 5 pF to 20 pF range (typ.), designed for high-frequency applications, and selected to match the requirements of the crystal or resonator (see [Figure 22](#)). C_{L1} and C_{L2} are usually the same size. The crystal manufacturer typically specifies a load capacitance which is the series combination of C_{L1} and C_{L2} . PCB and MCU pin capacitance must be included (10 pF can be used as a rough estimate of the combined pin and board capacitance) when sizing C_{L1} and C_{L2} .

Note: For information on selecting the crystal, refer to the application note AN2867 “Oscillator design guide for ST microcontrollers” available from the ST website www.st.com.

Figure 22. Typical application with an 8 MHz crystal



1. R_{EXT} value depends on the crystal characteristics.

Low-speed external clock generated from a crystal resonator

The low-speed external (LSE) clock can be supplied with a 32.768 kHz crystal resonator oscillator. All the information given in this paragraph are based on design simulation results obtained with typical external components specified in [Table 49](#). In the application, the resonator and the load capacitors have to be placed as close as possible to the oscillator pins in order to minimize output distortion and startup stabilization time. Refer to the crystal resonator manufacturer for more details on the resonator characteristics (frequency, package, accuracy).

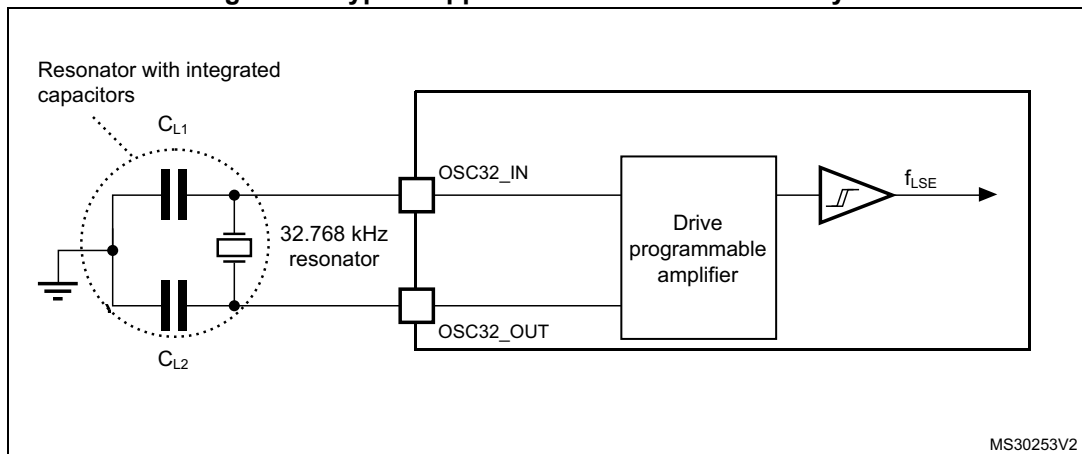
Table 49. LSE oscillator characteristics ($f_{LSE} = 32.768 \text{ kHz}$)⁽¹⁾

Symbol	Parameter	Conditions ⁽²⁾	Min	Typ	Max	Unit
$I_{DD(LSE)}$	LSE current consumption	LSEDRV[1:0] = 00 Low drive capability	-	250	-	nA
		LSEDRV[1:0] = 01 Medium low drive capability	-	315	-	
		LSEDRV[1:0] = 10 Medium high drive capability	-	500	-	
		LSEDRV[1:0] = 11 High drive capability	-	630	-	
$G_{m_{critmax}}$	Maximum critical crystal gm	LSEDRV[1:0] = 00 Low drive capability	-	-	0.5	$\mu A/V$
		LSEDRV[1:0] = 01 Medium low drive capability	-	-	0.75	
		LSEDRV[1:0] = 10 Medium high drive capability	-	-	1.7	
		LSEDRV[1:0] = 11 High drive capability	-	-	2.7	
$t_{SU(LSE)}$ ⁽³⁾	Startup time	V_{DD} is stabilized	-	2	-	s

1. Guaranteed by design.
2. Refer to the note and caution paragraphs below the table, and to the application note AN2867 "Oscillator design guide for ST microcontrollers".
3. $t_{SU(LSE)}$ is the startup time measured from the moment it is enabled (by software) to a stabilized 32.768 kHz oscillation is reached. This value is measured for a standard crystal and it can vary significantly with the crystal manufacturer

Note: For information on selecting the crystal, refer to the application note AN2867 "Oscillator design guide for ST microcontrollers" available from the ST website www.st.com.

Figure 23. Typical application with a 32.768 kHz crystal



Note: An external resistor is not required between OSC32_IN and OSC32_OUT and it is forbidden to add one.

6.3.8 Internal clock source characteristics

The parameters given in [Table 50](#) are derived from tests performed under ambient temperature and supply voltage conditions summarized in [Table 21: General operating conditions](#). The provided curves are characterization results, not tested in production.

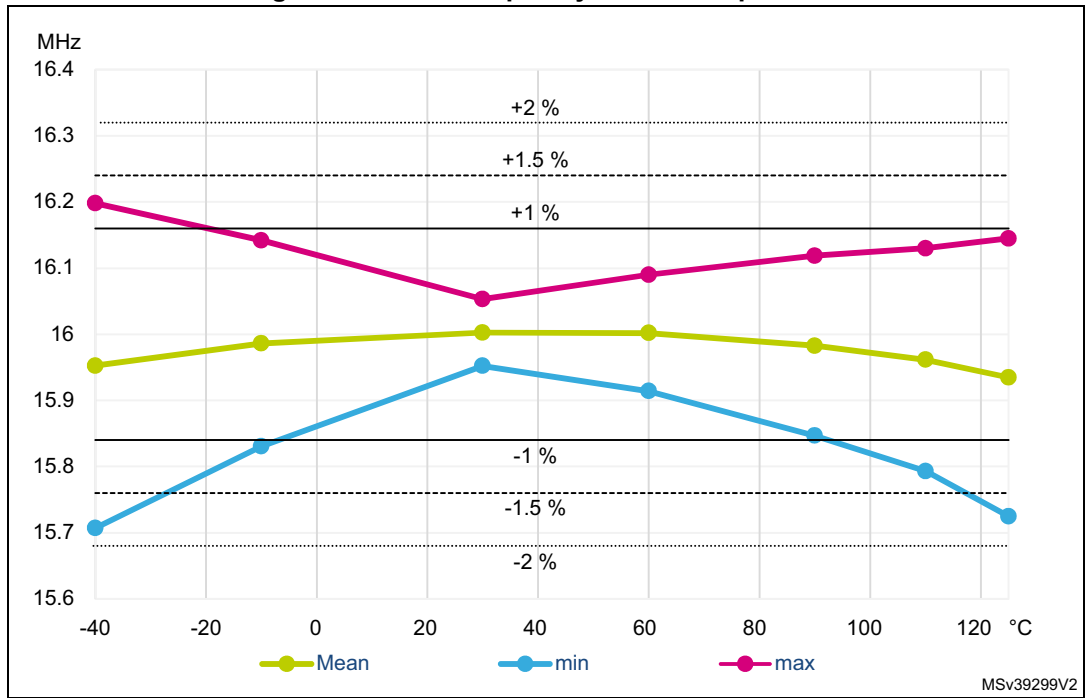
High-speed internal (HSI16) RC oscillator

Table 50. HSI16 oscillator characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
f_{HSI16}	HSI16 Frequency	$V_{\text{DD}}=3.0\text{ V}$, $T_{\text{J}}=30\text{ °C}$	15.88	-	16.08	MHz
TRIM	HSI16 user trimming step	Trimming code is not a multiple of 64	0.2	0.3	0.4	%
		Trimming code is a multiple of 64	-4	-6	-8	
$\text{DuCy}(\text{HSI16})^{(2)}$	Duty Cycle	-	45	-	55	%
$\Delta_{\text{Temp}}(\text{HSI16})$	HSI16 oscillator frequency drift over temperature	$T_{\text{J}}= 0\text{ to }85\text{ °C}$	-1	-	1	%
		$T_{\text{J}}= -40\text{ to }125\text{ °C}$	-2	-	1.5	%
$\Delta_{\text{VDD}}(\text{HSI16})$	HSI16 oscillator frequency drift over V_{DD}	$V_{\text{DD}}=1.71\text{ V to }3.6\text{ V}$	-0.1	-	0.05	%
$t_{\text{su}}(\text{HSI16})^{(2)}$	HSI16 oscillator start-up time	-	-	0.8	1.2	μs
$t_{\text{stab}}(\text{HSI16})^{(2)}$	HSI16 oscillator stabilization time	-	-	3	5	μs
$I_{\text{DD}}(\text{HSI16})^{(2)}$	HSI16 oscillator power consumption	-	-	155	190	μA

1. Guaranteed by characterization results.
2. Guaranteed by design.

Figure 24. HSI16 frequency versus temperature



Multi-speed internal (MSI) RC oscillator

Table 51. MSI oscillator characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min	Typ	Max	Unit	
f _{MSI}	MSI frequency after factory calibration, done at V _{DD} =3 V and T _J =30 °C	MSI mode	Range 0	98.7	100	101.3	kHz
			Range 1	197.4	200	202.6	
			Range 2	394.8	400	405.2	
			Range 3	789.6	800	810.4	
			Range 4	0.987	1	1.013	MHz
			Range 5	1.974	2	2.026	
			Range 6	3.948	4	4.052	
			Range 7	7.896	8	8.104	
			Range 8	15.79	16	16.21	
			Range 9	23.69	24	24.31	
			Range 10	31.58	32	32.42	
		Range 11	47.38	48	48.62		
		PLL mode XTAL= 32.768 kHz	Range 0	-	98.304	-	kHz
			Range 1	-	196.608	-	
			Range 2	-	393.216	-	
			Range 3	-	786.432	-	
			Range 4	-	1.016	-	MHz
			Range 5	-	1.999	-	
			Range 6	-	3.998	-	
			Range 7	-	7.995	-	
			Range 8	-	15.991	-	
			Range 9	-	23.986	-	
Range 10	-		32.014	-			
Range 11	-	48.005	-				
Δ _{TEMP} (MSI) ⁽²⁾	MSI oscillator frequency drift over temperature	MSI mode	T _J = -0 to 85 °C	-3.5	-	3	%
			T _J = -40 to 125 °C	-8	-	6	

Table 51. MSI oscillator characteristics⁽¹⁾ (continued)

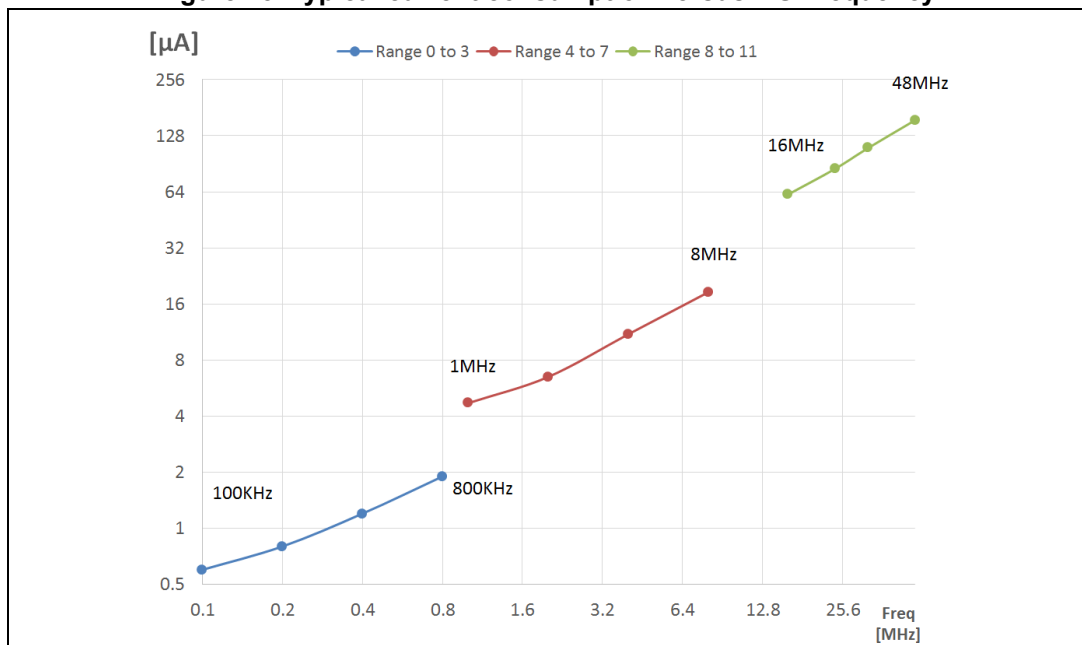
Symbol	Parameter	Conditions		Min	Typ	Max	Unit	
$\Delta V_{DD}(MSI)^{(2)}$	MSI oscillator frequency drift over V_{DD} (reference is 3 V)	MSI mode	Range 0 to 3	$V_{DD}=1.71\text{ V}$ to 3.6 V	-1.2	-	0.5	%
				$V_{DD}=2.4\text{ V}$ to 3.6 V	-0.5	-		
			Range 4 to 7	$V_{DD}=1.71\text{ V}$ to 3.6 V	-2.5	-	0.7	
				$V_{DD}=2.4\text{ V}$ to 3.6 V	-0.8	-		
			Range 8 to 11	$V_{DD}=1.71\text{ V}$ to 3.6 V	-5	-	1	
				$V_{DD}=2.4\text{ V}$ to 3.6 V	-1.6	-		
$\Delta F_{SAMPLING}(MSI)^{(2)(6)}$	Frequency variation in sampling mode ⁽³⁾	MSI mode	$T_J = -40\text{ to }85\text{ }^\circ\text{C}$		-	1	2	%
			$T_J = -40\text{ to }125\text{ }^\circ\text{C}$		-	2	4	
P_USB Jitter(MSI) ⁽⁶⁾	Period jitter for USB clock ⁽⁴⁾	PLL mode Range 11	for next transition	-	-	-	3.458	ns
			for paired transition	-	-	-	3.916	
MT_USB Jitter(MSI) ⁽⁶⁾	Medium term jitter for USB clock ⁽⁵⁾	PLL mode Range 11	for next transition	-	-	-	2	ns
			for paired transition	-	-	-	1	
CC jitter(MSI) ⁽⁶⁾	RMS cycle-to-cycle jitter	PLL mode Range 11		-	-	60	-	ps
P jitter(MSI) ⁽⁶⁾	RMS Period jitter	PLL mode Range 11		-	-	50	-	ps
$t_{SU}(MSI)^{(6)}$	MSI oscillator start-up time	Range 0		-	-	10	20	us
		Range 1		-	-	5	10	
		Range 2		-	-	4	8	
		Range 3		-	-	3	7	
		Range 4 to 7		-	-	3	6	
		Range 8 to 11		-	-	2.5	6	
$t_{STAB}(MSI)^{(6)}$	MSI oscillator stabilization time	PLL mode Range 11	10 % of final frequency	-	-	0.25	0.5	ms
			5 % of final frequency	-	-	0.5	1.25	
			1 % of final frequency	-	-	-	2.5	

Table 51. MSI oscillator characteristics⁽¹⁾ (continued)

Symbol	Parameter	Conditions		Min	Typ	Max	Unit	
I _{DD} (MSI) ⁽⁶⁾	MSI oscillator power consumption	MSI and PLL mode	Range 0	-	-	0.6	1	μA
			Range 1	-	-	0.8	1.2	
			Range 2	-	-	1.2	1.7	
			Range 3	-	-	1.9	2.5	
			Range 4	-	-	4.7	6	
			Range 5	-	-	6.5	9	
			Range 6	-	-	11	15	
			Range 7	-	-	18.5	25	
			Range 8	-	-	62	80	
			Range 9	-	-	85	110	
			Range 10	-	-	110	130	
			Range 11	-	-	155	190	

1. Guaranteed by characterization results.
2. This is a deviation for an individual part once the initial frequency has been measured.
3. Sampling mode means Low-power run/Low-power sleep modes with Temperature sensor disable.
4. Average period of MSI @48 MHz is compared to a real 48 MHz clock over 28 cycles. It includes frequency tolerance + jitter of MSI @48 MHz clock.
5. Only accumulated jitter of MSI @48 MHz is extracted over 28 cycles.
For next transition: min. and max. jitter of 2 consecutive frame of 28 cycles of the MSI @48 MHz, for 1000 captures over 28 cycles.
For paired transitions: min. and max. jitter of 2 consecutive frame of 56 cycles of the MSI @48 MHz, for 1000 captures over 56 cycles.
6. Guaranteed by design.

Figure 25. Typical current consumption versus MSI frequency



High-speed internal 48 MHz (HSI48) RC oscillator

Table 52. HSI48 oscillator characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
f_{HSI48}	HSI48 Frequency	$V_{\text{DD}}=3.0\text{V}$, $T_{\text{J}}=30^{\circ}\text{C}$	-	48	-	MHz
TRIM	HSI48 user trimming step	-	-	0.11 ⁽²⁾	0.18 ⁽²⁾	%
USER TRIM COVERAGE	HSI48 user trimming coverage	± 32 steps	± 3 ⁽³⁾	± 3.5 ⁽³⁾	-	%
DuCy(HSI48)	Duty Cycle	-	45 ⁽²⁾	-	55 ⁽²⁾	%
ACC _{HSI48_REL}	Accuracy of the HSI48 oscillator over temperature (factory calibrated)	$V_{\text{DD}} = 3.0\text{ V to } 3.6\text{ V}$, $T_{\text{J}} = -15\text{ to } 85\text{ }^{\circ}\text{C}$	-	-	± 3 ⁽³⁾	%
		$V_{\text{DD}} = 1.71\text{ V to } 3.6\text{ V}$, $T_{\text{J}} = -40\text{ to } 125\text{ }^{\circ}\text{C}$	-	-	± 4.5 ⁽³⁾	
D _{VDD} (HSI48)	HSI48 oscillator frequency drift with V_{DD}	$V_{\text{DD}} = 3\text{ V to } 3.6\text{ V}$	-	0.025 ⁽³⁾	0.05 ⁽³⁾	%
		$V_{\text{DD}} = 1.71\text{ V to } 3.6\text{ V}$	-	0.05 ⁽³⁾	0.1 ⁽³⁾	
t_{su} (HSI48)	HSI48 oscillator start-up time	-	-	2.5 ⁽²⁾	6 ⁽²⁾	μs
I _{DD} (HSI48)	HSI48 oscillator power consumption	-	-	340 ⁽²⁾	380 ⁽²⁾	μA
N _T jitter	Next transition jitter Accumulated jitter on 28 cycles ⁽⁴⁾	-	-	+/-0.15 ⁽²⁾	-	ns
P _T jitter	Paired transition jitter Accumulated jitter on 56 cycles ⁽⁴⁾	-	-	+/-0.25 ⁽²⁾	-	ns

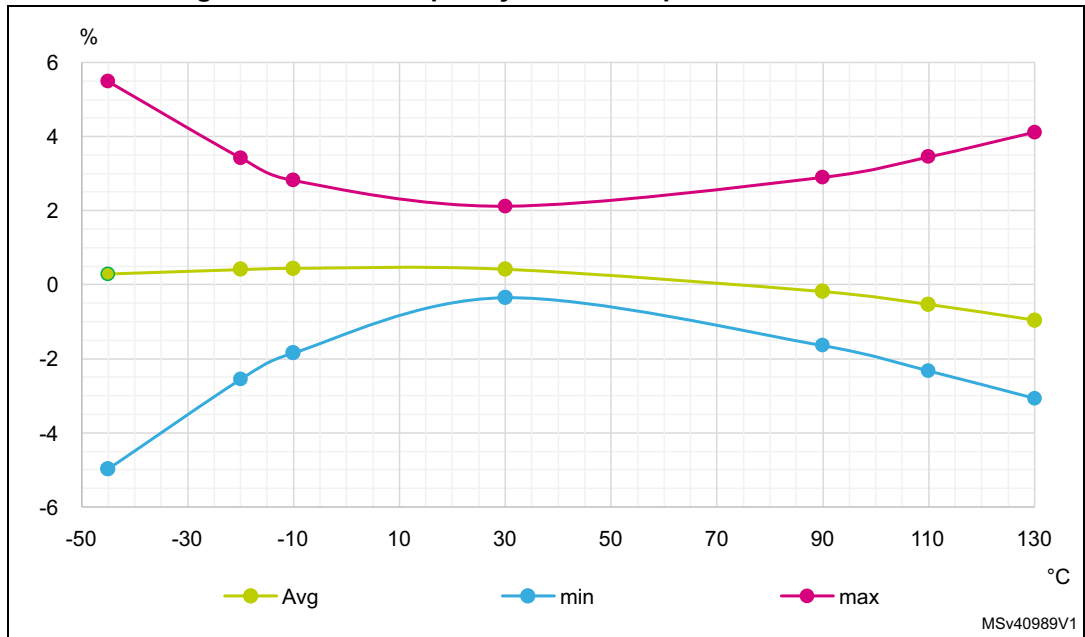
1. $V_{\text{DD}} = 3\text{ V}$, $T_{\text{J}} = -40\text{ to } 125^{\circ}\text{C}$ unless otherwise specified.

2. Guaranteed by design.

3. Guaranteed by characterization results.

4. Jitter measurement are performed without clock source activated in parallel.

Figure 26. HSI48 frequency versus temperature



Low-speed internal (LSI) RC oscillator

Table 53. LSI oscillator characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
f_{LSI}	LSI Frequency	$V_{DD} = 3.0\text{ V}$, $T_J = 30\text{ °C}$	31.04	-	32.96	kHz
		$V_{DD} = 1.71\text{ to }3.6\text{ V}$, $T_J = -40\text{ to }125\text{ °C}$	29.5	-	34	
$t_{SU(LSI)}^{(2)}$	LSI oscillator start-up time	-	-	80	130	μs
$t_{STAB(LSI)}^{(2)}$	LSI oscillator stabilization time	5% of final frequency	-	125	180	μs
$I_{DD(LSI)}^{(2)}$	LSI oscillator power consumption	-	-	110	180	nA

1. Guaranteed by characterization results.
2. Guaranteed by design.

6.3.9 PLL characteristics

The parameters given in [Table 54](#) are derived from tests performed under temperature and V_{DD} supply voltage conditions summarized in [Table 21: General operating conditions](#).

Table 54. PLL, PLLSAI1, PLLSAI2 characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
f_{PLL_IN}	PLL input clock ⁽²⁾	-	2.66	-	16	MHz
	PLL input clock duty cycle	-	45	-	55	%
$f_{PLL_P_OUT}$	PLL multiplier output clock P	Voltage scaling Range 1 Normal mode	2.0645	-	80	MHz
		Voltage scaling Range 1 Boost mode	2.0645	-	120	
		Voltage scaling Range 2	2.0645	-	26	
$f_{PLL_Q_OUT}$	PLL multiplier output clock Q	Voltage scaling Range 1 Normal mode	8	-	80	
		Voltage scaling Range 1 Boost mode	8	-	120	
		Voltage scaling Range 2	8	-	26	
$f_{PLL_R_OUT}$	PLL multiplier output clock R	Voltage scaling Range 1 Normal mode	8	-	80	
		Voltage scaling Range 1 Boost mode	8	-	120	
		Voltage scaling Range 2	8	-	26	
f_{VCO_OUT}	PLL VCO output	Voltage scaling Range 1	64	-	344	
		Voltage scaling Range 2	64	-	128	
t_{LOCK}	PLL lock time	-	-	15	40	μs
Jitter	RMS cycle-to-cycle jitter	System clock 80 MHz	-	40	-	$\pm ps$
	RMS period jitter		-	30	-	
$I_{DD(PLL)}$	PLL power consumption on V_{DD} ⁽¹⁾	VCO freq = 64 MHz	-	150	200	μA
		VCO freq = 96 MHz	-	200	260	
		VCO freq = 192 MHz	-	300	380	
		VCO freq = 344 MHz	-	520	650	

1. Guaranteed by design.
2. Take care of using the appropriate division factor M to obtain the specified PLL input clock values. The M factor is shared between the 3 PLLs.

6.3.10 Flash memory characteristics

Table 55. Flash memory characteristics⁽¹⁾

Symbol	Parameter	Conditions	Typ	Max	Unit
t _{prog}	64-bit programming time	-	81.69	83.35	μs
t _{prog_row}	One row (64 double word) programming time	Normal programming	5.23	5.33	ms
		Fast programming	3.80	3.88	
t _{prog_page}	One page (4 Kbytes) programming time	Normal programming	41.82	42.68	
		Fast programming	30.39	31.01	
t _{ERASE}	Page (4 Kbytes) erase time	-	22.02	24.47	
t _{prog_bank}	One bank (512 Kbytes) programming time	Normal programming	5.35	5.46	
		Fast programming	3.89	3.97	
t _{ME}	Mass erase time (one or two banks)	-	22.13	24.59	ms
I _{DD}	Average consumption from V _{DD}	Write mode	3.4	-	mA
		Erase mode	3.4	-	
	Maximum current (peak)	Write mode	7 (for 6 μs)	-	
		Erase mode	7 (for 67 μs)	-	

1. Guaranteed by design.

Table 56. Flash memory endurance and data retention

Symbol	Parameter	Conditions	Min ⁽¹⁾	Unit
N _{END}	Endurance	T _A = -40 to +105 °C	10	kcycles
t _{RET}	Data retention	1 kcycle ⁽²⁾ at T _A = 85 °C	30	Years
		1 kcycle ⁽²⁾ at T _A = 105 °C	15	
		1 kcycle ⁽²⁾ at T _A = 125 °C	7	
		10 kcycles ⁽²⁾ at T _A = 55 °C	30	
		10 kcycles ⁽²⁾ at T _A = 85 °C	15	
		10 kcycles ⁽²⁾ at T _A = 105 °C	10	

1. Guaranteed by characterization results.

2. Cycling performed over the whole temperature range.

6.3.11 EMC characteristics

Susceptibility tests are performed on a sample basis during device characterization.

Functional EMS (electromagnetic susceptibility)

While a simple application is executed on the device (toggling 2 LEDs through I/O ports), the device is stressed by two electromagnetic events until a failure occurs. The failure is indicated by the LEDs:

- **Electrostatic discharge (ESD)** (positive and negative) is applied to all device pins until a functional disturbance occurs. This test is compliant with the IEC 61000-4-2 standard.
- **FTB**: A Burst of Fast Transient voltage (positive and negative) is applied to V_{DD} and V_{SS} through a 100 pF capacitor, until a functional disturbance occurs. This test is compliant with the IEC 61000-4-4 standard.

A device reset allows normal operations to be resumed.

The test results are given in [Table 57](#). They are based on the EMS levels and classes defined in application note AN1709.

Table 57. EMS characteristics

Symbol	Parameter	Conditions	Level/Class
V_{FESD}	Voltage limits to be applied on any I/O pin to induce a functional disturbance	$V_{DD} = 3.3\text{ V}$, $T_A = +25\text{ °C}$, $f_{HCLK} = 120\text{ MHz}$, conforming to IEC 61000-4-2	3B
V_{EFTB}	Fast transient voltage burst limits to be applied through 100 pF on V_{DD} and V_{SS} pins to induce a functional disturbance	$V_{DD} = 3.3\text{ V}$, $T_A = +25\text{ °C}$, $f_{HCLK} = 120\text{ MHz}$, conforming to IEC 61000-4-4	5A

Designing hardened software to avoid noise problems

EMC characterization and optimization are performed at component level with a typical application environment and simplified MCU software. It should be noted that good EMC performance is highly dependent on the user application and the software in particular.

Therefore it is recommended that the user applies EMC software optimization and prequalification tests in relation with the EMC level requested for his application.

Software recommendations

The software flowchart must include the management of runaway conditions such as:

- Corrupted program counter
- Unexpected reset
- Critical Data corruption (control registers...)

Prequalification trials

Most of the common failures (unexpected reset and program counter corruption) can be reproduced by manually forcing a low state on the NRST pin or the Oscillator pins for 1 second.

To complete these trials, ESD stress can be applied directly on the device, over the range of specification values. When unexpected behavior is detected, the software can be hardened to prevent unrecoverable errors occurring (see application note AN1015).

Electromagnetic Interference (EMI)

The electromagnetic field emitted by the device are monitored while a simple application is executed (toggling 2 LEDs through the I/O ports). This emission test is compliant with IEC 61967-2 standard which specifies the test board and the pin loading.

Table 58. EMI characteristics

Symbol	Parameter	Conditions	Monitored frequency band	Max vs. [f _{HSE} /f _{HCLK}]	Unit
				8 MHz/120 MHz	
S _{EMI}	Peak level	V _{dd} =3.6 V, T _A =25°C, LQFP144 package, conforming to IEC61967-2	0.1 MHz to 30 MHz	16	dBμV
			30 MHz to 130 MHz	16	
			130 MHz to 1 GHz	19	
			1 GHz to 2 GHz	12	
			EMI Level	3.5	-

6.3.12 Electrical sensitivity characteristics

Based on three different tests (ESD, LU) using specific measurement methods, the device is stressed in order to determine its performance in terms of electrical sensitivity.

Electrostatic discharge (ESD)

Electrostatic discharges (a positive then a negative pulse separated by 1 second) are applied to the pins of each sample according to each pin combination. The sample size depends on the number of supply pins in the device (3 parts × (n+1) supply pins). This test conforms to the ANSI/JEDEC standard.

Table 59. ESD absolute maximum ratings

Symbol	Ratings	Conditions	Packages	Class	Maximum value ⁽¹⁾	Unit
V _{ESD(HBM)}	Electrostatic discharge voltage (human body model)	T _A = +25 °C, conforming to ANSI/ESDA/JEDEC JS-001	All	2	2000	V
V _{ESD(CDM)}	Electrostatic discharge voltage (charge device model)	T _A = +25 °C, conforming to ANSI/ESDA/JEDEC JS-002	LQFP64	C2a	500	V
			UFBGA169, UFBGA132, LQFP144, WLCSP100	C1	250	
			LQFP48, UFQFPN48	-	TBD	

1. Guaranteed by characterization results.

Static latch-up

Two complementary static tests are required on six parts to assess the latch-up performance:

- A supply overvoltage is applied to each power supply pin.
- A current injection is applied to each input, output and configurable I/O pin.

These tests are compliant with EIA/JESD78 IC latch-up standard.

Table 60. Electrical sensitivities

Symbol	Parameter	Conditions	Class
LU	Static latch-up class	T _A = +105 °C conforming to JESD78	II level A

6.3.13 I/O current injection characteristics

As a general rule, current injection to the I/O pins, due to external voltage below V_{SS} or above V_{DDIOx} (for standard, 3.3 V-capable I/O pins) should be avoided during normal product operation. However, in order to give an indication of the robustness of the microcontroller in cases when abnormal injection accidentally happens, susceptibility tests are performed on a sample basis during device characterization.

Functional susceptibility to I/O current injection

While a simple application is executed on the device, the device is stressed by injecting current into the I/O pins programmed in floating input mode. While current is injected into the I/O pin, one at a time, the device is checked for functional failures.

The failure is indicated by an out of range parameter: ADC error above a certain limit (higher than 5 LSB TUE), out of conventional limits of induced leakage current on adjacent pins (out of the $-5 \mu\text{A}/+0 \mu\text{A}$ range) or other functional failure (for example reset occurrence or oscillator frequency deviation).

The characterization results are given in [Table 61](#).

Negative induced leakage current is caused by negative injection and positive induced leakage current is caused by positive injection.

Table 61. I/O current injection susceptibility

Symbol	Description	Functional susceptibility		Unit
		Negative injection	Positive injection	
$I_{\text{INJ}}^{(1)}$	Injected current on all pins except TT_x	-5	NA	mA
	Injected current on pins PD7, OPAMP1_VINM, OPAMP2_VINM	0	NA	
	Injected current on TT_ax pins	-5	0	

1. Guaranteed by characterization.

6.3.14 I/O port characteristics

General input/output characteristics

Unless otherwise specified, the parameters given in [Table 62](#) are derived from tests performed under the conditions summarized in [Table 21: General operating conditions](#). All I/Os are designed as CMOS- and TTL-compliant.

Table 62. I/O static characteristics

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$V_{\text{IL}}^{(1)}$	I/O input low level voltage	$1.62 \text{ V} < V_{\text{DDIOx}} < 3.6 \text{ V}$	-	-	$0.3 \times V_{\text{DDIOx}}$ (2)	V
	I/O input low level voltage	$1.62 \text{ V} < V_{\text{DDIOx}} < 3.6 \text{ V}$	-	-	$0.39 \times V_{\text{DDIOx}} - 0.06$ (3)	
	I/O input low level voltage	$1.08 \text{ V} < V_{\text{DDIOx}} < 1.62 \text{ V}$	-	-	$0.43 \times V_{\text{DDIOx}} - 0.1$ (3)	

Table 62. I/O static characteristics (continued)

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$V_{IH}^{(1)}$	I/O input high level voltage	$1.62\text{ V} < V_{DDIOx} < 3.6\text{ V}$	$0.7 \times V_{DDIOx}^{(2)}$	-	-	V
	I/O input high level voltage	$1.62\text{ V} < V_{DDIOx} < 3.6\text{ V}$	$0.49 \times V_{DDIOx} + 0.26^{(3)}$	-	-	
	I/O input high level voltage	$1.08\text{ V} < V_{DDIOx} < 1.62\text{ V}$	$0.61 \times V_{DDIOx} + 0.05^{(3)}$	-	-	
$V_{hys}^{(3)}$	TT_xx, FT_xxx and NRST I/O input hysteresis	$1.62\text{ V} < V_{DDIOx} < 3.6\text{ V}$	-	200	-	mV
	FT_sx	$1.08\text{ V} < V_{DDIOx} < 1.62\text{ V}$	-	150	-	
I_{lkg}	FT_xx input leakage current ⁽³⁾	$V_{IN} \leq \text{Max}(V_{DDXXX})^{(4)}$	-	-	+/- 100	nA
		$\text{Max}(V_{DDXXX}) \leq V_{IN} \leq \text{Max}(V_{DDXXX}) + 1\text{ V}^{(4)(5)}$	-	-	650 ⁽³⁾⁽⁶⁾	
		$\text{Max}(V_{DDXXX}) + 1\text{ V} < V_{IN} \leq 5.5\text{ V}^{(3)(5)}$	-	-	200 ⁽⁶⁾	
	FT_lu, FT_u, PB2 and PC3 IO	$V_{IN} \leq \text{Max}(V_{DDXXX})^{(4)}$	-	-	+/- 150	
		$\text{Max}(V_{DDXXX}) \leq V_{IN} \leq \text{Max}(V_{DDXXX}) + 1\text{ V}^{(4)}$	-	-	2500 ⁽³⁾⁽⁷⁾	
		$\text{Max}(V_{DDXXX}) + 1\text{ V} < V_{IN} \leq 5.5\text{ V}^{(4)(5)(7)}$	-	-	250 ⁽⁷⁾	
	TT_xx input leakage current	$V_{IN} \leq \text{Max}(V_{DDXXX})^{(6)}$	-	-	+/-150	
$\text{Max}(V_{DDXXX}) \leq V_{IN} < 3.6\text{ V}^{(6)}$		-	-	2000 ⁽³⁾		
OPAMPx_VINM (x=1,2) dedicated input leakage current	-	-	-	(8)		
R_{PU}	Weak pull-up equivalent resistor ⁽⁹⁾	$V_{IN} = V_{SS}$	25	40	55	kΩ
R_{PD}	Weak pull-down equivalent resistor ⁽⁹⁾	$V_{IN} = V_{DDIOx}$	25	40	55	kΩ
C_{IO}	I/O pin capacitance	-	-	5	-	pF

1. Refer to [Figure 27: I/O input characteristics](#).

2. Tested in production.

3. Guaranteed by design.

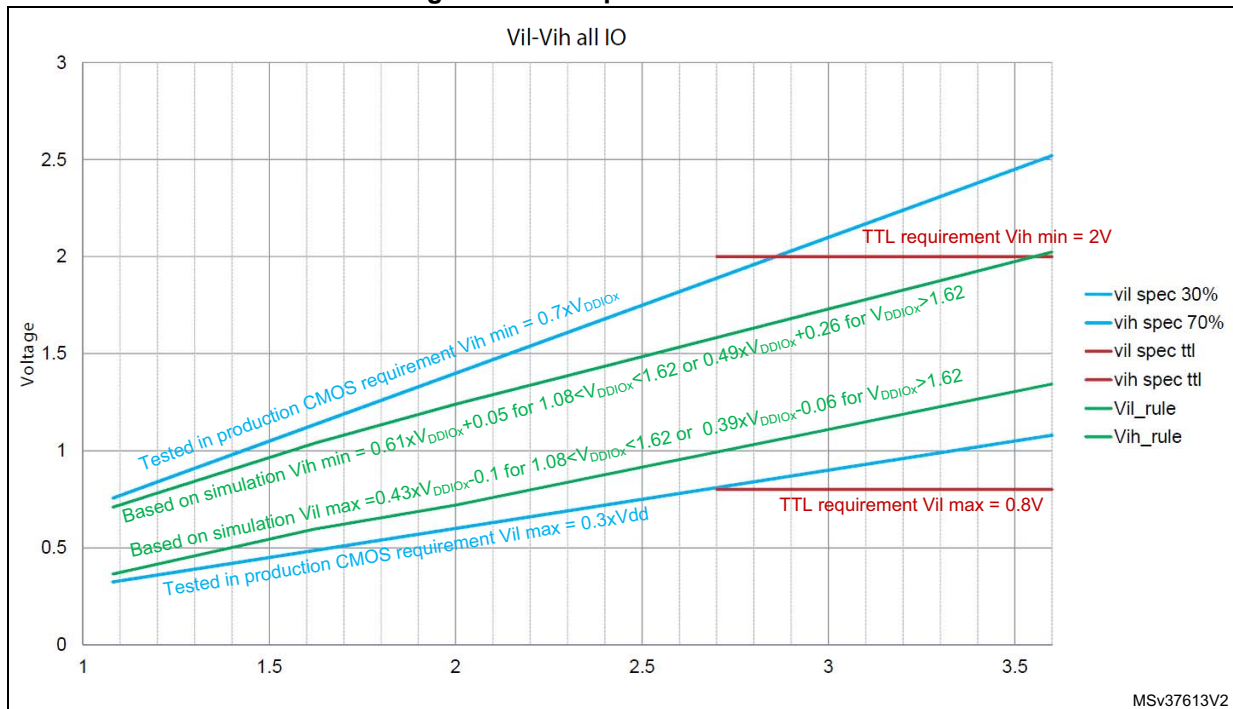
4. $\text{Max}(V_{DDXXX})$ is the maximum value of all the I/O supplies. Refer to [Table: Legend/Abbreviations used in the pinout table](#).

5. All TX_xx IO except FT_lu, FT_u, PB2 and PC3.

- This value represents the pad leakage of the IO itself. The total product pad leakage is provided by this formula:
 $I_{Total_leak_max} = 10 \mu A + [\text{number of IOs where } V_{IN} \text{ is applied on the pad}] \times I_{lk}(Max)$.
- To sustain a voltage higher than $MIN(V_{DD}, V_{DDA}, V_{DDUSB}, V_{DDIO2}) + 0.3 V$, the internal Pull-up and Pull-Down resistors must be disabled.
- Refer to I_{bias} in [Table 78: OPAMP characteristics](#) for the values of the OPAMP dedicated input leakage current.
- Pull-up and pull-down resistors are designed with a true resistance in series with a switchable PMOS/NMOS. This PMOS/NMOS contribution to the series resistance is minimal (~10% order).

All I/Os are CMOS- and TTL-compliant (no software configuration required). Their characteristics cover more than the strict CMOS-technology or TTL parameters. The coverage of these requirements is shown in [Figure 27](#) for standard I/Os, and in [Figure 27](#) for 5 V tolerant I/Os.

Figure 27. I/O input characteristics



Output driving current

The GPIOs (general purpose input/outputs) can sink or source up to ± 8 mA, and sink or source up to ± 20 mA (with a relaxed V_{OL}/V_{OH}).

In the user application, the number of I/O pins which can drive current must be limited to respect the absolute maximum rating specified in [Section 6.2](#):

- The sum of the currents sourced by all the I/Os on V_{DDIOx} , plus the maximum consumption of the MCU sourced on V_{DD} , cannot exceed the absolute maximum rating ΣI_{VDD} (see [Table 18: Voltage characteristics](#)).
- The sum of the currents sunk by all the I/Os on V_{SS} , plus the maximum consumption of the MCU sunk on V_{SS} , cannot exceed the absolute maximum rating ΣI_{VSS} (see [Table 18: Voltage characteristics](#)).

Output voltage levels

Unless otherwise specified, the parameters given in the table below are derived from tests performed under the ambient temperature and supply voltage conditions summarized in [Table 21: General operating conditions](#). All I/Os are CMOS- and TTL-compliant (FT OR TT unless otherwise specified).

Table 63. Output voltage characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min	Max	Unit
V_{OL}	Output low level voltage for an I/O pin	CMOS port ⁽²⁾ $ I_{IO} = 8 \text{ mA}$ $V_{DDIOx} \geq 2.7 \text{ V}$	-	0.4	V
V_{OH}	Output high level voltage for an I/O pin		$V_{DDIOx}-0.4$	-	
$V_{OL}^{(3)}$	Output low level voltage for an I/O pin	TTL port ⁽²⁾ $ I_{IO} = 8 \text{ mA}$ $V_{DDIOx} \geq 2.7 \text{ V}$	-	0.4	
$V_{OH}^{(3)}$	Output high level voltage for an I/O pin		2.4	-	
$V_{OL}^{(3)}$	Output low level voltage for an I/O pin	$ I_{IO} = 20 \text{ mA}$ $V_{DDIOx} \geq 2.7 \text{ V}$	-	1.3	
$V_{OH}^{(3)}$	Output high level voltage for an I/O pin		$V_{DDIOx}-1.3$	-	
$V_{OL}^{(3)}$	Output low level voltage for an I/O pin	$ I_{IO} = 4 \text{ mA}$ $V_{DDIOx} \geq 1.62 \text{ V}$	-	0.45	
$V_{OH}^{(3)}$	Output high level voltage for an I/O pin		$V_{DDIOx}-0.45$	-	
$V_{OL}^{(3)}$	Output low level voltage for an I/O pin	$ I_{IO} = 2 \text{ mA}$ $1.62 \text{ V} \geq V_{DDIOx} \geq 1.08 \text{ V}$	-	$0.35 \times V_{DDIOx}$	
$V_{OH}^{(3)}$	Output high level voltage for an I/O pin		$0.65 \times V_{DDIOx}$	-	
$V_{OLFM+}^{(3)}$	Output low level voltage for an FT I/O pin in FM+ mode (FT I/O with "f" option)	$ I_{IO} = 20 \text{ mA}$ $V_{DDIOx} \geq 2.7 \text{ V}$	-	0.4	
		$ I_{IO} = 10 \text{ mA}$ $V_{DDIOx} \geq 1.62 \text{ V}$	-	0.4	
		$ I_{IO} = 2 \text{ mA}$ $1.62 \text{ V} \geq V_{DDIOx} \geq 1.08 \text{ V}$	-	0.4	

1. The I_{IO} current sourced or sunk by the device must always respect the absolute maximum rating specified in [Table 18: Voltage characteristics](#), and the sum of the currents sourced or sunk by all the I/Os (I/O ports and control pins) must always respect the absolute maximum ratings ΣI_{IO} .
2. TTL and CMOS outputs are compatible with JEDEC standards JESD36 and JESD52.
3. Guaranteed by design.

Input/output AC characteristics

The definition and values of input/output AC characteristics are given in [Figure 28](#) and [Table 64](#), respectively.

Unless otherwise specified, the parameters given are derived from tests performed under the ambient temperature and supply voltage conditions summarized in [Table 21: General operating conditions](#).

Table 64. I/O AC characteristics⁽¹⁾⁽²⁾

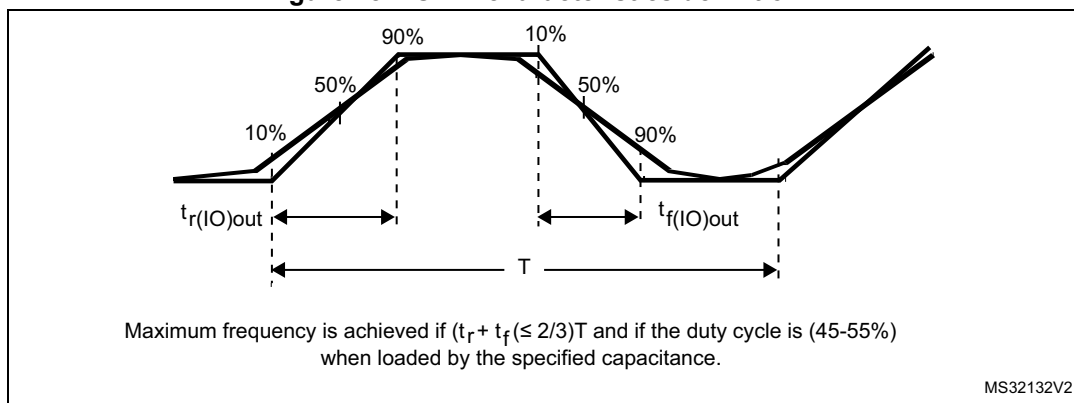
Speed	Symbol	Parameter	Conditions	Min	Max	Unit
00	Fmax	Maximum frequency	C=50 pF, 2.7 V ≤ V _{DDIOx} ≤ 3.6 V	-	5	MHz
			C=50 pF, 1.62 V ≤ V _{DDIOx} ≤ 2.7 V	-	1	
			C=50 pF, 1.08 V ≤ V _{DDIOx} ≤ 1.62 V	-	0.1	
			C=10 pF, 2.7 V ≤ V _{DDIOx} ≤ 3.6 V	-	10	
			C=10 pF, 1.62 V ≤ V _{DDIOx} ≤ 2.7 V	-	1.5	
			C=10 pF, 1.08 V ≤ V _{DDIOx} ≤ 1.62 V	-	0.1	
	Tr/Tf	Output rise and fall time	C=50 pF, 2.7 V ≤ V _{DDIOx} ≤ 3.6 V	-	25	ns
			C=50 pF, 1.62 V ≤ V _{DDIOx} ≤ 2.7 V	-	52	
			C=50 pF, 1.08 V ≤ V _{DDIOx} ≤ 1.62 V	-	140	
			C=10 pF, 2.7 V ≤ V _{DDIOx} ≤ 3.6 V	-	17	
			C=10 pF, 1.62 V ≤ V _{DDIOx} ≤ 2.7 V	-	37	
			C=10 pF, 1.08 V ≤ V _{DDIOx} ≤ 1.62 V	-	110	
01	Fmax	Maximum frequency	C=50 pF, 2.7 V ≤ V _{DDIOx} ≤ 3.6 V	-	25	MHz
			C=50 pF, 1.62 V ≤ V _{DDIOx} ≤ 2.7 V	-	10	
			C=50 pF, 1.08 V ≤ V _{DDIOx} ≤ 1.62 V	-	1	
			C=10 pF, 2.7 V ≤ V _{DDIOx} ≤ 3.6 V	-	50	
			C=10 pF, 1.62 V ≤ V _{DDIOx} ≤ 2.7 V	-	15	
			C=10 pF, 1.08 V ≤ V _{DDIOx} ≤ 1.62 V	-	1	
	Tr/Tf	Output rise and fall time	C=50 pF, 2.7 V ≤ V _{DDIOx} ≤ 3.6 V	-	9	ns
			C=50 pF, 1.62 V ≤ V _{DDIOx} ≤ 2.7 V	-	16	
			C=50 pF, 1.08 V ≤ V _{DDIOx} ≤ 1.62 V	-	40	
			C=10 pF, 2.7 V ≤ V _{DDIOx} ≤ 3.6 V	-	4.5	
			C=10 pF, 1.62 V ≤ V _{DDIOx} ≤ 2.7 V	-	9	
			C=10 pF, 1.08 V ≤ V _{DDIOx} ≤ 1.62 V	-	21	

Table 64. I/O AC characteristics⁽¹⁾⁽²⁾ (continued)

Speed	Symbol	Parameter	Conditions	Min	Max	Unit
10	Fmax	Maximum frequency	C=50 pF, 2.7 V ≤ V _{DDIOx} ≤ 3.6 V	-	50	MHz
			C=50 pF, 1.62 V ≤ V _{DDIOx} ≤ 2.7 V	-	25	
			C=50 pF, 1.08 V ≤ V _{DDIOx} ≤ 1.62 V	-	5	
			C=10 pF, 2.7 V ≤ V _{DDIOx} ≤ 3.6 V	-	100	
			C=10 pF, 1.62 V ≤ V _{DDIOx} ≤ 2.7 V	-	37.5	
			C=10 pF, 1.08 V ≤ V _{DDIOx} ≤ 1.62 V	-	5	
	Tr/Tf	Output rise and fall time	C=50 pF, 2.7 V ≤ V _{DDIOx} ≤ 3.6 V	-	5.8	ns
			C=50 pF, 1.62 V ≤ V _{DDIOx} ≤ 2.7 V	-	11	
			C=50 pF, 1.08 V ≤ V _{DDIOx} ≤ 1.62 V	-	28	
			C=10 pF, 2.7 V ≤ V _{DDIOx} ≤ 3.6 V	-	2.5	
			C=10 pF, 1.62 V ≤ V _{DDIOx} ≤ 2.7 V	-	5	
			C=10 pF, 1.08 V ≤ V _{DDIOx} ≤ 1.62 V	-	12	
11	Fmax	Maximum frequency	C=30 pF, 2.7 V ≤ V _{DDIOx} ≤ 3.6 V	-	120	MHz
			C=30 pF, 1.62 V ≤ V _{DDIOx} ≤ 2.7 V	-	50	
			C=30 pF, 1.08 V ≤ V _{DDIOx} ≤ 1.62 V	-	10	
			C=10 pF, 2.7 V ≤ V _{DDIOx} ≤ 3.6 V	-	180 ⁽³⁾	
			C=10 pF, 1.62 V ≤ V _{DDIOx} ≤ 2.7 V	-	75	
			C=10 pF, 1.08 V ≤ V _{DDIOx} ≤ 1.62 V	-	10	
	Tr/Tf	Output rise and fall time	C=30 pF, 2.7 V ≤ V _{DDIOx} ≤ 3.6 V	-	3.3	ns
			C=30 pF, 1.62 V ≤ V _{DDIOx} ≤ 2.7 V	-	6	
			C=30 pF, 1.08 V ≤ V _{DDIOx} ≤ 1.62 V	-	16	
Fm+	Fmax	Maximum frequency	C=50 pF, 1.6 V ≤ V _{DDIOx} ≤ 3.6 V	-	1	MHz
	Tf	Output fall time ⁽⁴⁾		-	5	ns

1. The I/O speed is configured using the OSPEEDRy[1:0] bits. The Fm+ mode is configured in the SYSCFG_CFGR1 register. Refer to the RM0432 reference manual for a description of GPIO Port configuration register.
2. Guaranteed by design.
3. This value represents the I/O capability but the maximum system frequency is limited to 120 MHz.
4. The fall time is defined between 70% and 30% of the output waveform accordingly to I²C specification.

Figure 28. I/O AC characteristics definition⁽¹⁾



1. Refer to [Table 64: I/O AC characteristics](#).

6.3.15 NRST pin characteristics

The NRST pin input driver uses the CMOS technology. It is connected to a permanent pull-up resistor, R_{PU} .

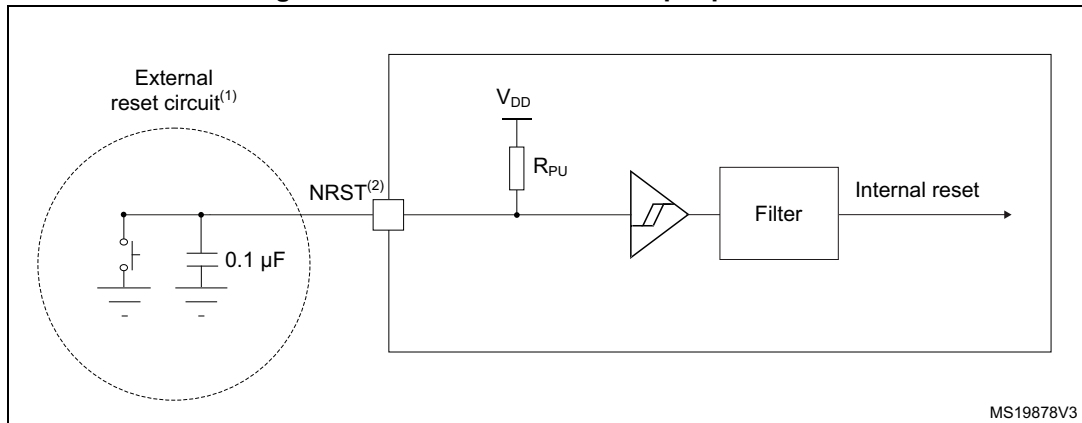
Unless otherwise specified, the parameters given in the table below are derived from tests performed under the ambient temperature and supply voltage conditions summarized in [Table 21: General operating conditions](#).

Table 65. NRST pin characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$V_{IL(NRST)}$	NRST input low level voltage	-	-	-	$0.3 \times V_{DDIOX}$	V
$V_{IH(NRST)}$	NRST input high level voltage	-	$0.7 \times V_{DDIOX}$	-	-	
$V_{hys(NRST)}$	NRST Schmitt trigger voltage hysteresis	-	-	200	-	mV
R_{PU}	Weak pull-up equivalent resistor ⁽²⁾	$V_{IN} = V_{SS}$	25	40	55	k Ω
$V_{F(NRST)}$	NRST input filtered pulse	-	-	-	70	ns
$V_{NF(NRST)}$	NRST input not filtered pulse	$1.71 \text{ V} \leq V_{DD} \leq 3.6 \text{ V}$	350	-	-	ns

- Guaranteed by design.
- The pull-up is designed with a true resistance in series with a switchable PMOS. This PMOS contribution to the series resistance is minimal (~10% order).

Figure 29. Recommended NRST pin protection



1. The reset network protects the device against parasitic resets.
2. The user must ensure that the level on the NRST pin can go below the $V_{IL(NRST)}$ max level specified in [Table 65: NRST pin characteristics](#). Otherwise the reset is not taken into account by the device.
3. The external capacitor on NRST must be placed as close as possible to the device.

6.3.16 Extended interrupt and event controller input (EXTI) characteristics

The pulse on the interrupt input must have a minimal length in order to guarantee that it is detected by the event controller.

Table 66. EXTI input characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
PLEC	Pulse length to event controller	-	20	-	-	ns

1. Guaranteed by design.

6.3.17 Analog switches booster

Table 67. Analog switches booster characteristics⁽¹⁾

Symbol	Parameter	Min	Typ	Max	Unit
V_{DD}	Supply voltage	1.71	-	3.6	V
$t_{SU(BOOST)}$	Booster startup time	-	-	240	µs
$I_{DD(BOOST)}$	Booster consumption for $1.71\text{ V} \leq V_{DD} \leq 2.0\text{ V}$	-	-	250	µA
	Booster consumption for $2.0\text{ V} \leq V_{DD} \leq 2.7\text{ V}$	-	-	500	
	Booster consumption for $2.7\text{ V} \leq V_{DD} \leq 3.6\text{ V}$	-	-	900	

1. Guaranteed by design.

6.3.18 Analog-to-digital converter characteristics

Unless otherwise specified, the parameters given in [Table 68](#) are preliminary values derived from tests performed under ambient temperature, f_{PCLK} frequency and V_{DDA} supply voltage conditions summarized in [Table 21: General operating conditions](#).

Note: It is recommended to perform a calibration after each power-up.

Table 68. ADC characteristics^{(1) (2)}

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
V_{DDA}	Analog supply voltage	-	1.62	-	3.6	V
V_{REF+}	Positive reference voltage	$V_{DDA} \geq 2\text{ V}$	2	-	V_{DDA}	V
		$V_{DDA} < 2\text{ V}$	V_{DDA}			V
V_{REF-}	Negative reference voltage	-	V_{SSA}			V
f_{ADC}	ADC clock frequency	Range 1	-	-	80	MHz
		Range 2	-	-	26	
f_s	Sampling rate for FAST channels	Resolution = 12 bits	-	-	5.33	Mpsps
		Resolution = 10 bits	-	-	6.15	
		Resolution = 8 bits	-	-	7.27	
		Resolution = 6 bits	-	-	8.88	
	Sampling rate for SLOW channels	Resolution = 12 bits	-	-	4.21	
		Resolution = 10 bits	-	-	4.71	
		Resolution = 8 bits	-	-	5.33	
		Resolution = 6 bits	-	-	6.15	
f_{TRIG}	External trigger frequency	$f_{ADC} = 80\text{ MHz}$ Resolution = 12 bits	-	-	5.33	MHz
		Resolution = 12 bits	-	-	15	$1/f_{ADC}$
$V_{AIN}^{(3)}$	Conversion voltage range(2)	-	0	-	V_{REF+}	V
R_{AIN}	External input impedance	-	-	-	50	k Ω
C_{ADC}	Internal sample and hold capacitor	-	-	5	-	pF
t_{STAB}	Power-up time	-	1			conversion cycle
t_{CAL}	Calibration time	$f_{ADC} = 80\text{ MHz}$	1.45			μs
		-	116			$1/f_{ADC}$

Table 68. ADC characteristics^{(1) (2)} (continued)

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
t_{LATR}	Trigger conversion latency Regular and injected channels without conversion abort	CKMODE = 00	1.5	2	2.5	$1/f_{ADC}$
		CKMODE = 01	-	-	2.0	
		CKMODE = 10	-	-	2.25	
		CKMODE = 11	-	-	2.125	
$t_{LATRINJ}$	Trigger conversion latency Injected channels aborting a regular conversion	CKMODE = 00	2.5	3	3.5	$1/f_{ADC}$
		CKMODE = 01	-	-	3.0	
		CKMODE = 10	-	-	3.25	
		CKMODE = 11	-	-	3.125	
t_s	Sampling time	$f_{ADC} = 80$ MHz	0.03125	-	8.00625	μs
		-	2.5	-	640.5	$1/f_{ADC}$
$t_{ADCVREG_STUP}$	ADC voltage regulator start-up time	-	-	-	20	μs
t_{CONV}	Total conversion time (including sampling time)	$f_{ADC} = 80$ MHz Resolution = 12 bits	0.1875	-	8.1625	μs
		Resolution = 12 bits	ts + 12.5 cycles for successive approximation = 15 to 653			$1/f_{ADC}$
$I_{DDA(ADC)}$	ADC consumption from the V_{DDA} supply	fs = 5 Msps	-	730	830	μA
		fs = 1 Msps	-	160	220	
		fs = 10 ksps	-	16	50	
$I_{DDV_S(ADC)}$	ADC consumption from the V_{REF+} single ended mode	fs = 5 Msps	-	130	160	μA
		fs = 1 Msps	-	30	40	
		fs = 10 ksps	-	0.6	2	
$I_{DDV_D(ADC)}$	ADC consumption from the V_{REF+} differential mode	fs = 5 Msps	-	260	310	μA
		fs = 1 Msps	-	60	70	
		fs = 10 ksps	-	1.3	3	

- Guaranteed by design
- The I/O analog switch voltage booster is enable when $V_{DDA} < 2.4$ V (BOOSTEN = 1 in the SYSCFG_CFGR1 when $V_{DDA} < 2.4$ V). It is disable when $V_{DDA} \geq 2.4$ V.
- V_{REF+} can be internally connected to V_{DDA} and V_{REF-} can be internally connected to V_{SSA} , depending on the package.
Refer to [Section 4: Pinouts and pin description](#) for further details.

The maximum value of R_{AIN} can be found in [Table 69: Maximum ADC RAIN](#).

Table 69. Maximum ADC $R_{AIN}^{(1)(2)}$

Resolution	Sampling cycle @80 MHz	Sampling time [ns] @80 MHz	R_{AIN} max (Ω)	
			Fast channels ⁽³⁾	Slow channels ⁽⁴⁾
12 bits	2.5	31.25	100	N/A
	6.5	81.25	330	100
	12.5	156.25	680	470
	24.5	306.25	1500	1200
	47.5	593.75	2200	1800
	92.5	1156.25	4700	3900
	247.5	3093.75	12000	10000
	640.5	8006.75	39000	33000
10 bits	2.5	31.25	120	N/A
	6.5	81.25	390	180
	12.5	156.25	820	560
	24.5	306.25	1500	1200
	47.5	593.75	2200	1800
	92.5	1156.25	5600	4700
	247.5	3093.75	12000	10000
	640.5	8006.75	47000	39000
8 bits	2.5	31.25	180	N/A
	6.5	81.25	470	270
	12.5	156.25	1000	680
	24.5	306.25	1800	1500
	47.5	593.75	2700	2200
	92.5	1156.25	6800	5600
	247.5	3093.75	15000	12000
	640.5	8006.75	50000	50000
6 bits	2.5	31.25	220	N/A
	6.5	81.25	560	330
	12.5	156.25	1200	1000
	24.5	306.25	2700	2200
	47.5	593.75	3900	3300
	92.5	1156.25	8200	6800
	247.5	3093.75	18000	15000
	640.5	8006.75	50000	50000

1. Guaranteed by design.
2. The I/O analog switch voltage booster is enable when $V_{DDA} < 2.4\text{ V}$ (BOOSTEN = 1 in the SYSCFG_CFGR1 when $V_{DDA} < 2.4\text{ V}$). It is disable when $V_{DDA} \geq 2.4\text{ V}$.
3. Fast channels are: PC0, PC1, PC2, PC3, PA0.
4. Slow channels are: all ADC inputs except the fast channels.

Table 70. ADC accuracy - limited test conditions 1⁽¹⁾(2)(3)

Sym-bol	Parameter	Conditions ⁽⁴⁾		Min	Typ	Max	Unit	
ET	Total unadjusted error		Single ended	Fast channel (max speed)	-	4	5	LSB
				Slow channel (max speed)	-	4	5	
			Differential	Fast channel (max speed)	-	3.5	4.5	
				Slow channel (max speed)	-	3.5	4.5	
EO	Offset error		Single ended	Fast channel (max speed)	-	1	2.5	
				Slow channel (max speed)	-	1	2.5	
			Differential	Fast channel (max speed)	-	1.5	2.5	
				Slow channel (max speed)	-	1.5	2.5	
EG	Gain error		Single ended	Fast channel (max speed)	-	2.5	4.5	
				Slow channel (max speed)	-	2.5	4.5	
			Differential	Fast channel (max speed)	-	2.5	3.5	
				Slow channel (max speed)	-	2.5	3.5	
ED	Differential linearity error		Single ended	Fast channel (max speed)	-	1	1.5	
				Slow channel (max speed)	-	1	1.5	
			Differential	Fast channel (max speed)	-	1	1.2	
				Slow channel (max speed)	-	1	1.2	
EL	Integral linearity error		Single ended	Fast channel (max speed)	-	1.5	2.5	
				Slow channel (max speed)	-	1.5	2.5	
			Differential	Fast channel (max speed)	-	1	2	
				Slow channel (max speed)	-	1	2	
ENOB	Effective number of bits		Single ended	Fast channel (max speed)	10.4	10.5	-	bits
				Slow channel (max speed)	10.4	10.5	-	
			Differential	Fast channel (max speed)	10.8	10.9	-	
				Slow channel (max speed)	10.8	10.9	-	
SINAD	Signal-to-noise and distortion ratio		Single ended	Fast channel (max speed)	64.4	65	-	dB
				Slow channel (max speed)	64.4	65	-	
			Differential	Fast channel (max speed)	66.8	67.4	-	
				Slow channel (max speed)	66.8	67.4	-	
SNR	Signal-to-noise ratio		Single ended	Fast channel (max speed)	65	66	-	
				Slow channel (max speed)	65	66	-	
			Differential	Fast channel (max speed)	67	68	-	
				Slow channel (max speed)	67	68	-	

ADC clock frequency ≤ 80 MHz,
 Sampling rate ≤ 5.33 Msps,
 V_{DDA} = VREF+ = 3 V,
 TA = 25 °C

Table 70. ADC accuracy - limited test conditions 1⁽¹⁾(2)(3) (continued)

Sym- bol	Parameter	Conditions ⁽⁴⁾		Min	Typ	Max	Unit	
THD	Total harmonic distortion	ADC clock frequency ≤ 80 MHz, Sampling rate ≤ 5.33 Msps, V _{DDA} = V _{REF+} = 3 V, TA = 25 °C	Single ended	Fast channel (max speed)	-	-74	-73	dB
				Slow channel (max speed)	-	-74	-73	
			Differential	Fast channel (max speed)	-	-79	-76	
				Slow channel (max speed)	-	-79	-76	

1. Guaranteed by design.
2. ADC DC accuracy values are measured after internal calibration.
3. ADC accuracy vs. negative Injection Current: Injecting negative current on any analog input pins should be avoided as this significantly reduces the accuracy of the conversion being performed on another analog input. It is recommended to add a Schottky diode (pin to ground) to analog pins which may potentially inject negative current.
4. The I/O analog switch voltage booster is enable when V_{DDA} < 2.4 V (BOOSTEN = 1 in the SYSCFG_CFGR1 when V_{DDA} < 2.4 V). It is disable when V_{DDA} ≥ 2.4 V. No oversampling.

Table 71. ADC accuracy - limited test conditions 2⁽¹⁾⁽²⁾⁽³⁾

Symbol	Parameter	Conditions ⁽⁴⁾		Min	Typ	Max	Unit
ET	Total unadjusted error	Single ended	Fast channel (max speed)	-	4	6.5	LSB
			Slow channel (max speed)	-	4	6.5	
		Differential	Fast channel (max speed)	-	3.5	5.5	
			Slow channel (max speed)	-	3.5	5.5	
EO	Offset error	Single ended	Fast channel (max speed)	-	1	4.5	
			Slow channel (max speed)	-	1	5	
		Differential	Fast channel (max speed)	-	1.5	3	
			Slow channel (max speed)	-	1.5	3	
EG	Gain error	Single ended	Fast channel (max speed)	-	2.5	6	
			Slow channel (max speed)	-	2.5	6	
		Differential	Fast channel (max speed)	-	2.5	3.5	
			Slow channel (max speed)	-	2.5	3.5	
ED	Differential linearity error	Single ended	Fast channel (max speed)	-	1	1.5	
			Slow channel (max speed)	-	1	1.5	
		Differential	Fast channel (max speed)	-	1	1.2	
			Slow channel (max speed)	-	1	1.2	
EL	Integral linearity error	Single ended	Fast channel (max speed)	-	1.5	3.5	
			Slow channel (max speed)	-	1.5	3.5	
		Differential	Fast channel (max speed)	-	1	3	
			Slow channel (max speed)	-	1	2.5	
ENOB	Effective number of bits	Single ended	Fast channel (max speed)	10	10.5	-	bits
			Slow channel (max speed)	10	10.5	-	
		Differential	Fast channel (max speed)	10.7	10.9	-	
			Slow channel (max speed)	10.7	10.9	-	
SINAD	Signal-to-noise and distortion ratio	Single ended	Fast channel (max speed)	62	65	-	dB
			Slow channel (max speed)	62	65	-	
		Differential	Fast channel (max speed)	66	67.4	-	
			Slow channel (max speed)	66	67.4	-	
SNR	Signal-to-noise ratio	Single ended	Fast channel (max speed)	64	66	-	
			Slow channel (max speed)	64	66	-	
		Differential	Fast channel (max speed)	66.5	68	-	
			Slow channel (max speed)	66.5	68	-	

Table 71. ADC accuracy - limited test conditions 2⁽¹⁾(2)(3) (continued)

Sym-bol	Parameter	Conditions ⁽⁴⁾		Min	Typ	Max	Unit	
THD	Total harmonic distortion	ADC clock frequency ≤ 80 MHz, Sampling rate ≤ 5.33 Msps, 2 V ≤ V _{DDA}	Single ended	Fast channel (max speed)	-	-74	-65	dB
				Slow channel (max speed)	-	-74	-67	
			Differential	Fast channel (max speed)	-	-79	-70	
				Slow channel (max speed)	-	-79	-71	

1. Guaranteed by design.
2. ADC DC accuracy values are measured after internal calibration.
3. ADC accuracy vs. negative Injection Current: Injecting negative current on any analog input pins should be avoided as this significantly reduces the accuracy of the conversion being performed on another analog input. It is recommended to add a Schottky diode (pin to ground) to analog pins which may potentially inject negative current.
4. The I/O analog switch voltage booster is enable when V_{DDA} < 2.4 V (BOOSTEN = 1 in the SYSCFG_CFGR1 when V_{DDA} < 2.4 V). It is disable when V_{DDA} ≥ 2.4 V. No oversampling.

Table 72. ADC accuracy - limited test conditions 3⁽¹⁾⁽²⁾⁽³⁾

Symbol	Parameter	Conditions ⁽⁴⁾		Min	Typ	Max	Unit
ET	Total unadjusted error	Single ended	Fast channel (max speed)	-	5.5	7.5	LSB
			Slow channel (max speed)	-	4.5	6.5	
		Differential	Fast channel (max speed)	-	4.5	7.5	
			Slow channel (max speed)	-	4.5	5.5	
EO	Offset error	Single ended	Fast channel (max speed)	-	2	5	
			Slow channel (max speed)	-	2.5	5	
		Differential	Fast channel (max speed)	-	2	3.5	
			Slow channel (max speed)	-	2.5	3	
EG	Gain error	Single ended	Fast channel (max speed)	-	4.5	7	
			Slow channel (max speed)	-	3.5	6	
		Differential	Fast channel (max speed)	-	3.5	4	
			Slow channel (max speed)	-	3.5	5	
ED	Differential linearity error	Single ended	Fast channel (max speed)	-	1.2	1.5	
			Slow channel (max speed)	-	1.2	1.5	
		Differential	Fast channel (max speed)	-	1	1.2	
			Slow channel (max speed)	-	1	1.2	
EL	Integral linearity error	Single ended	Fast channel (max speed)	-	3	3.5	
			Slow channel (max speed)	-	2.5	3.5	
		Differential	Fast channel (max speed)	-	2	2.5	
			Slow channel (max speed)	-	2	2.5	
ENOB	Effective number of bits	Single ended	Fast channel (max speed)	10	10.4	-	
			Slow channel (max speed)	10	10.4	-	
		Differential	Fast channel (max speed)	10.6	10.7	-	
			Slow channel (max speed)	10.6	10.7	-	
SINAD	Signal-to-noise and distortion ratio	Single ended	Fast channel (max speed)	62	64	-	
			Slow channel (max speed)	62	64	-	
		Differential	Fast channel (max speed)	65	66	-	
			Slow channel (max speed)	65	66	-	
SNR	Signal-to-noise ratio	Single ended	Fast channel (max speed)	63	65	-	
			Slow channel (max speed)	63	65	-	
		Differential	Fast channel (max speed)	66	67	-	
			Slow channel (max speed)	66	67	-	

Table 72. ADC accuracy - limited test conditions 3⁽¹⁾(2)(3) (continued)

Sym- bol	Parameter	Conditions ⁽⁴⁾		Min	Typ	Max	Unit	
THD	Total harmonic distortion	ADC clock frequency ≤ 80 MHz, Sampling rate ≤ 5.33 Msps, 1.65 V ≤ V _{DDA} = V _{REF+} ≤ 3.6 V, Voltage scaling Range 1	Single ended	Fast channel (max speed)	-	-69	-67	dB
				Slow channel (max speed)	-	-71	-67	
			Differential	Fast channel (max speed)	-	-72	-71	
				Slow channel (max speed)	-	-72	-71	

1. Guaranteed by design.
2. ADC DC accuracy values are measured after internal calibration.
3. ADC accuracy vs. negative Injection Current: Injecting negative current on any analog input pins should be avoided as this significantly reduces the accuracy of the conversion being performed on another analog input. It is recommended to add a Schottky diode (pin to ground) to analog pins which may potentially inject negative current.
4. The I/O analog switch voltage booster is enable when V_{DDA} < 2.4 V (BOOSTEN = 1 in the SYSCFG_CFGR1 when V_{DDA} < 2.4 V). It is disable when V_{DDA} ≥ 2.4 V. No oversampling.

Table 73. ADC accuracy - limited test conditions 4⁽¹⁾⁽²⁾⁽³⁾

Symbol	Parameter	Conditions ⁽⁴⁾		Min	Typ	Max	Unit
ET	Total unadjusted error	Single ended	Fast channel (max speed)	-	5	5.4	LSB
			Slow channel (max speed)	-	4	5	
		Differential	Fast channel (max speed)	-	4	5	
			Slow channel (max speed)	-	3.5	4.5	
EO	Offset error	Single ended	Fast channel (max speed)	-	2	4	
			Slow channel (max speed)	-	2	4	
		Differential	Fast channel (max speed)	-	2	3.5	
			Slow channel (max speed)	-	2	3.5	
EG	Gain error	Single ended	Fast channel (max speed)	-	4	4.5	
			Slow channel (max speed)	-	4	4.5	
		Differential	Fast channel (max speed)	-	3	4	
			Slow channel (max speed)	-	3	4	
ED	Differential linearity error	Single ended	Fast channel (max speed)	-	1	1.5	
			Slow channel (max speed)	-	1	1.5	
		Differential	Fast channel (max speed)	-	1	1.2	
			Slow channel (max speed)	-	1	1.2	
EL	Integral linearity error	Single ended	Fast channel (max speed)	-	2.5	3	
			Slow channel (max speed)	-	2.5	3	
		Differential	Fast channel (max speed)	-	2	2.5	
			Slow channel (max speed)	-	2	2.5	
ENOB	Effective number of bits	Single ended	Fast channel (max speed)	10.2	10.5	-	bits
			Slow channel (max speed)	10.2	10.5	-	
		Differential	Fast channel (max speed)	10.6	10.7	-	
			Slow channel (max speed)	10.6	10.7	-	
SINAD	Signal-to-noise and distortion ratio	Single ended	Fast channel (max speed)	63	65	-	dB
			Slow channel (max speed)	63	65	-	
		Differential	Fast channel (max speed)	65	66	-	
			Slow channel (max speed)	65	66	-	
SNR	Signal-to-noise ratio	Single ended	Fast channel (max speed)	64	65	-	
			Slow channel (max speed)	64	65	-	
		Differential	Fast channel (max speed)	66	67	-	
			Slow channel (max speed)	66	67	-	

Table 73. ADC accuracy - limited test conditions 4⁽¹⁾(2)(3) (continued)

Sym-bol	Parameter	Conditions ⁽⁴⁾		Min	Typ	Max	Unit	
THD	Total harmonic distortion	ADC clock frequency ≤ 26 MHz, 1.65 V ≤ V _{DDA} = VREF+ ≤ 3.6 V, Voltage scaling Range 2	Single ended	Fast channel (max speed)	-	-71	-69	dB
				Slow channel (max speed)	-	-71	-69	
			Differential	Fast channel (max speed)	-	-73	-72	
				Slow channel (max speed)	-	-73	-72	

1. Guaranteed by design.
2. ADC DC accuracy values are measured after internal calibration.
3. ADC accuracy vs. negative Injection Current: Injecting negative current on any analog input pins should be avoided as this significantly reduces the accuracy of the conversion being performed on another analog input. It is recommended to add a Schottky diode (pin to ground) to analog pins which may potentially inject negative current.
4. The I/O analog switch voltage booster is enable when V_{DDA} < 2.4 V (BOOSTEN = 1 in the SYSCFG_CFGR1 when V_{DDA} < 2.4 V). It is disable when V_{DDA} ≥ 2.4 V. No oversampling.

Figure 30. ADC accuracy characteristics

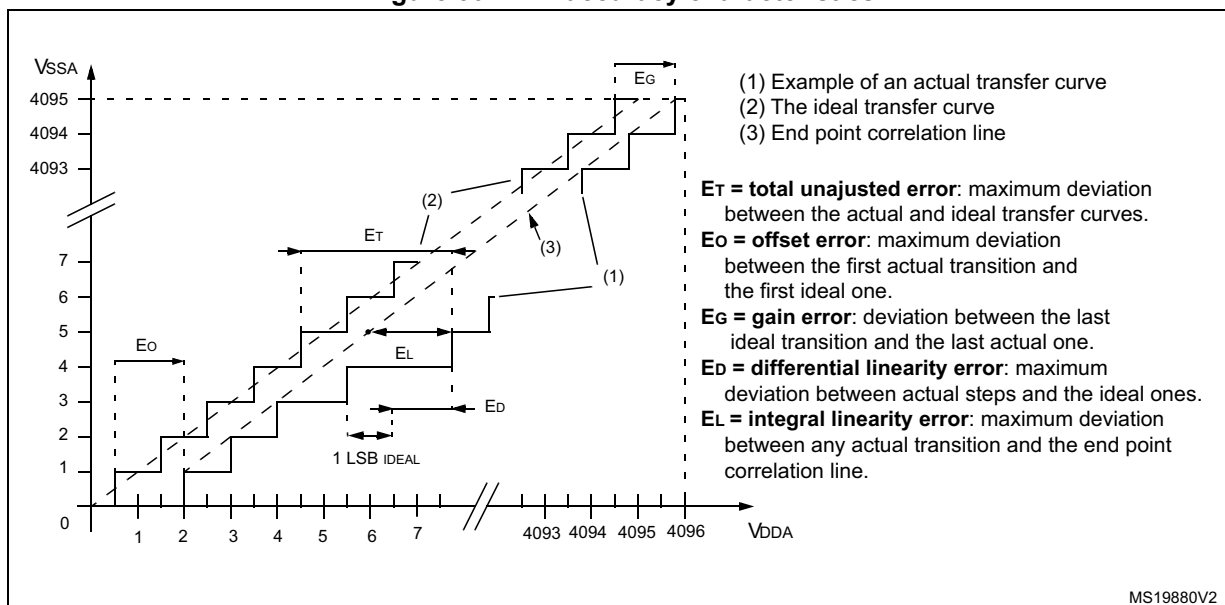
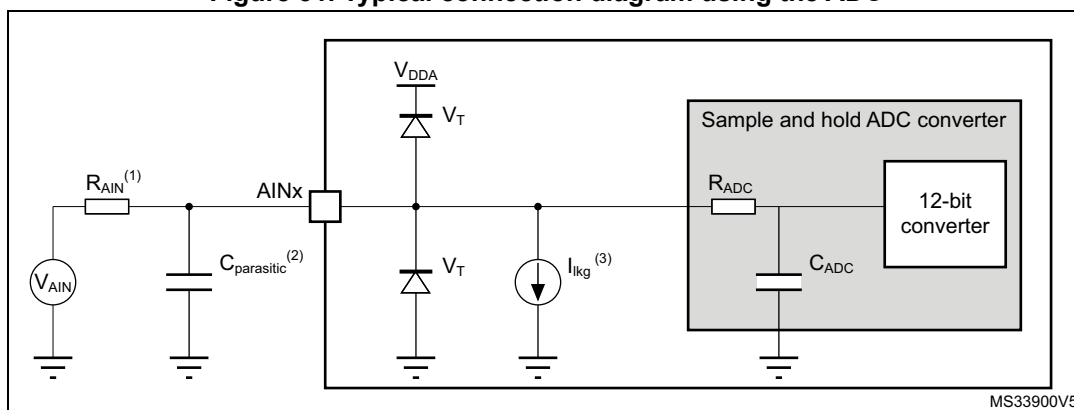


Figure 31. Typical connection diagram using the ADC



1. Refer to [Table 68: ADC characteristics](#) for the values of R_{AIN} and C_{ADC} .
2. $C_{parasitic}$ represents the capacitance of the PCB (dependent on soldering and PCB layout quality) plus the pad capacitance (refer to [Table 62: I/O static characteristics](#) for the value of the pad capacitance). A high $C_{parasitic}$ value downgrades the conversion accuracy. To remedy this, f_{ADC} should be reduced.
3. Refer to [Table 62: I/O static characteristics](#) for the values of I_{kg} .

General PCB design guidelines

Power supply decoupling should be performed as shown in [Figure 17: STM32L4Q5xx power supply scheme](#). The 10 nF capacitor should be ceramic (good quality) and it should be placed as close as possible to the chip.

6.3.19 Digital-to-Analog converter characteristics

Table 74. DAC characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min	Typ	Max	Unit	
V _{DDA}	Analog supply voltage for DAC ON	DAC output buffer OFF, DAC_OUT pin not connected (internal connection only)	1.71	-	3.6	V	
		Other modes	1.80	-			
V _{REF+}	Positive reference voltage	DAC output buffer OFF, DAC_OUT pin not connected (internal connection only)	1.71	-	V _{DDA}		
		Other modes	1.80	-			
V _{REF-}	Negative reference voltage	-	V _{SSA}				
R _L	Resistive load	DAC output buffer ON	connected to V _{SSA}	5	-		-
			connected to V _{DDA}	25	-	-	
R _O	Output Impedance	DAC output buffer OFF	9.6	11.7	13.8	kΩ	
R _{BON}	Output impedance sample and hold mode, output buffer ON	V _{DD} = 2.7 V	-	-	2	kΩ	
		V _{DD} = 2.0 V	-	-	3.5		
R _{BOFF}	Output impedance sample and hold mode, output buffer OFF	V _{DD} = 2.7 V	-	-	16.5	kΩ	
		V _{DD} = 2.0 V	-	-	18.0		
C _L	Capacitive load	DAC output buffer ON	-	-	50	pF	
C _{SH}		Sample and hold mode	-	0.1	1	μF	
V _{DAC_OUT}	Voltage on DAC_OUT output	DAC output buffer ON	0.2	-	V _{REF+} - 0.2	V	
		DAC output buffer OFF	0	-	V _{REF+}		
t _{SETTLING}	Settling time (full scale: for a 12-bit code transition between the lowest and the highest input codes when DAC_OUT reaches final value ±0.5LSB, ±1 LSB, ±2 LSB, ±4 LSB, ±8 LSB)	Normal mode DAC output buffer ON CL ≤ 50 pF, RL ≥ 5 kΩ	±0.5 LSB	-	1.7	3	μs
			±1 LSB	-	1.6	2.9	
			±2 LSB	-	1.55	2.85	
			±4 LSB	-	1.48	2.8	
			±8 LSB	-	1.4	2.75	
		Normal mode DAC output buffer OFF, ±1LSB, CL = 10 pF	-	2	2.5		
t _{WAKEUP} ⁽²⁾	Wakeup time from off state (setting the ENx bit in the DAC Control register) until final value ±1 LSB	Normal mode DAC output buffer ON CL ≤ 50 pF, RL ≥ 5 kΩ	-	4.2	7.5	μs	
		Normal mode DAC output buffer OFF, CL ≤ 10 pF	-	2	5		
PSRR	V _{DDA} supply rejection ratio	Normal mode DAC output buffer ON CL ≤ 50 pF, RL = 5 kΩ, DC	-	-80	-28	dB	

Table 74. DAC characteristics⁽¹⁾ (continued)

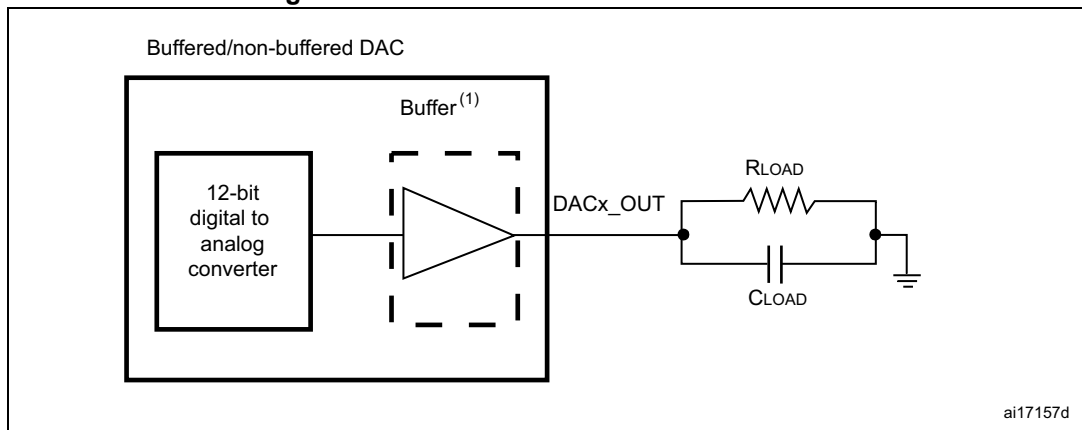
Symbol	Parameter	Conditions	Min	Typ	Max	Unit	
$T_{W_to_W}$	Minimal time between two consecutive writes into the DAC_DORx register to guarantee a correct DAC_OUT for a small variation of the input code (1 LSB) DAC_MCR:MODEx[2:0] = 000 or 001 DAC_MCR:MODEx[2:0] = 010 or 011	$CL \leq 50 \text{ pF}$, $RL \geq 5 \text{ k}\Omega$ $CL \leq 10 \text{ pF}$	1 1.4	-	-	μs	
t_{SAMP}	Sampling time in sample and hold mode (code transition between the lowest input code and the highest input code when DACOUT reaches final value $\pm 1\text{LSB}$)	DAC_OUT pin connected				ms	
			DAC output buffer ON, $C_{\text{SH}} = 100 \text{ nF}$	-	0.7		3.5
			DAC output buffer OFF, $C_{\text{SH}} = 100 \text{ nF}$	-	10.5	18	
		DAC_OUT pin not connected (internal connection only)				μs	
		DAC output buffer OFF	-	2	3.5	μs	
I_{leak}	Output leakage current	Sample and hold mode, DAC_OUT pin connected	-	-	-(3)	nA	
C_{int}	Internal sample and hold capacitor	-	5.2	7	8.8	pF	
t_{TRIM}	Middle code offset trim time	DAC output buffer ON	50	-	-	μs	
V_{offset}	Middle code offset for 1 trim code step	$V_{\text{REF+}} = 3.6 \text{ V}$	-	1500	-	μV	
		$V_{\text{REF+}} = 1.8 \text{ V}$	-	750	-		
$I_{\text{DDA(DAC)}}$	DAC consumption from V_{DDA}	DAC output buffer ON	No load, middle code (0x800)	-	315	500	μA
			No load, worst code (0xF1C)	-	450	670	
		DAC output buffer OFF	No load, middle code (0x800)	-	-	0.2	
		Sample and hold mode, $C_{\text{SH}} = 100 \text{ nF}$	-	$315 \times \frac{T_{\text{on}}}{T_{\text{on}} + T_{\text{off}}}$ (4)	$670 \times \frac{T_{\text{on}}}{T_{\text{on}} + T_{\text{off}}}$ (4)		

Table 74. DAC characteristics⁽¹⁾ (continued)

Symbol	Parameter	Conditions	Min	Typ	Max	Unit	
I _{DDV} (DAC)	DAC consumption from V _{REF+}	DAC output buffer ON	No load, middle code (0x800)	-	185	240	μA
			No load, worst code (0xF1C)	-	340	400	
		DAC output buffer OFF	No load, middle code (0x800)	-	155	205	
		Sample and hold mode, buffer ON, C _{SH} = 100 nF, worst case		-	185 x Ton/(Ton + Toff) ⁽⁴⁾	400 x Ton/(Ton + Toff) ⁽⁴⁾	
		Sample and hold mode, buffer OFF, C _{SH} = 100 nF, worst case		-	155 x Ton/(Ton + Toff) ⁽⁴⁾	205 x Ton/(Ton + Toff) ⁽⁴⁾	

1. Guaranteed by design.
2. In buffered mode, the output can overshoot above the final value for low input code (starting from min value).
3. Refer to [Table 62: I/O static characteristics](#).
4. Ton is the Refresh phase duration. Toff is the Hold phase duration. Refer to RM0432 reference manual for more details.

Figure 32. 12-bit buffered / non-buffered DAC



1. The DAC integrates an output buffer that can be used to reduce the output impedance and to drive external loads directly without the use of an external operational amplifier. The buffer can be bypassed by configuring the BOFFx bit in the DAC_CR register.

Table 75. DAC accuracy⁽¹⁾

Symbol	Parameter	Conditions	Min	Typ	Max	Unit	
DNL	Differential non linearity ⁽²⁾	DAC output buffer ON	-	-	±2	LSB	
		DAC output buffer OFF	-	-	±2		
-	monotonicity	10 bits	guaranteed				
INL	Integral non linearity ⁽³⁾	DAC output buffer ON CL ≤ 50 pF, RL ≥ 5 kΩ	-	-	±4		
		DAC output buffer OFF CL ≤ 50 pF, no RL	-	-	±4		
Offset	Offset error at code 0x800 ⁽³⁾	DAC output buffer ON CL ≤ 50 pF, RL ≥ 5 kΩ	V _{REF+} = 3.6 V	-	-		±12
			V _{REF+} = 1.8 V	-	-		±25
		DAC output buffer OFF CL ≤ 50 pF, no RL		-	-		±8
Offset1	Offset error at code 0x001 ⁽⁴⁾	DAC output buffer OFF CL ≤ 50 pF, no RL	-	-	±5		
OffsetCal	Offset Error at code 0x800 after calibration	DAC output buffer ON CL ≤ 50 pF, RL ≥ 5 kΩ	V _{REF+} = 3.6 V	-	-	±5	
			V _{REF+} = 1.8 V	-	-	±7	
Gain	Gain error ⁽⁵⁾	DAC output buffer ON CL ≤ 50 pF, RL ≥ 5 kΩ	-	-	±0.5	%	
		DAC output buffer OFF CL ≤ 50 pF, no RL	-	-	±0.5		
TUE	Total unadjusted error	DAC output buffer ON CL ≤ 50 pF, RL ≥ 5 kΩ	-	-	±30	LSB	
		DAC output buffer OFF CL ≤ 50 pF, no RL	-	-	±12		
TUECal	Total unadjusted error after calibration	DAC output buffer ON CL ≤ 50 pF, RL ≥ 5 kΩ	-	-	±23	LSB	
SNR	Signal-to-noise ratio	DAC output buffer ON CL ≤ 50 pF, RL ≥ 5 kΩ 1 kHz, BW 500 kHz	-	71.2	-	dB	
		DAC output buffer OFF CL ≤ 50 pF, no RL, 1 kHz BW 500 kHz	-	71.6	-		
THD	Total harmonic distortion	DAC output buffer ON CL ≤ 50 pF, RL ≥ 5 kΩ, 1 kHz	-	-78	-	dB	
		DAC output buffer OFF CL ≤ 50 pF, no RL, 1 kHz	-	-79	-		

Table 75. DAC accuracy⁽¹⁾ (continued)

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
SINAD	Signal-to-noise and distortion ratio	DAC output buffer ON CL ≤ 50 pF, RL ≥ 5 kΩ, 1 kHz	-	70.4	-	dB
		DAC output buffer OFF CL ≤ 50 pF, no RL, 1 kHz	-	71	-	
ENOB	Effective number of bits	DAC output buffer ON CL ≤ 50 pF, RL ≥ 5 kΩ, 1 kHz	-	11.4	-	bits
		DAC output buffer OFF CL ≤ 50 pF, no RL, 1 kHz	-	11.5	-	

1. Guaranteed by design.
2. Difference between two consecutive codes - 1 LSB.
3. Difference between measured value at Code i and the value at Code i on a line drawn between Code 0 and last Code 4095.
4. Difference between the value measured at Code (0x001) and the ideal value.
5. Difference between ideal slope of the transfer function and measured slope computed from code 0x000 and 0xFF when buffer is OFF, and from code giving 0.2 V and (V_{REF+} - 0.2) V when buffer is ON.

6.3.20 Voltage reference buffer characteristics

Table 76. VREFBUF characteristics⁽¹⁾

Symbol	Parameter	Conditions		Min	Typ	Max	Unit
V _{DDA}	Analog supply voltage	Normal mode	V _{RS} = 0	2.4	-	3.6	V
			V _{RS} = 1	2.8	-	3.6	
		Degraded mode ⁽²⁾	V _{RS} = 0	1.65	-	2.4	
			V _{RS} = 1	1.65	-	2.8	
V _{REFBUF_OUT}	Voltage reference output	Normal mode V _{DDA} = 3 V / 30°C I _{load} = 100 µA @ ±3σ	V _{RS} = 0	2.037	2.042	2.047	
			V _{RS} = 1	2.494	2.5	2.506	
		Degraded mode ⁽²⁾	V _{RS} = 0	V _{DDA} -150 mV	-	V _{DDA}	
			V _{RS} = 1	V _{DDA} -150 mV	-	V _{DDA}	
TRIM	Trim step resolution	-	-	-	±0.05	±0.1	%
CL	Load capacitor	-	-	0.5	1	1.5	µF
esr	Equivalent Serial Resistor of Cload	-	-	-	-	2	Ω
I _{load}	Static load current	-	-	-	-	4	mA
I _{line_reg}	Line regulation	2.8 V ≤ V _{DDA} ≤ 3.6 V	I _{load} = 500 µA	-	200	1000	ppm/V
			I _{load} = 4 mA	-	100	500	
I _{load_reg}	Load regulation	500 µA ≤ I _{load} ≤ 4 mA	Normal mode	-	50	500	ppm/mA
T _{Coeff}	Temperature coefficient	-40 °C < T _J < +125 °C		-	-	T _{coeff_vrefint} + 50	ppm/ °C
		0 °C < T _J < +50 °C		-	-	T _{coeff_vrefint} + 50	
PSRR	Power supply rejection	DC		40	60	-	dB
		100 kHz		25	40	-	
t _{START}	Start-up time	CL = 0.5 µF ⁽³⁾		-	300	350	µs
		CL = 1.1 µF ⁽³⁾		-	500	650	
		CL = 1.5 µF ⁽³⁾		-	650	800	
I _{INRUSH}	Control of maximum DC current drive on VREFBUF_OUT during start-up phase ⁽⁴⁾	-	-	-	8	-	mA

Table 76. VREFBUF characteristics⁽¹⁾ (continued)

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$I_{DDA}(VREFBUF)$	VREFBUF consumption from V_{DDA}	$I_{load} = 0 \mu A$	-	16	25	μA
		$I_{load} = 500 \mu A$	-	18	30	
		$I_{load} = 4 mA$	-	35	50	

1. Guaranteed by design, unless otherwise specified.
2. In degraded mode, the voltage reference buffer can not maintain accurately the output voltage which follows (V_{DDA} - drop voltage).
3. The capacitive load must include a 100 nF capacitor in order to cut-off the high frequency noise.
4. To correctly control the VREFBUF inrush current during start-up phase and scaling change, the V_{DDA} voltage must be in the range [2.4 V to 3.6 V] and [2.8 V to 3.6 V] respectively for $V_{RS} = 0$ and $V_{RS} = 1$.

6.3.21 Comparator characteristics

Table 77. COMP characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min	Typ	Max	Unit	
V_{DDA}	Analog supply voltage	-	1.62	-	3.6	V	
V_{IN}	Comparator input voltage range	-	0	-	V_{DDA}		
$V_{BG}^{(2)}$	Scaler input voltage	-	V_{REFINT}				
V_{SC}	Scaler offset voltage	-	-	± 5	± 10	mV	
$I_{DDA(SCALER)}$	Scaler static consumption from V_{DDA}	BRG_EN=0 (bridge disable)	-	200	300	nA	
		BRG_EN=1 (bridge enable)	-	0.8	1	μA	
t_{START_SCALER}	Scaler startup time	-	-	100	200	μs	
t_{START}	Comparator startup time to reach propagation delay specification	High-speed mode	$V_{DDA} \geq 2.7 V$	-	-	5	μs
			$V_{DDA} < 2.7 V$	-	-	7	
		Medium mode	$V_{DDA} \geq 2.7 V$	-	-	15	
			$V_{DDA} < 2.7 V$	-	-	25	
Ultra-low-power mode		-	-	80			
$t_D^{(3)}$	Propagation delay for 200 mV step with 100 mV overdrive	High-speed mode	$V_{DDA} \geq 2.7 V$	-	55	80	ns
			$V_{DDA} < 2.7 V$	-	65	100	
		Medium mode	$V_{DDA} \geq 2.7 V$	-	0.55	0.9	μs
			$V_{DDA} < 2.7 V$	-	0.65	1	
Ultra-low-power mode		-	5	12			
V_{offset}	Comparator offset error	Full common mode range	-	± 5	± 20	mV	
V_{hys}	Comparator hysteresis	No hysteresis		-	0	-	mV
		Low hysteresis		-	8	-	
		Medium hysteresis		-	15	-	
		High hysteresis		-	27	-	

Table 77. COMP characteristics⁽¹⁾ (continued)

Symbol	Parameter	Conditions		Min	Typ	Max	Unit
I _{DDA} (COMP)	Comparator consumption from V _{DDA}	Ultra-low-power mode	Static	-	400	600	nA
			With 50 kHz ±100 mV overdrive square signal	-	1200	-	
		Medium mode	Static	-	5	7	µA
			With 50 kHz ±100 mV overdrive square signal	-	6	-	
		High-speed mode	Static	-	70	100	
			With 50 kHz ±100 mV overdrive square signal	-	75	-	
I _{bias}	Comparator input bias current	-		-	-	-(4)	nA

1. Guaranteed by design, unless otherwise specified.
2. Refer to [Table 24: Embedded internal voltage reference](#).
3. Guaranteed by characterization results.
4. Mostly I/O leakage when used in analog mode. Refer to I_{Ikg} parameter in [Table 62: I/O static characteristics](#).

6.3.22 Operational amplifiers characteristics

Table 78. OPAMP characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
V _{DDA}	Analog supply voltage	-	1.8	-	3.6	V
CMIR	Common mode input range	-	0	-	V _{DDA}	V
V _I OFFSET	Input offset voltage	25 °C, No Load on output.	-	-	±1.5	mV
		All voltage/Temp.	-	-	±3	
ΔV _I OFFSET	Input offset voltage drift	Normal mode	-	±5	-	µV/°C
		Low-power mode	-	±10	-	
TRIMOFFSETP TRIMLPOFFSETP	Offset trim step at low common input voltage (0.1 × V _{DDA})	-	-	0.8	1.1	mV
TRIMOFFSETN TRIMLPOFFSETN	Offset trim step at high common input voltage (0.9 × V _{DDA})	-	-	1	1.35	

Table 78. OPAMP characteristics⁽¹⁾ (continued)

Symbol	Parameter	Conditions		Min	Typ	Max	Unit
I _{LOAD}	Drive current	Normal mode	V _{DDA} ≥ 2 V	-	-	500	μA
		Low-power mode		-	-	100	
I _{LOAD_PGA}	Drive current in PGA mode	Normal mode	V _{DDA} ≥ 2 V	-	-	450	
		Low-power mode		-	-	50	
R _{LOAD}	Resistive load (connected to VSSA or to VDDA)	Normal mode	V _{DDA} < 2 V	4	-	-	kΩ
		Low-power mode		20	-	-	
R _{LOAD_PGA}	Resistive load in PGA mode (connected to VSSA or to VDDA)	Normal mode	V _{DDA} < 2 V	4.5	-	-	
		Low-power mode		40	-	-	
C _{LOAD}	Capacitive load	-		-	-	50	pF
CMRR	Common mode rejection ratio	Normal mode		-	-85	-	dB
		Low-power mode		-	-90	-	
PSRR	Power supply rejection ratio	Normal mode	C _{LOAD} ≤ 50 pf, R _{LOAD} ≥ 4 kΩ DC	70	85	-	dB
		Low-power mode	C _{LOAD} ≤ 50 pf, R _{LOAD} ≥ 20 kΩ DC	72	90	-	
GBW	Gain Bandwidth Product	Normal mode	V _{DDA} ≥ 2.4 V (OPA_RANGE = 1)	550	1600	2200	kHz
		Low-power mode		100	420	600	
		Normal mode	V _{DDA} < 2.4 V (OPA_RANGE = 0)	250	700	950	
		Low-power mode		40	180	280	
SR ⁽²⁾	Slew rate (from 10 and 90% of output voltage)	Normal mode	V _{DDA} ≥ 2.4 V	-	700	-	V/ms
		Low-power mode		-	180	-	
		Normal mode	V _{DDA} < 2.4 V	-	300	-	
		Low-power mode		-	80	-	
AO	Open loop gain	Normal mode		55	110	-	dB
		Low-power mode		45	110	-	
V _{OHSAT} ⁽²⁾	High saturation voltage	Normal mode	I _{load} = max or R _{load} = min Input at V _{DDA} .	V _{DDA} - 100	-	-	mV
		Low-power mode		V _{DDA} - 50	-	-	
V _{OLSAT} ⁽²⁾	Low saturation voltage	Normal mode	I _{load} = max or R _{load} = min Input at 0.	-	-	100	
		Low-power mode		-	-	50	
Φ _m	Phase margin	Normal mode		-	74	-	°
		Low-power mode		-	66	-	

Table 78. OPAMP characteristics⁽¹⁾ (continued)

Symbol	Parameter	Conditions		Min	Typ	Max	Unit
GM	Gain margin	Normal mode		-	13	-	dB
		Low-power mode		-	20	-	
t _{WAKEUP}	Wake up time from OFF state.	Normal mode	C _{LOAD} ≤ 50 pf, R _{LOAD} ≥ 4 kΩ follower configuration	-	5	10	μs
		Low-power mode	C _{LOAD} ≤ 50 pf, R _{LOAD} ≥ 20 kΩ follower configuration	-	10	30	
I _{bias}	OPAMP input bias current	General purpose input (all packages except UFBGA132 and UFBGA169 only)		-	-	(3)	nA
		Dedicated input (UFBGA132 and UFBGA169 only)	T _J ≤ 75 °C	-	-	1	
			T _J ≤ 85 °C	-	-	3	
			T _J ≤ 105 °C	-	-	8	
		T _J ≤ 125 °C	-	-	15		
PGA gain ⁽²⁾	Non inverting gain value	-		-	2	-	-
				-	4	-	
				-	8	-	
				-	16	-	
R _{network}	R2/R1 internal resistance values in PGA mode ⁽⁴⁾	PGA Gain = 2		-	80/80	-	kΩ/kΩ
		PGA Gain = 4		-	120/40	-	
		PGA Gain = 8		-	140/20	-	
		PGA Gain = 16		-	150/10	-	
Delta R	Resistance variation (R1 or R2)	-		-15	-	15	%
PGA gain error	PGA gain error	-		-1	-	1	%
PGA BW	PGA bandwidth for different non inverting gain	Gain = 2	-	-	GBW/2	-	MHz
		Gain = 4	-	-	GBW/4	-	
		Gain = 8	-	-	GBW/8	-	
		Gain = 16	-	-	GBW/16	-	

Table 78. OPAMP characteristics⁽¹⁾ (continued)

Symbol	Parameter	Conditions		Min	Typ	Max	Unit
en	Voltage noise density	Normal mode	at 1 kHz, Output loaded with 4 kΩ	-	500	-	nV/√Hz
		Low-power mode	at 1 kHz, Output loaded with 20 kΩ	-	600	-	
		Normal mode	at 10 kHz, Output loaded with 4 kΩ	-	180	-	
		Low-power mode	at 10 kHz, Output loaded with 20 kΩ	-	290	-	
I _{DDA} (OPAMP) ⁽²⁾	OPAMP consumption from V _{DDA}	Normal mode	no Load, quiescent mode	-	120	260	μA
		Low-power mode		-	45	100	

1. Guaranteed by design, unless otherwise specified.
2. Guaranteed by characterization results.
3. Mostly I/O leakage, when used in analog mode. Refer to I_{lkg} parameter in [Table 62: I/O static characteristics](#).
4. R2 is the internal resistance between OPAMP output and OPAMP inverting input. R1 is the internal resistance between OPAMP inverting input and ground. The PGA gain = 1+R2/R1

6.3.23 Temperature sensor characteristics

Table 79. TS characteristics

Symbol	Parameter	Min	Typ	Max	Unit
$T_L^{(1)}$	V_{TS} linearity with temperature	-	± 1	± 2	$^{\circ}\text{C}$
Avg_Slope ⁽²⁾	Average slope	2.3	2.5	2.7	mV/ $^{\circ}\text{C}$
V_{30}	Voltage at 30 $^{\circ}\text{C}$ ($\pm 5^{\circ}\text{C}$) ⁽³⁾	0.742	0.76	0.785	V
$t_{\text{START}}^{(1)}$ (TS_BUF) ⁽¹⁾	Sensor Buffer Start-up time in continuous mode ⁽⁴⁾	-	8	15	μs
$t_{\text{START}}^{(1)}$	Start-up time when entering in continuous mode ⁽⁴⁾	-	70	120	μs
$t_{\text{S_temp}}^{(1)}$	ADC sampling time when reading the temperature	5	-	-	μs
$I_{\text{DD}}(\text{TS})^{(1)}$	Temperature sensor consumption from V_{DD} , when selected by ADC	-	4.7	7	μA

1. Guaranteed by design.
2. Guaranteed by characterization results.
3. Measured at $V_{\text{DDA}} = 3.0\text{ V} \pm 10\text{ mV}$. The V_{30} ADC conversion result is stored in the TS_CAL1 byte. Refer to [Table 8: Temperature sensor calibration values](#).
4. Continuous mode means Run/Sleep modes, or temperature sensor enable in Low-power run/Low-power sleep modes.

6.3.24 V_{BAT} monitoring characteristics

Table 80. V_{BAT} monitoring characteristics⁽¹⁾

Symbol	Parameter	Min	Typ	Max	Unit
R	Resistor bridge for V_{BAT}	-	39	-	k Ω
Q	Ratio on V_{BAT} measurement	-	3	-	-
$E_r^{(2)}$	Error on Q	-10	-	10	%
$t_{\text{S_vbat}}^{(2)}$	ADC sampling time when reading the VBAT	12	-	-	μs

1. $1.55\text{ V} < V_{\text{BAT}} < 3.6\text{ V}$
2. Guaranteed by design.

Table 81. V_{BAT} charging characteristics

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
R_{BC}	Battery charging resistor	VBRS = 0	-	5	-	k Ω
		VBRS = 1	-	1.5	-	

6.3.25 DFSDM characteristics

Unless otherwise specified, the parameters given in [Table 82](#) for DFSDM are derived from tests performed under the ambient temperature, f_{APB2} frequency and V_{DD} supply voltage conditions summarized in [Table 21: General operating conditions](#).

- Output speed is set to $OSPEEDRy[1:0] = 10$
- Capacitive load $C = 30\text{ pF}$
- Measurement points are done at CMOS levels: $0.5 \times V_{DD}$

Refer to [Section 6.3.14: I/O port characteristics](#) for more details on the input/output alternate function characteristics (DFSDM1_CKINy, DFSDM1_DATINy, DFSDM1_CKOUT for DFSDM).

Table 82. DFSDM characteristics⁽¹⁾

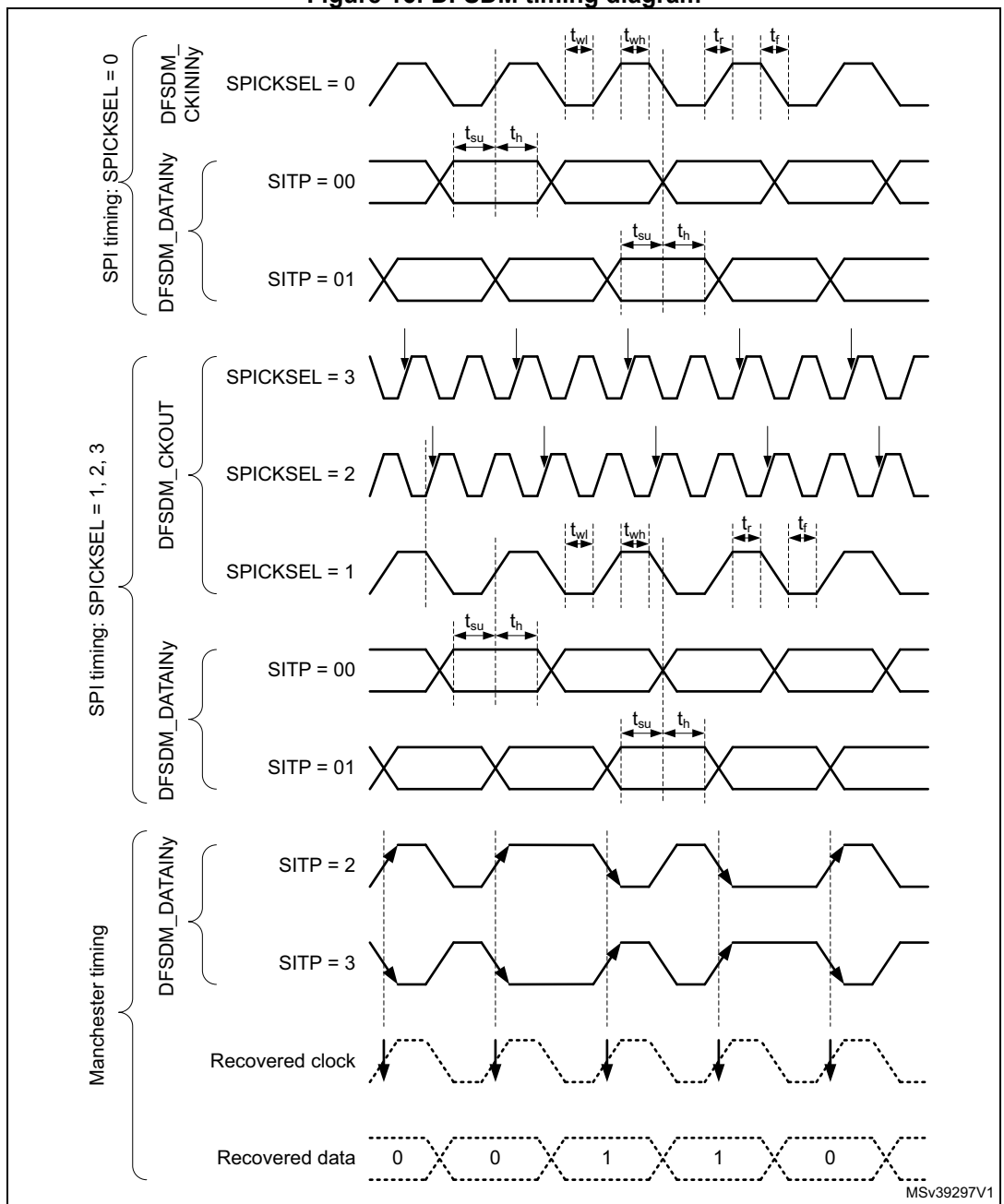
Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$f_{DFSDMCLK}$	DFSDM clock	$1.71 < V_{DD} < 3.6\text{ V}$	-	-	f_{SYSCLK}	
f_{CKIN} ($1/T_{CKIN}$)	Input clock frequency	SPI mode (SITP[1:0]=0,1), External clock mode (SPICKSEL[1:0]=0), $1.71 < V_{DD} < 3.6\text{ V}$	-	-	20	MHz
		SPI mode (SITP[1:0]=0,1), External clock mode (SPICKSEL[1:0]=0), $2.7 < V_{DD} < 3.6\text{ V}$	-	-	20	
		SPI mode (SITP[1:0]=0,1), Internal clock mode (SPICKSEL[1:0]≠0), $1.71 < V_{DD} < 3.6\text{ V}$	-	-	20	
		SPI mode (SITP[1:0]=0,1), Internal clock mode (SPICKSEL[1:0]≠0), $2.7 < V_{DD} < 3.6\text{ V}$	-	-	20	
f_{CKOUT}	Output clock frequency	$1.71 < V_{DD} < 3.6\text{ V}$	-	-	20	
$DuCy_{CKOUT}$	Output clock frequency duty cycle	$1.71 < V_{DD} < 3.6\text{ V}$	45	50	55	%

Table 82. DFSDM characteristics⁽¹⁾ (continued)

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$t_{wh(CKIN)}$ $t_{wl(CKIN)}$	Input clock high and low time	SPI mode (SITP[1:0]=0,1), External clock mode (SPICKSEL[1:0]=0), 1.71 < V _{DD} < 3.6 V	TCKIN/2-0.5	T _{CKIN} /2	-	ns
t_{su}	Data input setup time	SPI mode (SITP[1:0]=0,1), External clock mode (SPICKSEL[1:0]=0), 1.71 < V _{DD} < 3.6 V	2	-	-	
t_h	Data input hold time	SPI mode (SITP[1:0]=0,1), External clock mode (SPICKSEL[1:0]=0), 1.71 < V _{DD} < 3.6 V	0.5	-	-	
T _{Manchester}	Manchester data period (recovered clock period)	Manchester mode (SITP[1:0]=2,3), Internal clock mode (SPICKSEL[1:0]≠0), 1.71 < V _{DD} < 3.6 V	(CKOUTDIV+1) * T _{DFSDMCLK}	-	(2*CKOUTDIV) * T _{DFSDMCLK}	

1. Guaranteed by characterization results.

Figure 16: DFSDM timing diagram



6.3.26 Timer characteristics

The parameters given in the following tables are guaranteed by design.

Refer to [Section 6.3.14: I/O port characteristics](#) for details on the input/output alternate function characteristics (output compare, input capture, external clock, PWM output).

Table 83. TIMx⁽¹⁾ characteristics

Symbol	Parameter	Conditions	Min	Max	Unit
t _{res(TIM)}	Timer resolution time	-	1	-	t _{TIMxCLK}
		f _{TIMxCLK} = 120 MHz	8.33	-	ns
f _{EXT}	Timer external clock frequency on CH1 to CH4	-	0	f _{TIMxCLK} /2	MHz
		f _{TIMxCLK} = 120 MHz	0	60	MHz
Res _{TIM}	Timer resolution	TIMx (except TIM2 and TIM5)	-	16	bit
		TIM2 and TIM5	-	32	
t _{COUNTER}	16-bit counter clock period	-	1	65536	t _{TIMxCLK}
		f _{TIMxCLK} = 120 MHz	0.00833	546.13	µs
t _{MAX_COUNT}	Maximum possible count with 32-bit counter	-	-	65536 × 65536	t _{TIMxCLK}
		f _{TIMxCLK} = 120 MHz	-	35.77	s

1. TIMx is used as a general term in which x stands for 1,2,3,4,5,6,7,8,15,16 or 17.

Table 84. IWDG min/max timeout period at 32 kHz (LSI)⁽¹⁾

Prescaler divider	PR[2:0] bits	Min timeout RL[11:0]= 0x000	Max timeout RL[11:0]= 0xFFFF	Unit
/4	0	0.125	512	ms
/8	1	0.250	1024	
/16	2	0.500	2048	
/32	3	1.0	4096	
/64	4	2.0	8192	
/128	5	4.0	16384	
/256	6 or 7	8.0	32768	

1. The exact timings still depend on the phasing of the APB interface clock versus the LSI clock so that there is always a full RC period of uncertainty.

Table 85. WWDG min/max timeout value at 120 MHz (PCLK)

Prescaler	WDGTB	Min timeout value	Max timeout value	Unit
1	0	0.0341	2.1845	ms
2	1	0.0683	4.3691	
4	2	0.1356	8.7381	
8	3	0.2731	17.4763	

6.3.27 Communication interfaces characteristics

I²C interface characteristics

The I2C interface meets the timings requirements of the I²C-bus specification and user manual rev. 03 for:

- Standard-mode (Sm): with a bit rate up to 100 kbit/s
- Fast-mode (Fm): with a bit rate up to 400 kbit/s
- Fast-mode Plus (Fm+): with a bit rate up to 1 Mbit/s.

The I2C timings requirements are guaranteed by design when the I2C peripheral is properly configured (refer to RM0432 reference manual).

The SDA and SCL I/O requirements are met with the following restrictions: the SDA and SCL I/O pins are not “true” open-drain. When configured as open-drain, the PMOS connected between the I/O pin and V_{DDIOx} is disabled, but is still present. Only FT_f I/O pins support Fm+ low level output current maximum requirement. Refer to [Section 6.3.14: I/O port characteristics](#) for the I2C I/Os characteristics.

All I2C SDA and SCL I/Os embed an analog filter. Refer to [Table 86](#) below for the analog filter characteristics:

Table 86. I2C analog filter characteristics⁽¹⁾

Symbol	Parameter	Min	Max	Unit
t _{AF}	Maximum pulse width of spikes that are suppressed by the analog filter	50 ⁽²⁾	260 ⁽³⁾	ns

1. Guaranteed by design.
2. Spikes with widths below t_{AF(min)} are filtered.
3. Spikes with widths above t_{AF(max)} are not filtered.

SPI characteristics

Unless otherwise specified, the parameters given in [Table 87](#) for SPI are derived from tests performed under the ambient temperature, f_{PCLKx} frequency and supply voltage conditions summarized in [Table 21: General operating conditions](#).

- Output speed is set to $OSPEEDRy[1:0] = 11$
- Capacitive load $C = 30$ pF
- Measurement points are done at CMOS levels: $0.5 \times V_{DD}$

Refer to [Section 6.3.14: I/O port characteristics](#) for more details on the input/output alternate function characteristics (NSS, SCK, MOSI, MISO for SPI).

Table 87. SPI characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min	Typ	Max ⁽²⁾	Unit
f_{SCK} $1/t_{c(SCK)}$	SPI clock frequency	Master mode $2.7\text{ V} < V_{DD} < 3.6\text{ V}$ Voltage Range V1	-	-	60	MHz
		Master mode $1.71\text{ V} < V_{DD} < 3.6\text{ V}$ Voltage Range V1			45	
		Master transmitter mode $1.71\text{ V} < V_{DD} < 3.6\text{ V}$ Voltage Range V1			60	
		Slave receiver mode $1.71\text{ V} < V_{DD} < 3.6\text{ V}$ Voltage Range V1			60	
		Slave mode transmitter/full duplex $2.7\text{ V} < V_{DD} < 3.6\text{ V}$ Voltage Range V1			32	
		Slave mode transmitter/full duplex $1.71\text{ V} < V_{DD} < 3.6\text{ V}$ Voltage Range V1			22	
		$1.71\text{ V} < V_{DD} < 3.6\text{ V}$ Voltage Range V2			13	
		$1.08\text{ V} < V_{DD} < 1.32\text{ V}^{(3)}$			12	
$t_{su(NSS)}$	NSS setup time	Slave mode, SPI prescaler = 2	$4 \times T_{PCLK}$	-	-	ns
$t_{h(NSS)}$	NSS hold time	Slave mode, SPI prescaler = 2	$2 \times T_{PCLK}$	-	-	
$t_{w(SCKH)}$ $t_{w(SCKL)}$	SCK high and low time	Master mode	$T_{PCLK} - 1$	T_{PCLK}	$T_{PCLK} + 1$	
$t_{su(MI)}$	Data input setup time	Master mode	1.5	-	-	
$t_{su(SI)}$		Slave mode	2	-	-	
$t_{h(MI)}$	Data input hold time	Master mode	5.5	-	-	
$t_{h(SI)}$		Slave mode	5	-	-	
$t_{a(SO)}$	Data output access time	Slave mode	9	-	34	

Table 87. SPI characteristics⁽¹⁾ (continued)

Symbol	Parameter	Conditions	Min	Typ	Max ⁽²⁾	Unit
$t_{dis(SO)}$	Data output disable time	Slave mode	9	-	16	ns
$t_{v(SO)}$	Data output valid time	Slave mode $2.7\text{ V} < V_{DD} < 3.6\text{ V}$ Voltage Range V1	-	13	15.5	
		Slave mode $1.71\text{ V} < V_{DD} < 3.6\text{ V}$ Voltage Range V1	-	11	23	
		Slave mode $1.71\text{ V} < V_{DD} < 3.6\text{ V}$ Voltage Range V2	-	13	22	
		Slave mode ⁽³⁾ $1.08\text{ V} < V_{DD} < 1.32\text{ V}$	-	27	30	
$t_{v(MO)}$		Master mode	-	2	4	
$t_{h(SO)}$	Data output hold time	Slave mode $1.71\text{ V} < V_{DD} < 3.6\text{ V}$	9	-	-	
		Slave mode ⁽³⁾ $1.08 < V_{DD} < 1.32\text{ V}$	24	-	-	
$t_{h(MO)}$		Master mode	1	-	-	

1. Guaranteed by characterization results.
2. The maximum frequency in Slave transmitter mode is determined by the sum of $t_{v(SO)}$ and $t_{su(MI)}$ which has to fit into SCK low or high-phase preceding the SCK sampling edge. This value can be achieved when the SPI communicates with a master having $t_{su(MI)} = 0$ while $Duty(SCK) = 50\%$.
3. SPI mapped on GPIOG port which is supplied by VDDIO2 specified down to 1.08V. SPI is tested in this voltage.

Figure 33. SPI timing diagram - slave mode and CPHA = 0

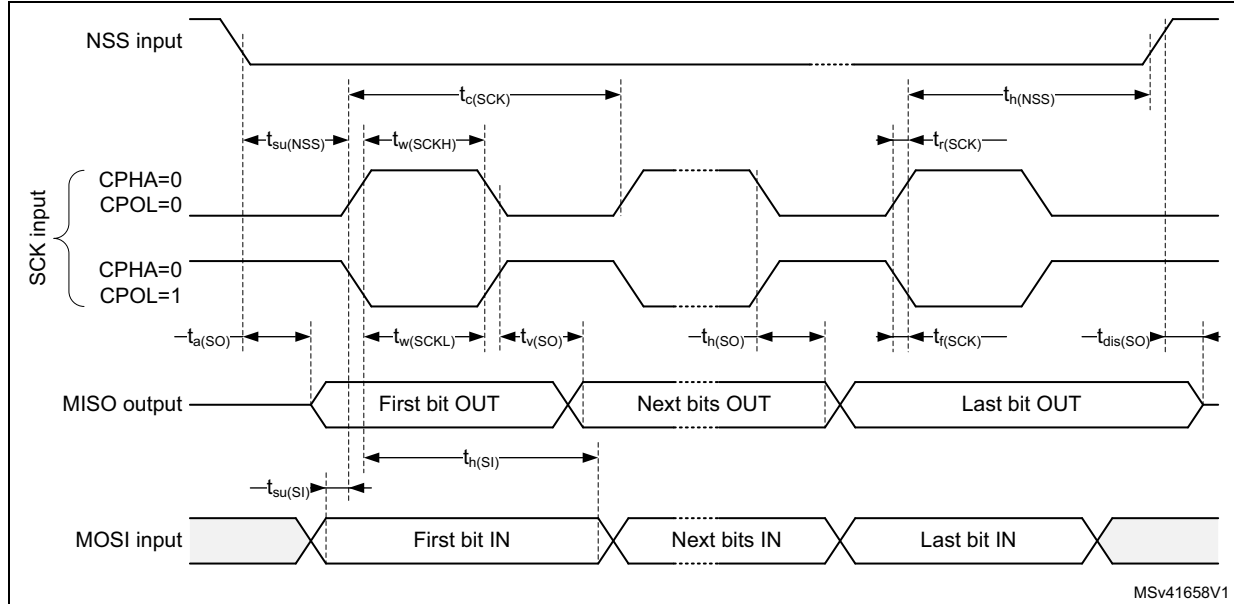
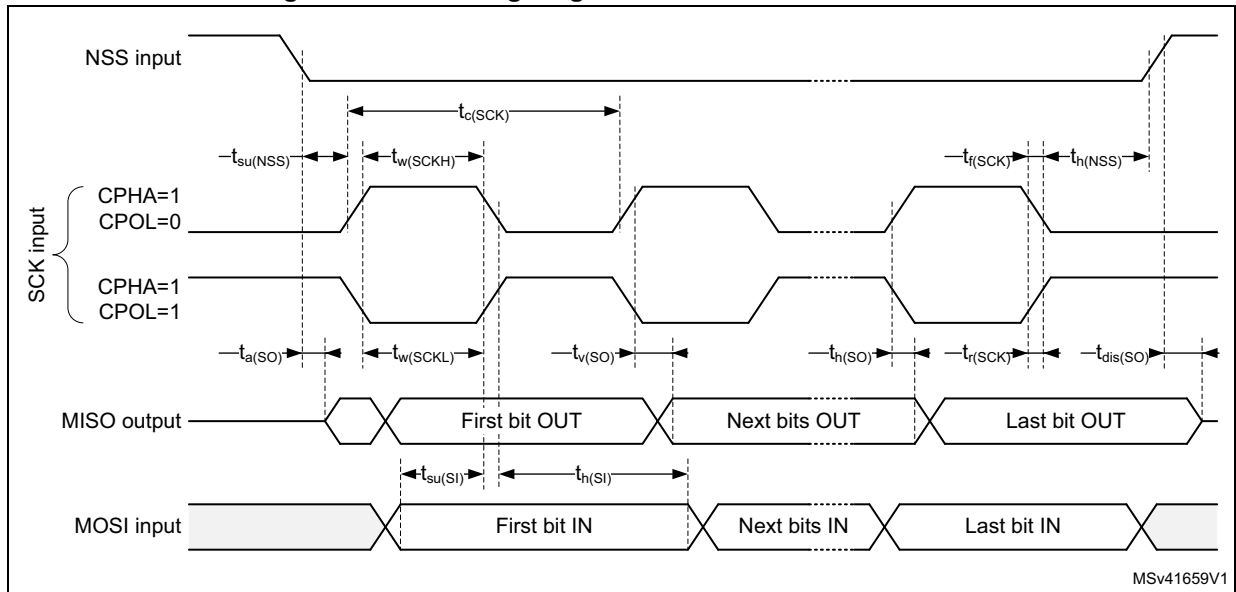
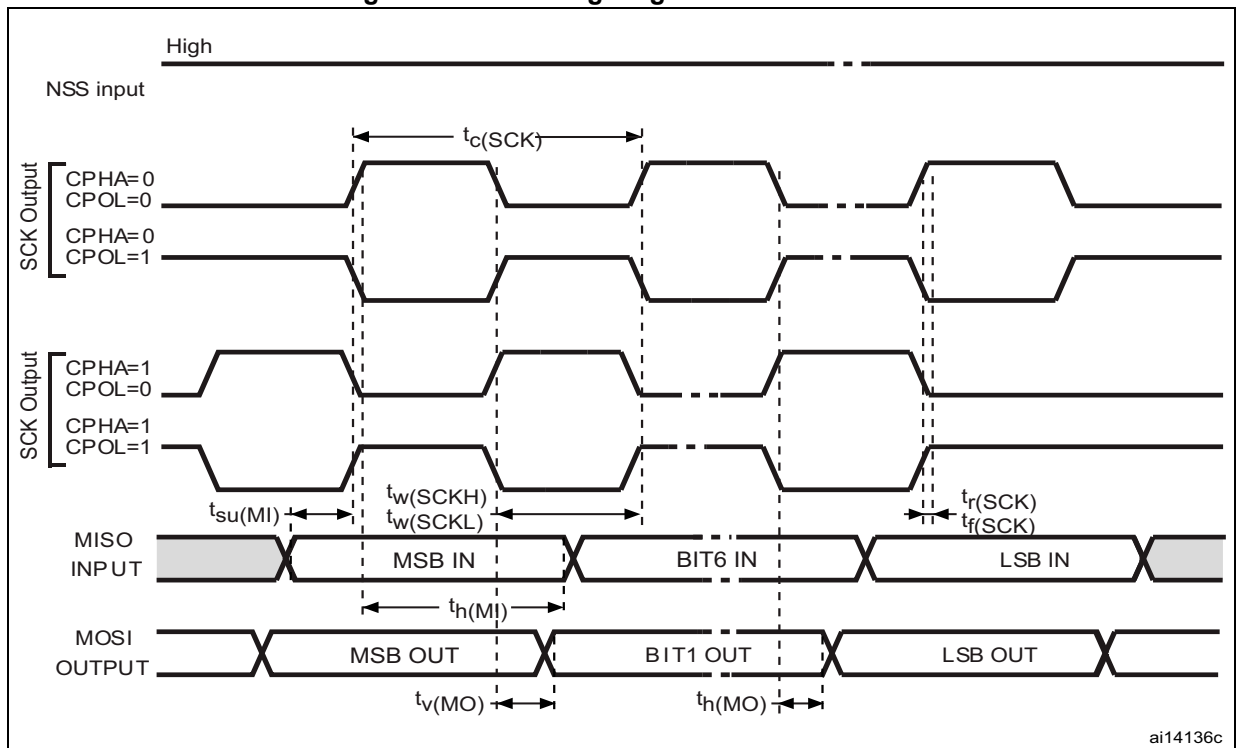


Figure 34. SPI timing diagram - slave mode and CPHA = 1



1. Measurement points are done at CMOS levels: 0.3 V_{DD} and 0.7 V_{DD} .

Figure 35. SPI timing diagram - master mode



1. Measurement points are done at CMOS levels: 0.3 V_{DD} and 0.7 V_{DD} .

SAI characteristics

Unless otherwise specified, the parameters given in [Table 88](#) for SAI are derived from tests performed under the ambient temperature, f_{PCLKx} frequency and V_{DD} supply voltage conditions summarized in [Table 21: General operating conditions](#), with the following configuration:

- Output speed is set to $OSPEEDRy[1:0] = 10$
- Capacitive load $C = 30$ pF
- Measurement points are done at CMOS levels: $0.5 \times V_{DD}$

Refer to [Section 6.3.14: I/O port characteristics](#) for more details on the input/output alternate function characteristics (CK,SD,FS).

Table 88. SAI characteristics⁽¹⁾

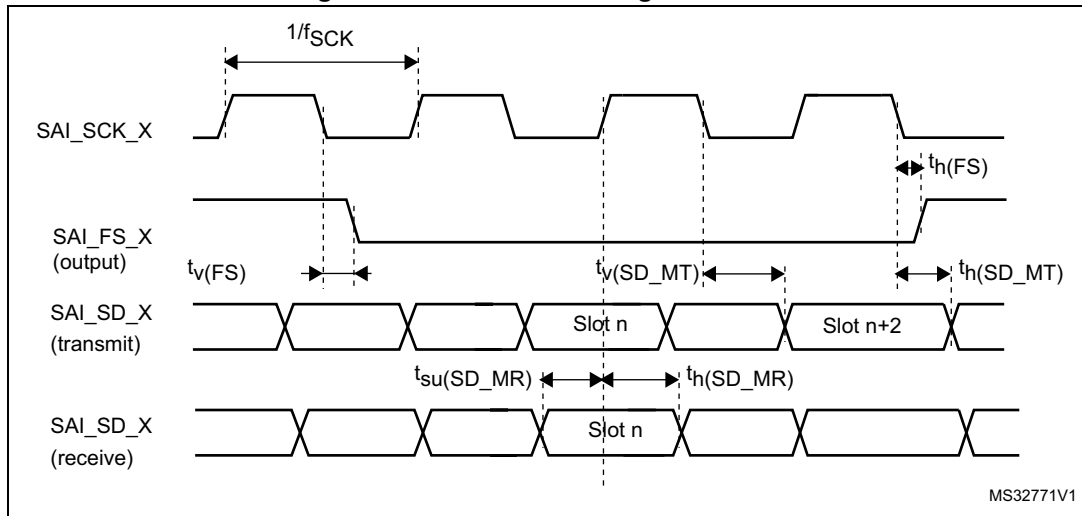
Symbol	Parameter	Conditions	Min	Max	Unit
f_{MCLK}	SAI Main clock output	-	-	50	MHz
f_{CK}	SAI clock frequency ⁽²⁾	Master transmitter $2.7 \leq V_{DD} \leq 3.6$ Voltage Range V1	-	30	
		Master transmitter $1.71 \leq V_{DD} \leq 3.6$ Voltage Range V1	-	20	
		Master receiver Voltage Range V1	-	15	
		Slave transmitter $2.7 \leq V_{DD} \leq 3.6$ Voltage Range V1	-	38	
		Slave transmitter $1.71 \leq V_{DD} \leq 3.6$ Voltage Range V1	-	26	
		Slave receiver Voltage Range V1	-	50	
		Voltage Range V2	-	13	
		$1.08V < V_{DD} < 1.32V$	-	8	
$t_{v(FS)}$	FS valid time	Master mode $2.7 \leq V_{DD} \leq 3.6$	-	15	ns
		Master mode $1.71 \leq V_{DD} \leq 3.6$	-	22	
$t_{h(FS)}$	FS hold time	Master mode	22	-	
$t_{su(FS)}$	FS setup time	Slave mode	2.5	-	
$t_{h(FS)}$	FS hold time	Slave mode	1	-	
$t_{su(SD_A_MR)}$	Data input setup time	Master receiver	3	-	
$t_{su(SD_B_SR)}$		Slave receiver	1	-	
$t_{h(SD_A_MR)}$	Data input hold time	Master receiver	3.5	-	
$t_{h(SD_B_SR)}$		Slave receiver	1.5	-	

Table 88. SAI characteristics⁽¹⁾ (continued)

Symbol	Parameter	Conditions	Min	Max	Unit
$t_{v(SD_B_ST)}$	Data output valid time	Slave transmitter (after enable edge) $2.7 \leq V_{DD} \leq 3.6$	-	13	ns
		Slave transmitter (after enable edge) $1.71 \leq V_{DD} \leq 3.6$	-	19	
		Slave transmitter (after enable edge) $1.08V < V_{DD} < 1.32V$	-	52	
		Slave transmitter Voltage Range V2	-	23.5	
$t_{h(SD_B_ST)}$	Data output hold time	Slave transmitter (after enable edge)	10	-	
$t_{v(SD_A_MT)}$	Data output valid time	Master transmitter (after enable edge) $2.7 \leq V_{DD} \leq 3.6$	-	16.5	
		Master transmitter (after enable edge) $1.71 \leq V_{DD} \leq 3.6$	-	24	
		Master transmitter (after enable edge) $1.08V < V_{DD} < 1.32V$	-	58	
$t_{h(SD_A_MT)}$	Data output hold time	Master transmitter (after enable edge)	10	-	

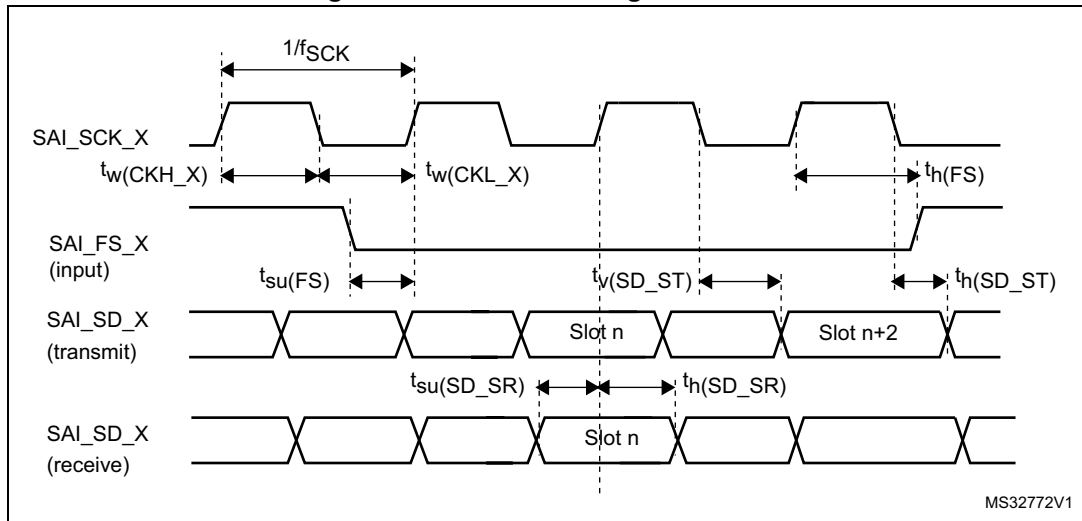
1. Guaranteed by characterization results.
2. APB clock frequency must be at least twice SAI clock frequency.

Figure 36. SAI master timing waveforms



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Figure 37. SAI slave timing waveforms



USB OTG full speed (FS) characteristics

The device’s USB interface is fully compliant with the USB specification version 2.0 and is USB-IF certified (for Full-speed device operation).

Table 89. USB electrical characteristics

Symbol	Parameter	Conditions	Min ⁽¹⁾	Typ	Max ⁽¹⁾	Unit
V _{DDUSB}	USB OTG full speed transceiver operating voltage	-	3.0 ⁽²⁾	-	3.6	V
V _{DI} ⁽³⁾	Differential input sensitivity	Over V _{CM} range	0.2	-	-	
V _{CM} ⁽³⁾	Differential input common mode range	Includes V _{DI} range	0.8	-	2.5	
V _{SE} ⁽³⁾	Single ended receiver input threshold	-	0.8	-	2.0	
V _{OL}	Static output level low	R _L of 1.5 kΩ to 3.6 V ⁽⁴⁾	-	-	0.3	
V _{OH}	Static output level high	R _L of 15 kΩ to 3.6 V ⁽⁴⁾	2.8	-	3.6	
R _{PD} ⁽³⁾	Pull down resistor on PA11, PA12 (USB_FS_DP/DM)	V _{IN} = V _{DD}	14.25	-	24.8	kΩ
R _{PU} ⁽³⁾	Pull Up Resistor on PA12 (USB_FS_DP)	V _{IN} = V _{SS} during idle	0.9	1.25	1.575	
	Pull Up Resistor on PA12 (USB_FS_DP)	V _{IN} = V _{SS} during reception	1.425	2.25	3.09	
	Pull Up Resistor on PA10 (OTG_FS_ID)	-	-	-	14.5	

1. All the voltages are measured from the local ground potential.

2. The STM32L4Q5xx USB OTG full speed transceiver functionality is ensured down to 2.7 V but not the full USB full speed electrical characteristics which are degraded in the 2.7-to-3.0 V V_{DD} voltage range.
3. Guaranteed by design.
4. R_L is the load connected on the USB OTG full speed drivers.

Note: When VBUS sensing feature is enabled, PA9 should be left at its default state (floating input), not as alternate function. A typical 200 μ A current consumption of the sensing block (current to voltage conversion to determine the different sessions) can be observed on PA9 when the feature is enabled.

Figure 38. USB OTG timings – definition of data signal rise and fall time

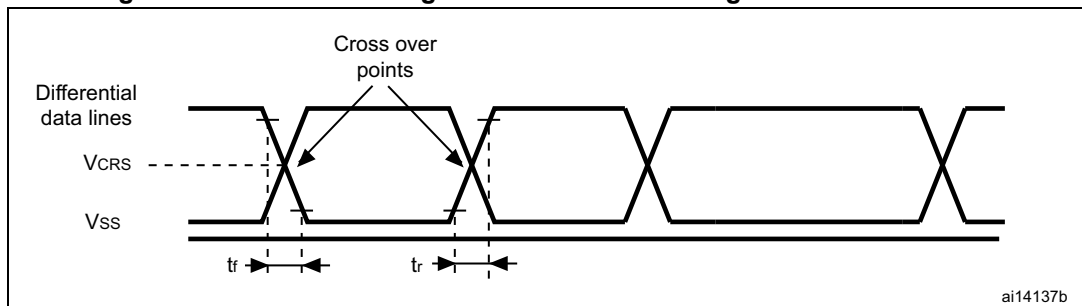


Table 90. USB OTG electrical characteristics⁽¹⁾

Driver characteristics					
Symbol	Parameter	Conditions	Min	Max	Unit
t_{rLS}	Rise time in LS ⁽²⁾	$C_L = 200$ to 600 pF	75	300	ns
t_{fLS}	Fall time in LS ⁽²⁾				
t_{rfmLS}	Rise/ fall time matching in LS	t_r / t_f	80	125	%
t_{rFS}	Rise time in FS ⁽²⁾	$C_L = 50$ pF	4	20	ns
t_{fFS}	Fall time in FS ⁽²⁾	$C_L = 50$ pF			
t_{rfmFS}	Rise/ fall time matching in FS	t_r / t_f	90	111	%
V_{CRS}	Output signal crossover voltage (LS/FS)	-	1.3	2.0	V
Z_{DRV}	Output driver impedance ⁽³⁾	Driving high or low	28	44	Ω

1. Guaranteed by design
2. Measured from 10% to 90% of the data signal. For more detailed informations, please refer to USB Specification - Chapter 7 (version 2.0).
3. No external termination series resistors are required on DP (D+) and DM (D-) pins since the matching impedance is included in the embedded driver.

Table 91. USB BCD DC electrical characteristics⁽¹⁾

Driver characteristics						
Symbol	Parameter	Conditions	Min	Typ	Max	Unit
I _{DD(USBBCD)}	Primary detection mode consumption	-	-	-	300	μA
	Secondary detection mode consumption	-	-	-		
RDAT_LKG	Data line leakage resistance	-	300	-	-	kΩ
VDAT_LKG	Data line leakage voltage	-	0.0	-	3.6	V
RDCP_DAT	Dedicated charging port resistance across D+/D-	-	-	-	200	Ω
VLGC_HI	Logic high	-	2.0	-	3.6	V
VLGC_LOW	Logic low	-	-	-	0.8	V
VLGC	Logic threshold	-	0.8	-	2.0	V
VDAT_REF	Data detect voltage	-	0.25	-	0.4	V
VDP_SRC	D+ source voltage	-	0.5	-	0.7	V
VDM_SRC	D- source voltage	-	0.5	-	0.7	V
IDP_SINK	D+ sink current	-	25	-	175	μA
IDM_SINK	D- sink current	-	25	-	175	μA

1. Guaranteed by design

CAN (controller area network) interface

Refer to [Section 6.3.14: I/O port characteristics](#) for more details on the input/output alternate function characteristics (CAN_TX and CAN_RX).

6.3.28 FSMC characteristics

Unless otherwise specified, the parameters given in [Table 92](#) to [Table 105](#) for the FMC interface are derived from tests performed under the ambient temperature, f_{HCLK} frequency and V_{DD} supply voltage conditions summarized in [Table 21](#), with the following configuration:

- Output speed is set to OSPEEDRy[1:0] = 11
- Capacitive load C = 30 pF
- Measurement points are done at CMOS levels: 0.5 x V_{DD}

Refer to [Section 6.3.14: I/O port characteristics](#) for more details on the input/output characteristics.

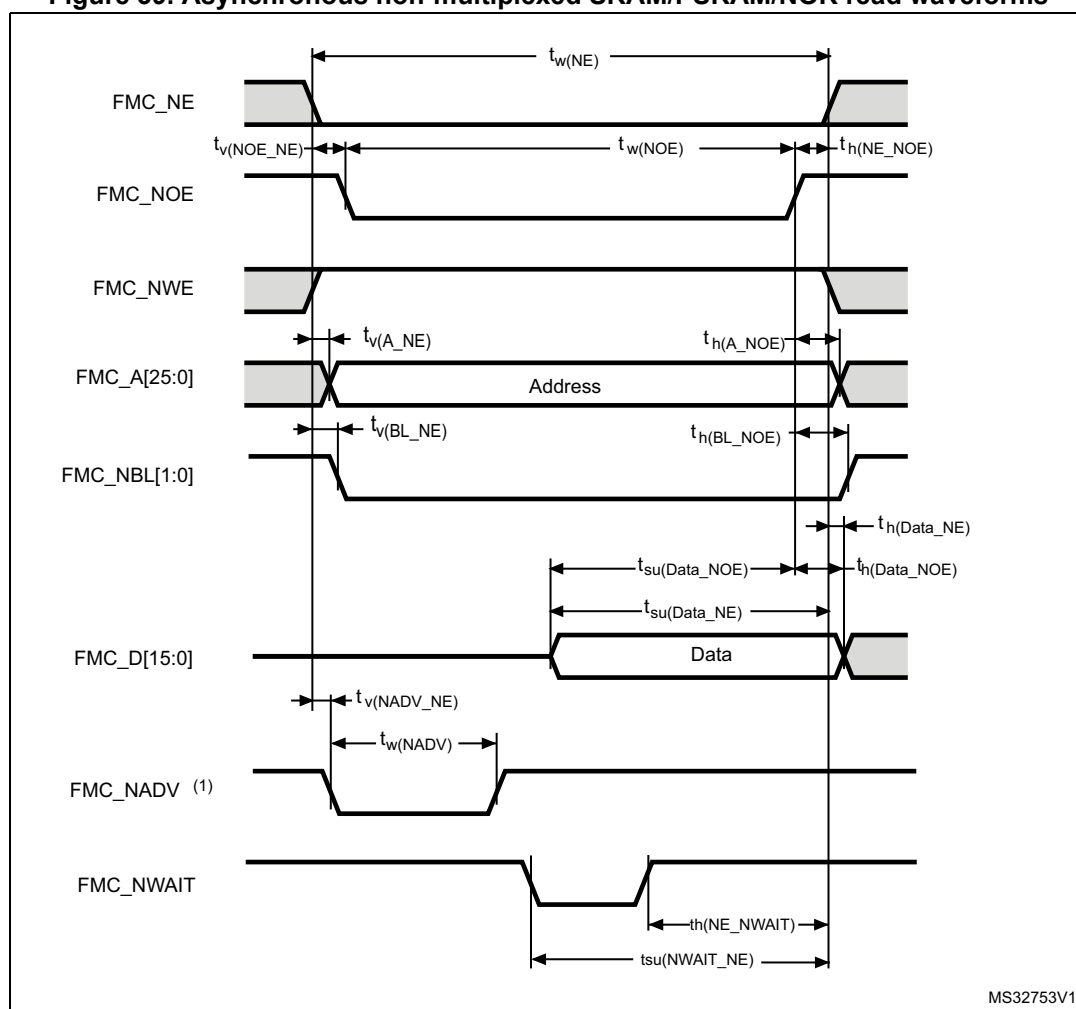
Asynchronous waveforms and timings

Figure 39 through Figure 42 represent asynchronous waveforms and Table 92 through Table 99 provide the corresponding timings. The results shown in these tables are obtained with the following FMC configuration:

- AddressSetupTime(ADDSET) = 0x1
- AddressHoldTime(ADDHLD) = 0x1
- DataHoldTime(DATAHLD) = 0x1 (1THCLK for read operations and 2THCLK for write operations)
- ByteLaneSetup(NBLSET)=0x1
- DataSetupTime(DATAST) = 0x1 (except for asynchronous NWAIT mode, DataSetupTime = 0x5)
- BusTurnAroundDuration = 0x0

In all timing tables, the THCLK is the HCLK clock period.

Figure 39. Asynchronous non-multiplexed SRAM/PSRAM/NOR read waveforms



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Table 92. Asynchronous non-multiplexed SRAM/PSRAM/NOR read timings⁽¹⁾

Symbol	Parameter	Min	Max	Unit
$t_{w(NE)}$	FMC_NE low time	$3T_{HCLK}-0.5$	$3T_{HCLK}+1$	ns
$t_{v(NOE_NE)}$	FMC_NEx low to FMC_NOE low	0	1	
$t_{w(NOE)}$	FMC_NOE low time	$2T_{HCLK}-0.5$	$2T_{HCLK}+1$	
$t_{h(NE_NOE)}$	FMC_NOE high to FMC_NE high hold time	T_{HCLK}	-	
$t_{v(A_NE)}$	FMC_NEx low to FMC_A valid	-	1	
$t_{h(A_NOE)}$	Address hold time after FMC_NOE high	$2T_{HCLK}-1$	-	
$t_{su(Data_NE)}$	Data to FMC_NEx high setup time	$T_{HCLK}+14$	-	
$t_{su(Data_NOE)}$	Data to FMC_NOEx high setup time	14	-	
$t_{h(Data_NOE)}$	Data hold time after FMC_NOE high	0	-	
$t_{h(Data_NE)}$	Data hold time after FMC_NEx high	0	-	
$t_{v(NADV_NE)}$	FMC_NEx low to FMC_NADV low	-	0	
$t_{w(NADV)}$	FMC_NADV low time	-	$T_{HCLK}+1.5$	

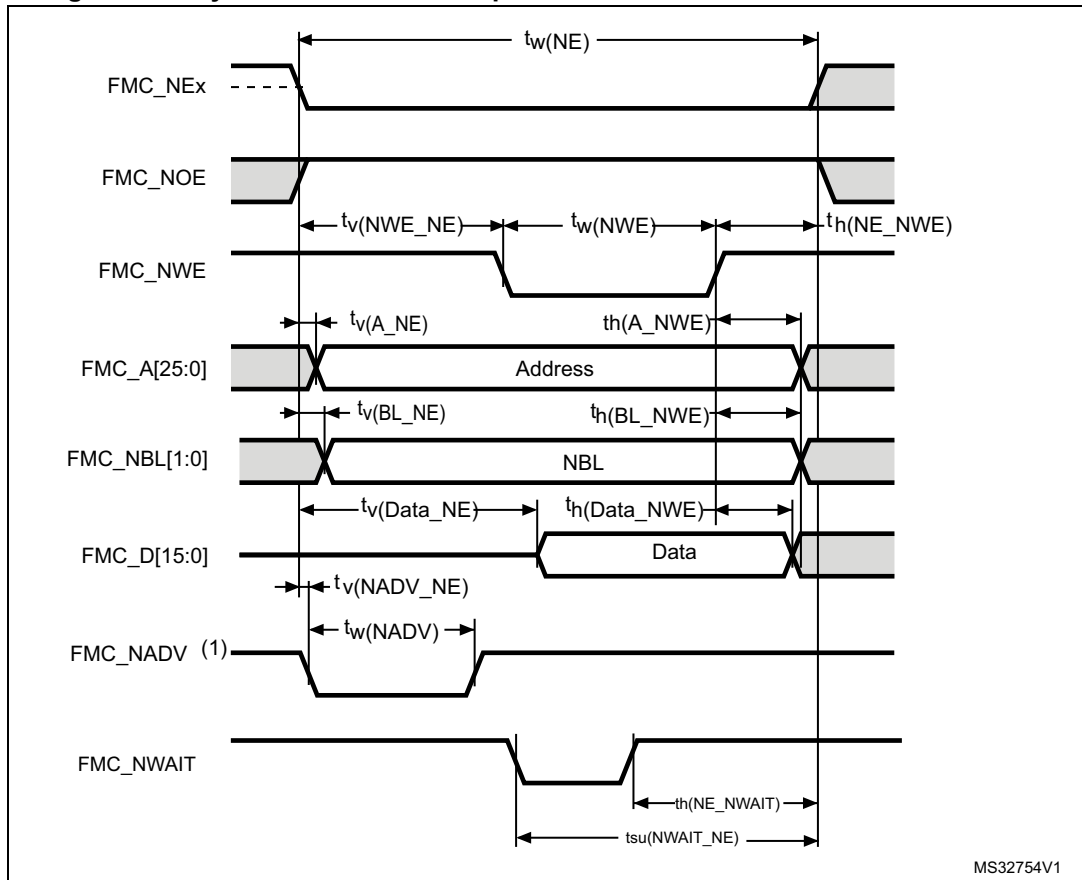
1. Guaranteed by characterization results.

Table 93. Asynchronous non-multiplexed SRAM/PSRAM/NOR read-NWAIT timings⁽¹⁾

Symbol	Parameter	Min	Max	Unit
$t_{w(NE)}$	FMC_NE low time	$8T_{HCLK}-0.5$	$8T_{HCLK}+1$	ns
$t_{w(NOE)}$	FMC_NWE low time	$7T_{HCLK}-0.5$	$7T_{HCLK}+0.5$	
$t_{w(NWAIT)}$	FMC_NWAIT low time	T_{HCLK}	-	
$t_{su(NWAIT_NE)}$	FMC_NWAIT valid before FMC_NEx high	$5T_{HCLK}+12.5$	-	
$t_{h(NE_NWAIT)}$	FMC_NEx hold time after FMC_NWAIT invalid	$4T_{HCLK}+12$	-	

1. Guaranteed by characterization results.

Figure 40. Asynchronous non-multiplexed SRAM/PSRAM/NOR write waveforms



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Table 94. Asynchronous non-multiplexed SRAM/PSRAM/NOR write timings⁽¹⁾

Symbol	Parameter	Min	Max	Unit
$t_{w(NE)}$	FMC_NE low time	$4T_{HCLK}-0.5$	$4T_{HCLK}+1$	ns
$t_{v(NWE_NE)}$	FMC_NEx low to FMC_NWE low	$T_{HCLK}-0.5$	$T_{HCLK}+1$	
$t_{w(NWE)}$	FMC_NWE low time	$T_{HCLK}-0.5$	$T_{HCLK}+1$	
$t_{h(NE_NWE)}$	FMC_NWE high to FMC_NE high hold time	$2T_{HCLK}-0.5$	-	
$t_{v(A_NE)}$	FMC_NEx low to FMC_A valid	-	0	
$t_{h(A_NWE)}$	Address hold time after FMC_NWE high	$2T_{HCLK}-1$	-	
$t_{v(BL_NE)}$	FMC_NEx low to FMC_NBL valid	-	T_{HCLK}	
$t_{h(BL_NWE)}$	FMC_NBL hold time after FMC_NWE high	$2T_{HCLK}-0.5$	-	
$t_{v(Data_NE)}$	Data to FMC_NEx low to Data valid	-	$T_{HCLK}+3$	
$t_{h(Data_NWE)}$	Data hold time after FMC_NWE high	$2T_{HCLK}+1$	-	
$t_{v(NADV_NE)}$	FMC_NEx low to FMC_NADV low	-	1	
$t_{w(NADV)}$	FMC_NADV low time	-	$T_{HCLK}+1.5$	

1. Guaranteed by characterization results.

Table 95. Asynchronous non-multiplexed SRAM/PSRAM/NOR write-NWAIT timings⁽¹⁾

Symbol	Parameter	Min	Max	Unit
$t_{w(NE)}$	FMC_NE low time	$9T_{HCLK}-0.5$	$9T_{HCLK}+1.5$	ns
$t_{w(NWE)}$	FMC_NWE low time	$6T_{HCLK}-0.5$	$6T_{HCLK}+1$	
$t_{su(NWAIT_NE)}$	FMC_NWAIT valid before FMC_NEx high	$7T_{HCLK}-13$	-	
$t_h(NE_NWAIT)$	FMC_NEx hold time after FMC_NWAIT invalid	$5T_{HCLK}+13$	-	

1. Guaranteed by characterization results.

Figure 41. Asynchronous multiplexed PSRAM/NOR read waveforms

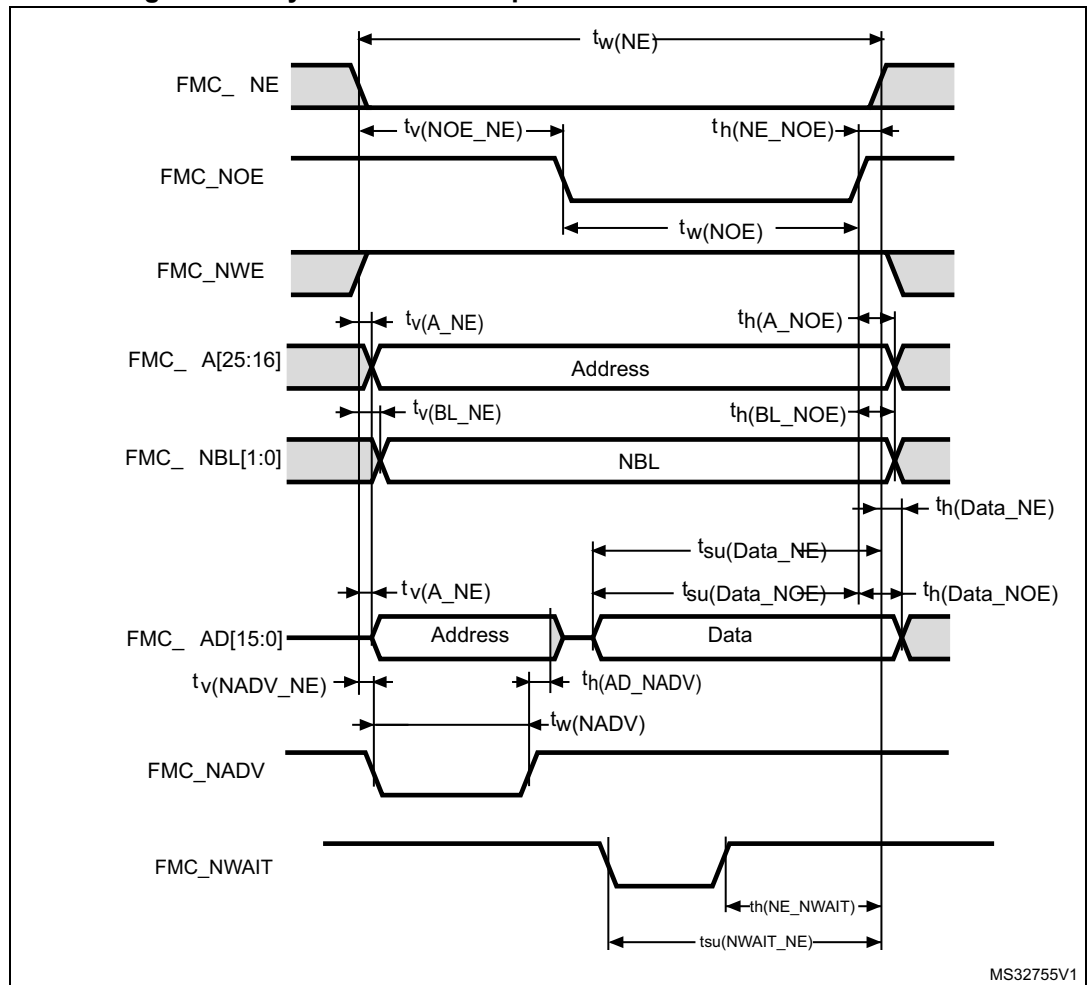


Table 96. Asynchronous multiplexed PSRAM/NOR read timings⁽¹⁾

Symbol	Parameter	Min	Max	Unit
$t_{w(NE)}$	FMC_NE low time	$4T_{HCLK}-0.5$	$4T_{HCLK}+1$	ns
$t_{v(NOE_NE)}$	FMC_NEx low to FMC_NOE low	$2T_{HCLK}-0.5$	$2T_{HCLK}+1$	
$t_{w(NOE)}$	FMC_NOE low time	$T_{HCLK}-0.5$	$T_{HCLK}+0.5$	
$t_{h(NE_NOE)}$	FMC_NOE high to FMC_NE high hold time	$T_{HCLK}-1$	-	
$t_{v(A_NE)}$	FMC_NEx low to FMC_A valid	-	3	
$t_{v(NADV_NE)}$	FMC_NEx low to FMC_NADV low	0.5	1.5	
$t_{w(NADV)}$	FMC_NADV low time	T_{HCLK}	$T_{HCLK}+1.5$	
$t_{h(AD_NADV)}$	FMC_AD(address) valid hold time after FMC_NADV high	$T_{HCLK}-3$	-	
$t_{h(A_NOE)}$	Address hold time after FMC_NOE high	Address held until next read operation	-	
$t_{su(Data_NE)}$	Data to FMC_NEx high setup time	$T_{HCLK}+14$	-	
$t_{su(Data_NOE)}$	Data to FMC_NOE high setup time	14	-	
$t_{h(Data_NE)}$	Data hold time after FMC_NEx high	0	-	
$t_{h(Data_NOE)}$	Data hold time after FMC_NOE high	0	-	

1. Guaranteed by characterization results.

Table 97. Asynchronous multiplexed PSRAM/NOR read-NWAIT timings⁽¹⁾

Symbol	Parameter	Min	Max	Unit
$t_{w(NE)}$	FMC_NE low time	$8T_{HCLK}-0.5$	$9T_{HCLK}+1$	ns
$t_{w(NOE)}$	FMC_NOE low time	$5T_{HCLK}-0.5$	$6T_{HCLK}+1$	
$t_{su(NWAIT_NE)}$	FMC_NWAIT valid before FMC_NEx high	$5T_{HCLK}+12$	-	
$t_{h(NE_NWAIT)}$	FMC_NEx hold time after FMC_NWAIT invalid	$4T_{HCLK}+11$	-	

1. Guaranteed by characterization results.

Figure 42. Asynchronous multiplexed PSRAM/NOR write waveforms

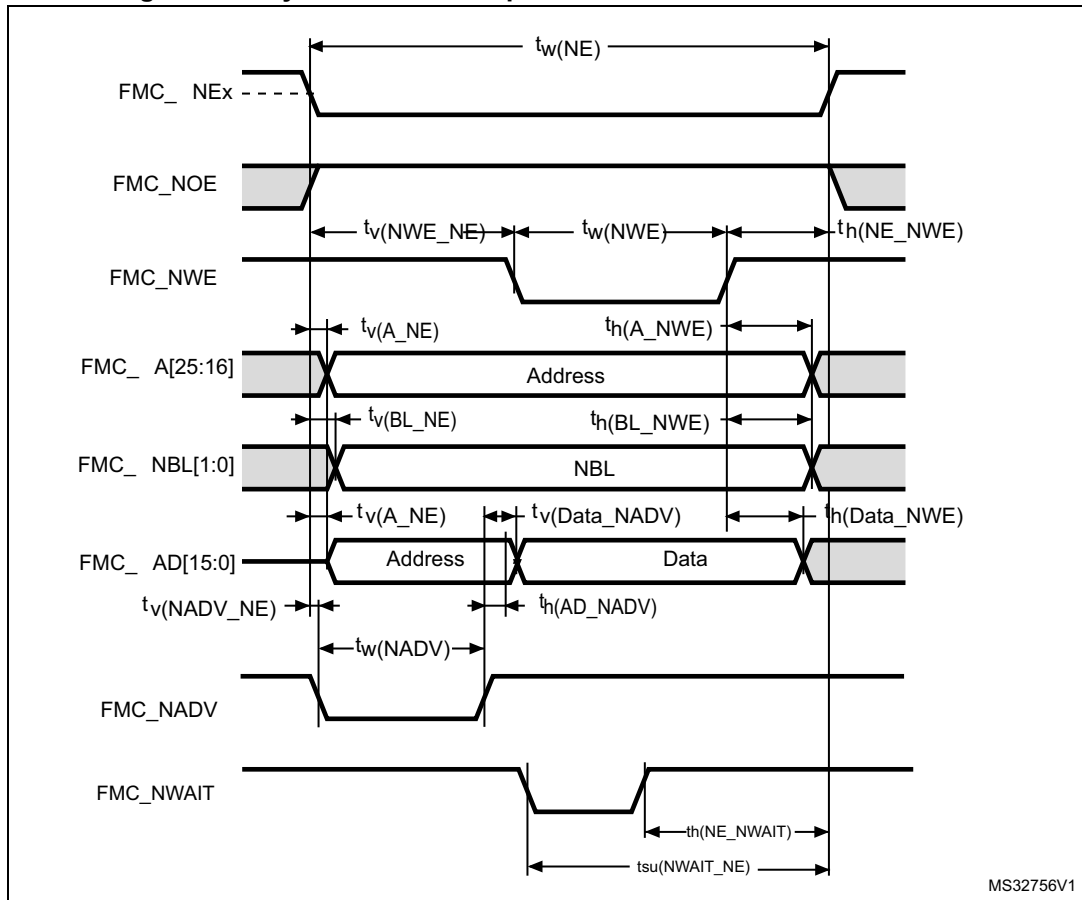


Table 98. Asynchronous multiplexed PSRAM/NOR write timings⁽¹⁾

Symbol	Parameter	Min	Max	Unit
$t_{w(NE)}$	FMC_NE low time	$5T_{HCLK}-0.5$	$5T_{HCLK}+1$	ns
$t_{v(NWE_NE)}$	FMC_NEx low to FMC_NWE low	$T_{HCLK}-0.5$	$T_{HCLK}+1$	
$t_{w(NWE)}$	FMC_NWE low time	$2T_{HCLK}-0.5$	$2T_{HCLK}+0.5$	
$t_{h(NE_NWE)}$	FMC_NWE high to FMC_NE high hold time	$2T_{HCLK}-0.5$	-	
$t_{v(A_NE)}$	FMC_NEx low to FMC_A valid	-	3	
$t_{v(NADV_NE)}$	FMC_NEx low to FMC_NADV low	0	1	
$t_{w(NADV)}$	FMC_NADV low time	$T_{HCLK}+0.5$	$T_{HCLK}+1.5$	
$t_{h(AD_NADV)}$	FMC_AD(address) valid hold time after FMC_NADV high	$T_{HCLK}-3$	-	
$t_{h(A_NWE)}$	Address hold time after FMC_NWE high	Address held until next write operation	-	
$t_{h(BL_NWE)}$	FMC_BL hold time after FMC_NWE high	$2T_{HCLK}-0.5$	-	
$t_{v(BL_NE)}$	FMC_NEx low to FMC_BL valid	-	T_{HCLK}	
$t_{v(Data_NADV)}$	FMC_NADV high to Data valid	-	$T_{HCLK}+2$	
$t_{h(Data_NWE)}$	Data hold time after FMC_NWE high	$2T_{HCLK}+0.5$	-	

1. Guaranteed by characterization results.

Table 99. Asynchronous multiplexed PSRAM/NOR write-NWAIT timings⁽¹⁾

Symbol	Parameter	Min	Max	Unit
$t_{w(NE)}$	FMC_NE low time	$10T_{HCLK}-0.5$	$10T_{HCLK}+1$	ns
$t_{w(NWE)}$	FMC_NWE low time	$7T_{HCLK}-0.5$	$7T_{HCLK}+0.5$	
$t_{su(NWAIT_NE)}$	FMC_NWAIT valid before FMC_NEx high	$7T_{HCLK}+12.5$	-	
$t_{h(NE_NWAIT)}$	FMC_NEx hold time after FMC_NWAIT invalid	$5T_{HCLK}+13$	-	

1. Guaranteed by characterization results.

Synchronous waveforms and timings

Figure 43 through Figure 46 represent synchronous waveforms and Table 100 through Table 103 provide the corresponding timings. The results shown in these tables are obtained with the following FMC configuration:

- BurstAccessMode = FMC_BurstAccessMode_Enable
- MemoryType = FMC_MemoryType_CRAM
- WriteBurst = FMC_WriteBurst_Enable
- CLKDivision = 1
- DataLatency = 1 for NOR Flash; DataLatency = 0 for PSRAM

In all timing tables, the T_{HCLK} is the HCLK clock period.

- For $2.7\text{ V} \leq V_{DD} \leq 3.6\text{ V}$, maximum FMC_CLK = 60 MHz for CLKDIV = 0x1 and 54 MHz for CLKDIV = 0x0 at CL = 30 pF (on FMC_CLK).
- For $1.71\text{ V} \leq V_{DD} \leq 2.7\text{ V}$, maximum FMC_CLK = 60 MHz for CLKDIV = 0x1 and 32 MHz for CLKDIV = 0x0 at CL = 20 pF (on FMC_CLK).

Figure 43. Synchronous multiplexed NOR/PSRAM read timings

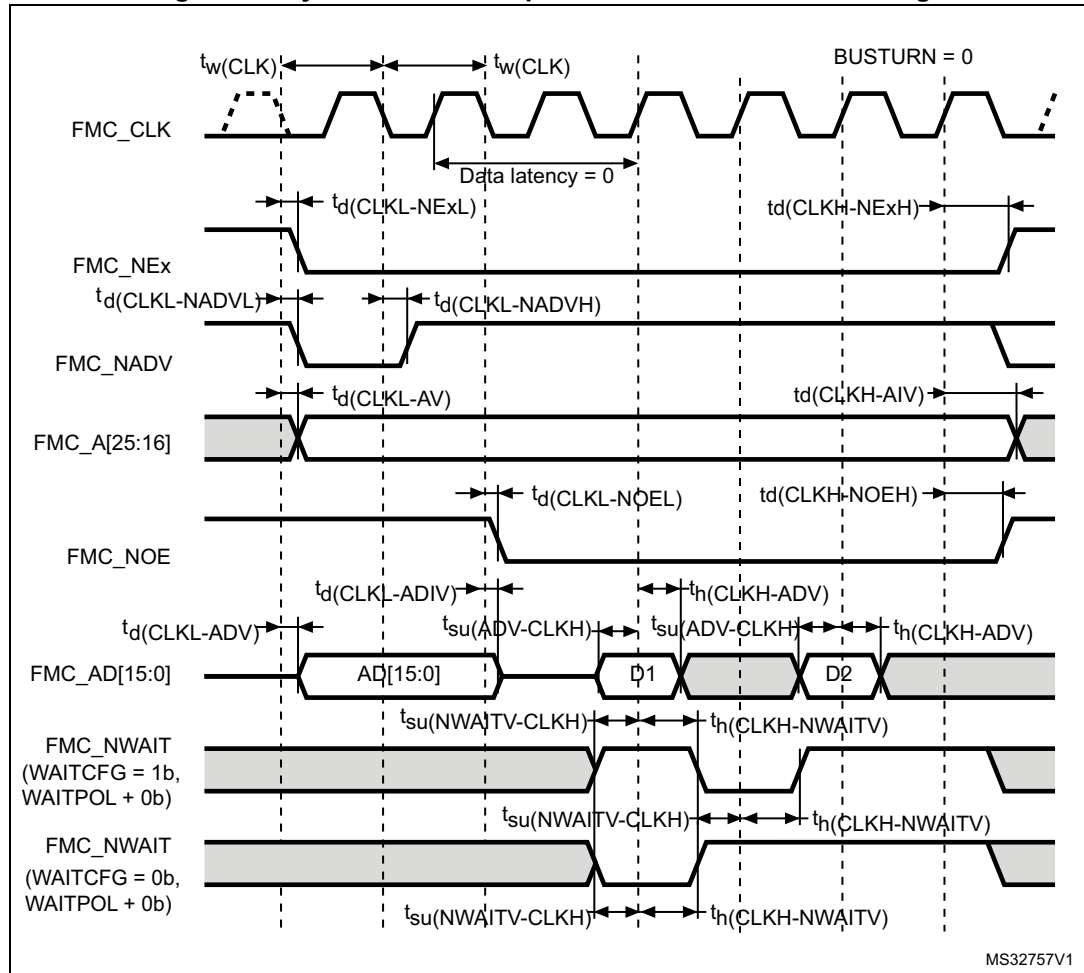


Table 100. Synchronous multiplexed NOR/PSRAM read timings⁽¹⁾⁽²⁾

Symbol	Parameter	Min	Max	Unit
$t_{w(\text{CLK})}$	FMC_CLK period	$RxT_{\text{HCLK}}-0.5^{(2)}$	-	ns
$t_{d(\text{CLKL-NExL})}$	FMC_CLK low to FMC_NEx low ($x=0..2$)	-	2.5	
$t_{d(\text{CLKH-NExH})}$	FMC_CLK high to FMC_NEx high ($x=0..2$)	$RxT_{\text{HCLK}}/2 + 1^{(2)}$	-	
$t_{d(\text{CLKL-NADV})}$	FMC_CLK low to FMC_NADV low	-	2.5	
$t_{d(\text{CLKL-NADVH})}$	FMC_CLK low to FMC_NADV high	2	-	
$t_{d(\text{CLKL-AV})}$	FMC_CLK low to FMC_Ax valid ($x=16..25$)	-	5.5	
$t_{d(\text{CLKH-AIV})}$	FMC_CLK high to FMC_Ax invalid ($x=16..25$)	$RxT_{\text{HCLK}}/2 + 1^{(2)}$	-	
$t_{d(\text{CLKL-NOEL})}$	FMC_CLK low to FMC_NOE low	-	2	
$t_{d(\text{CLKH-NOEH})}$	FMC_CLK high to FMC_NOE high	$RxT_{\text{HCLK}}/2 + 1^{(2)}$	-	
$t_{d(\text{CLKL-ADV})}$	FMC_CLK low to FMC_AD[15:0] valid	-	3	
$t_{d(\text{CLKL-ADIV})}$	FMC_CLK low to FMC_AD[15:0] invalid	0	-	
$t_{su(\text{ADV-CLKH})}$	FMC_A/D[15:0] valid data before FMC_CLK high	2	-	
$t_h(\text{CLKH-ADV})$	FMC_A/D[15:0] valid data after FMC_CLK high	4	-	
$t_{su(\text{NWAIT-CLKH})}$	FMC_NWAIT valid before FMC_CLK high	1.5	-	
$t_h(\text{CLKH-NWAIT})$	FMC_NWAIT valid after FMC_CLK high	4	-	

1. Guaranteed by characterization results.
2. Clock ratio $R = (\text{HCLK period} / \text{FMC_CLK period})$.

Figure 44. Synchronous multiplexed PSRAM write timings

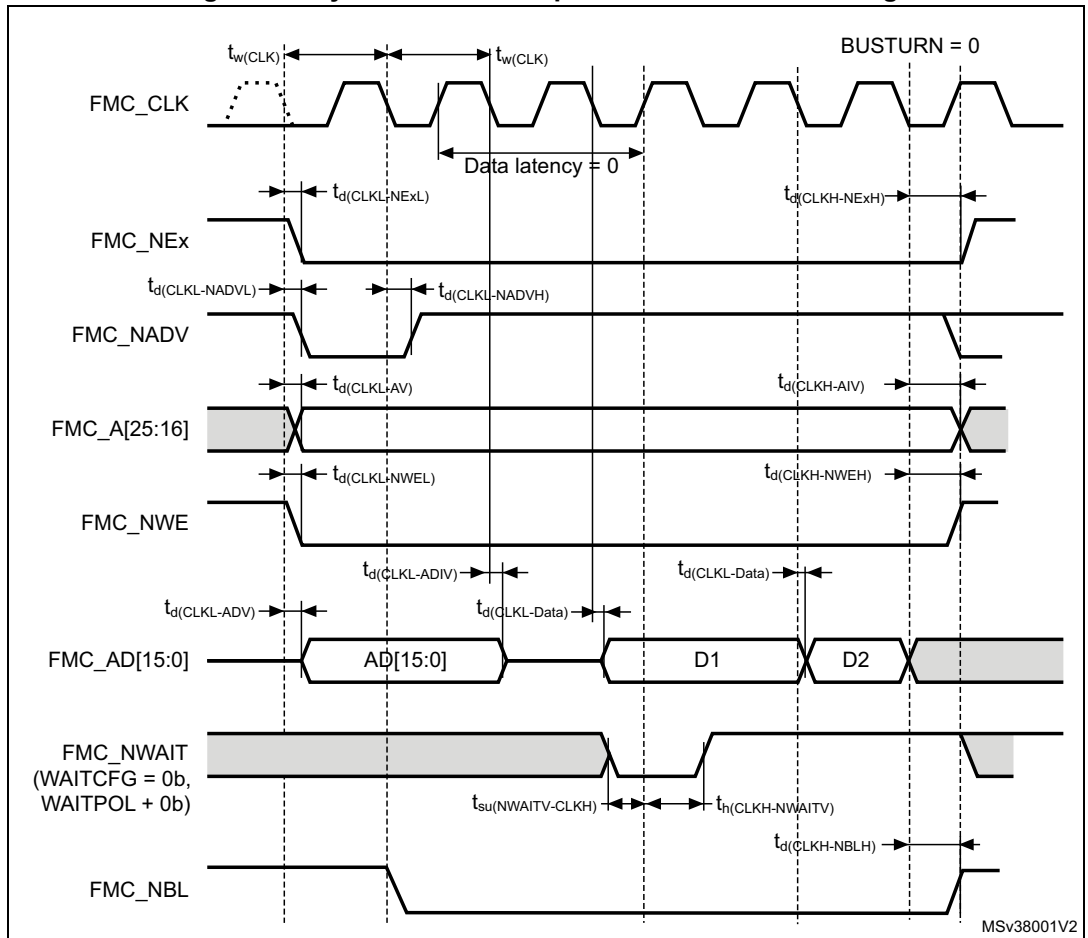


Table 101. Synchronous multiplexed PSRAM write timings⁽¹⁾⁽²⁾

Symbol	Parameter	Min	Max	Unit
$t_{w(CLK)}$	FMC_CLK period , VDD range= 2.7 to 3.6 V	$RxT_{HCLK} \cdot 0.5^{(2)}$	-	ns
$t_{d(CLKL-NExL)}$	FMC_CLK low to FMC_NEx low (x=0..2)	-	2.5	
$t_{d(CLKH-NExH)}$	FMC_CLK high to FMC_NEx high (x= 0...2)	$RxT_{HCLK}/2 + 1^{(2)}$	-	
$t_{d(CLKL-NADVl)}$	FMC_CLK low to FMC_NADV low	-	2.5	
$t_{d(CLKL-NADVh)}$	FMC_CLK low to FMC_NADV high	2	-	
$t_{d(CLKL-AV)}$	FMC_CLK low to FMC_Ax valid (x=16...25)	-	5.5	
$t_{d(CLKH-AIV)}$	FMC_CLK high to FMC_Ax invalid (x=16...25)	$RxT_{HCLK}/2 + 1^{(2)}$	-	
$t_{d(CLKL-NWEL)}$	FMC_CLK low to FMC_NWE low	-	2	
$t_{d(CLKH-NWEH)}$	FMC_CLK high to FMC_NWE high	$RxT_{HCLK}/2 + 1^{(2)}$	-	
$t_{d(CLKL-ADV)}$	FMC_CLK low to FMC_AD[15:0] valid	-	3	
$t_{d(CLKL-ADIV)}$	FMC_CLK low to FMC_AD[15:0] invalid	0	-	
$t_{d(CLKL-DATA)}$	FMC_A/D[15:0] valid data after FMC_CLK low	-	3.5	
$t_{d(CLKL-NBLL)}$	FMC_CLK low to FMC_NBL low	1	-	
$t_{d(CLKH-NBLH)}$	FMC_CLK high to FMC_NBL high	$RxT_{HCLK}/2 + 1.5^{(2)}$	-	
$t_{su(NWAIT-CLKH)}$	FMC_NWAIT valid before FMC_CLK high	1.5	-	
$t_h(CLKH-NWAIT)$	FMC_NWAIT valid after FMC_CLK high	4	-	

1. Guaranteed by characterization results.
2. Clock ratio R = (HCLK period /FMC_CLK period).

Figure 45. Synchronous non-multiplexed NOR/PSRAM read timings

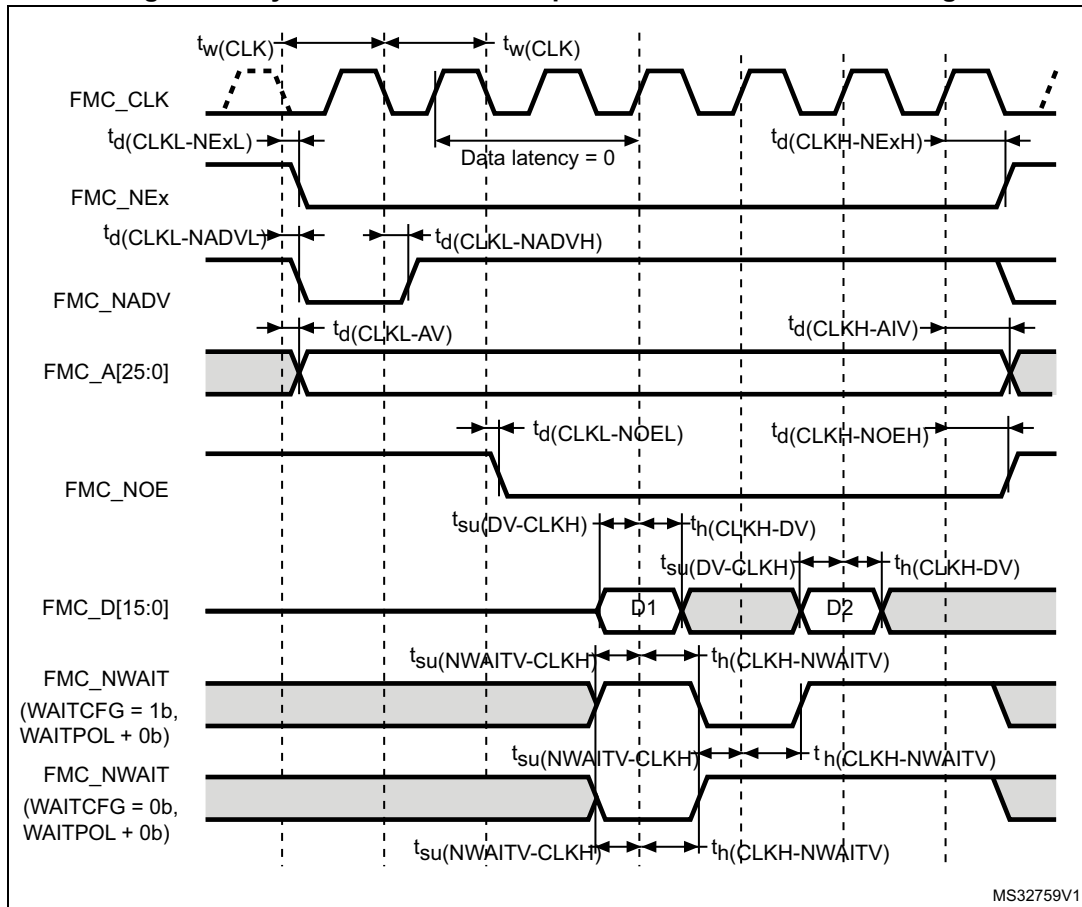


Table 102. Synchronous non-multiplexed NOR/PSRAM read timings⁽¹⁾⁽²⁾

Symbol	Parameter	Min	Max	Unit
$t_w(\text{CLK})$	FMC_CLK period	$RxT_{HCLK} - 0.5^{(2)}$	-	ns
$t_d(\text{CLKL-NExL})$	FMC_CLK low to FMC_NEx low ($x=0..2$)	-	2.5	
$t_d(\text{CLKH-NExH})$	FMC_CLK high to FMC_NEx high ($x=0..2$)	$RxT_{HCLK}/2 + 1^{(2)}$	-	
$t_d(\text{CLKL-NADV})$	FMC_CLK low to FMC_NADV low	-	2.5	
$t_d(\text{CLKL-NADVH})$	FMC_CLK low to FMC_NADV high	2	-	
$t_d(\text{CLKL-AV})$	FMC_CLK low to FMC_Ax valid ($x=16..25$)	-	5.5	
$t_d(\text{CLKH-AIV})$	FMC_CLK high to FMC_Ax invalid ($x=16..25$)	$RxT_{HCLK}/2 + 0.5^{(2)}$	-	
$t_d(\text{CLKL-NOEL})$	FMC_CLK low to FMC_NOE low	-	2	
$t_d(\text{CLKH-NOEH})$	FMC_CLK high to FMC_NOE high	$RxT_{HCLK}/2 + 1^{(2)}$	-	
$t_{su}(\text{DV-CLKH})$	FMC_D[15:0] valid data before FMC_CLK high	2	-	
$t_h(\text{CLKH-DV})$	FMC_D[15:0] valid data after FMC_CLK high	4	-	

Table 102. Synchronous non-multiplexed NOR/PSRAM read timings⁽¹⁾⁽²⁾ (continued)

Symbol	Parameter	Min	Max	Unit
$t_{su}(NWAIT-CLKH)$	FMC_NWAIT valid before FMC_CLK high	1.5	-	ns
$t_h(CLKH-NWAIT)$	FMC_NWAIT valid after FMC_CLK high	4	-	

1. Guaranteed by characterization results.
2. Clock ratio R = (HCLK period / FMC_CLK period).

Figure 46. Synchronous non-multiplexed PSRAM write timings

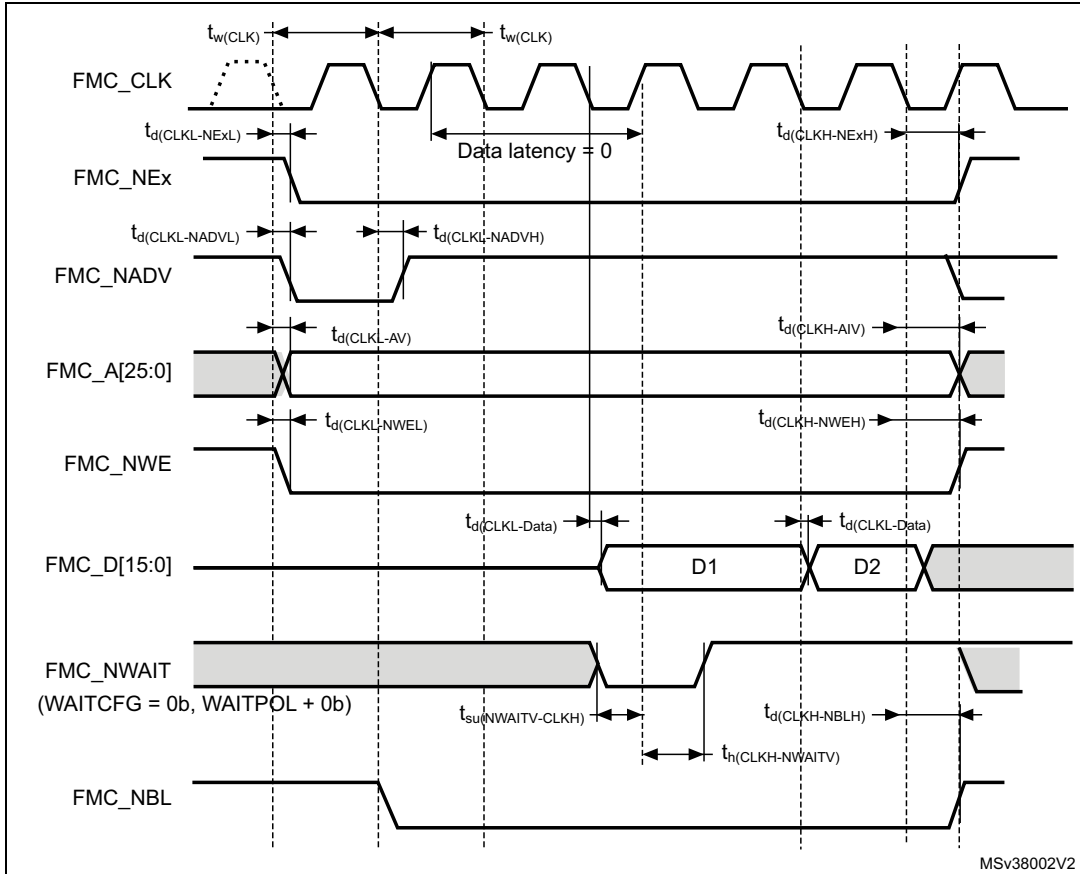


Table 103. Synchronous non-multiplexed PSRAM write timings⁽¹⁾⁽²⁾

Symbol	Parameter	Min	Max	Unit
$t_{w(\text{CLK})}$	FMC_CLK period	$RxT_{\text{HCLK}}-0.5^{(2)}$	-	ns
$t_{d(\text{CLKL-NEXL})}$	FMC_CLK low to FMC_NEx low ($x=0..2$)	-	2.5	
$t_{d(\text{CLKH-NEXH})}$	FMC_CLK high to FMC_NEx high ($x=0..2$)	$RxT_{\text{HCLK}}/2 + 1^{(2)}$	-	
$t_{d(\text{CLKL-NADVL})}$	FMC_CLK low to FMC_NADV low	-	2.5	
$t_{d(\text{CLKL-NADVH})}$	FMC_CLK low to FMC_NADV high	2	-	
$t_{d(\text{CLKL-AV})}$	FMC_CLK low to FMC_Ax valid ($x=16..25$)	-	5.5	
$t_{d(\text{CLKH-AIV})}$	FMC_CLK high to FMC_Ax invalid ($x=16..25$)	$RxT_{\text{HCLK}}/2 + 0.5^{(2)}$	-	
$t_{d(\text{CLKL-NWEL})}$	FMC_CLK low to FMC_NWE low	-	2	
$t_{d(\text{CLKH-NWEH})}$	FMC_CLK high to FMC_NWE high	$RxT_{\text{HCLK}}/2 + 1^{(2)}$	-	
$t_{d(\text{CLKL-Data})}$	FMC_D[15:0] valid data after FMC_CLK low	-	3.5	
$t_{d(\text{CLKL-NBLL})}$	FMC_CLK low to FMC_NBL low	1	-	
$t_{d(\text{CLKH-NBLH})}$	FMC_CLK high to FMC_NBL high	$RxT_{\text{HCLK}}/2 + 1.5^{(2)}$	-	
$t_{su(\text{NWAIT-CLKH})}$	FMC_NWAIT valid before FMC_CLK high	1.5	-	
$t_h(\text{CLKH-NWAIT})$	FMC_NWAIT valid after FMC_CLK high	4	-	

1. Guaranteed by characterization results.

2. Clock ratio $R = (\text{HCLK period} / \text{FMC_CLK period})$.

NAND controller waveforms and timings

Figure 47 through Figure 50 represent synchronous waveforms, and Table 104 and Table 105 provide the corresponding timings. The results shown in these tables are obtained with the following FMC configuration:

- COM.FMC_SetupTime = 0x01
- COM.FMC_WaitSetupTime = 0x03
- COM.FMC_HoldSetupTime = 0x02
- COM.FMC_HiZSetupTime = 0x01
- ATT.FMC_SetupTime = 0x01
- ATT.FMC_WaitSetupTime = 0x03
- ATT.FMC_HoldSetupTime = 0x02
- ATT.FMC_HiZSetupTime = 0x01
- Bank = FMC_Bank_NAND
- MemoryDataWidth = FMC_MemoryDataWidth_16b
- ECC = FMC_ECC_Enable
- ECCPageSize = FMC_ECCPageSize_512Bytes
- TCLRSetupTime = 0
- TARSetupTime = 0

In all timing tables, the T_{HCLK} is the HCLK clock period.

Figure 47. NAND controller waveforms for read access

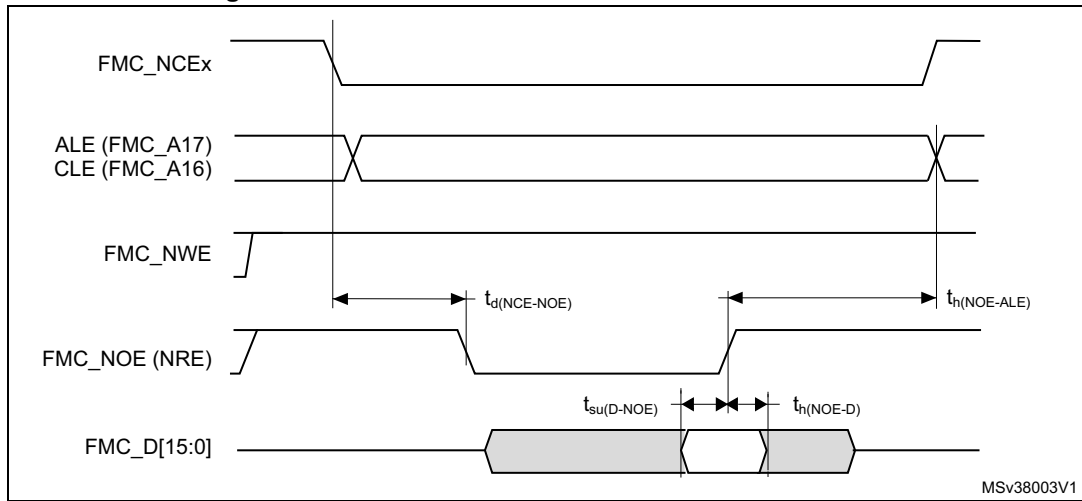


Figure 48. NAND controller waveforms for write access

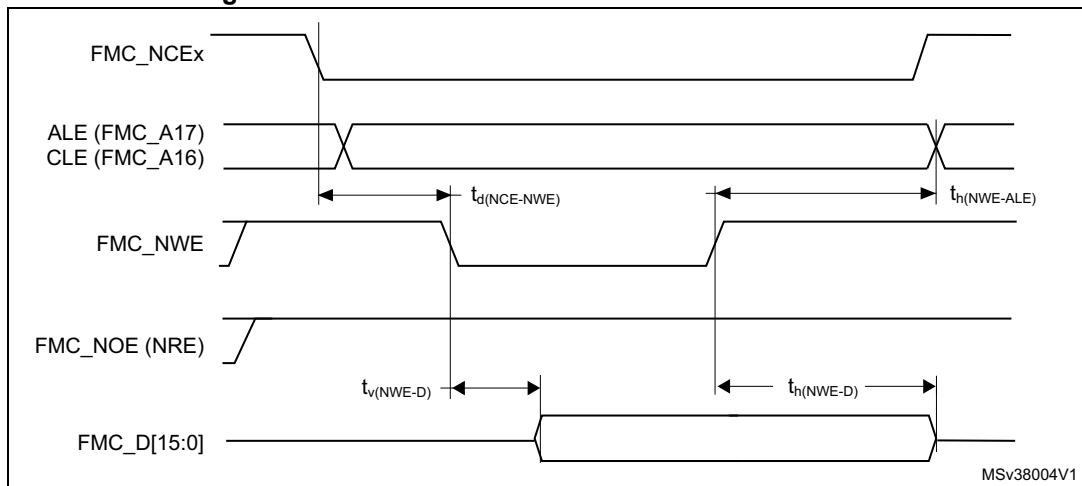


Figure 49. NAND controller waveforms for common memory read access

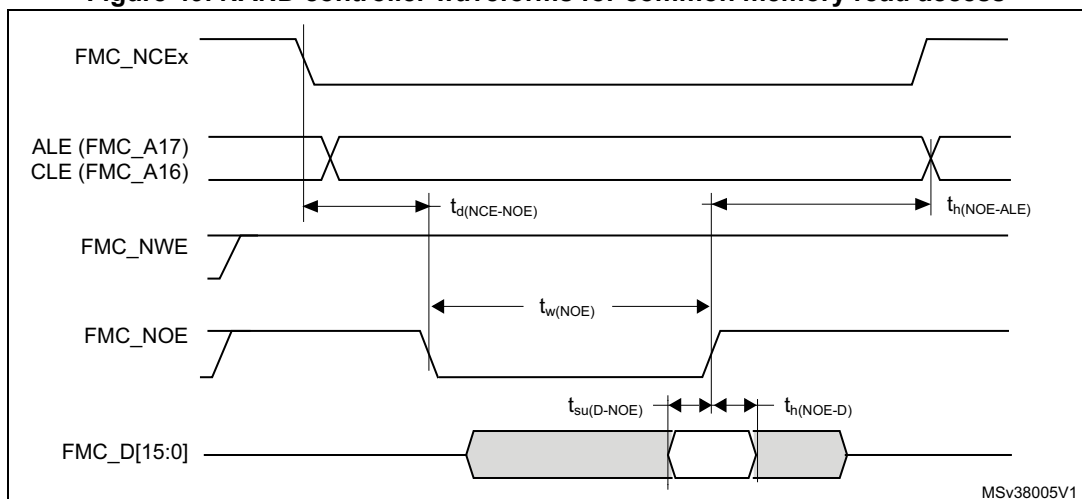


Figure 50. NAND controller waveforms for common memory write access

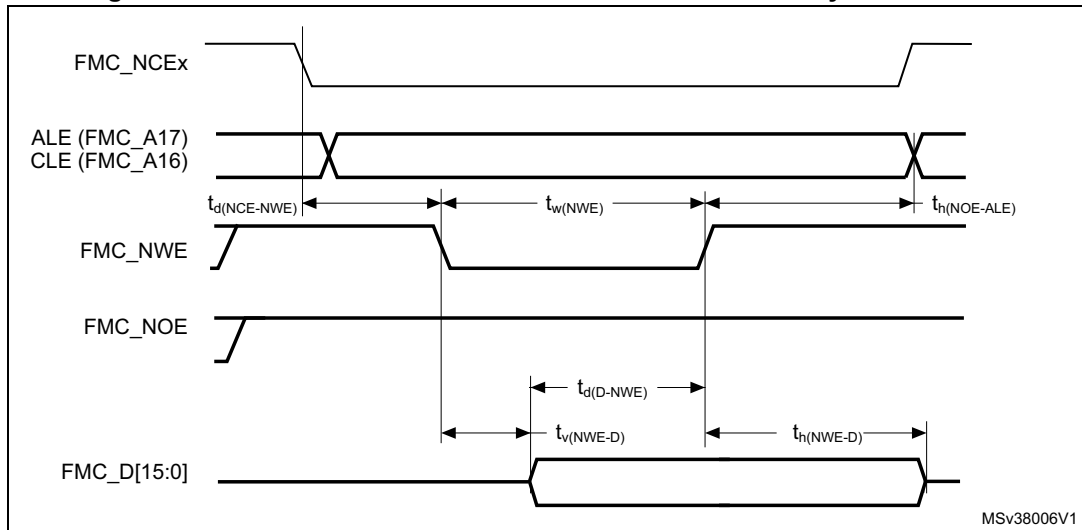


Table 104. Switching characteristics for NAND Flash read cycles⁽¹⁾

Symbol	Parameter	Min	Max	Unit
$T_{w(NOE)}$	FMC_NOE low width	$4T_{HCLK}-0.5$	$4T_{HCLK}+0.5$	ns
$T_{su(D-NOE)}$	FMC_D[15-0] valid data before FMC_NOE high	14	-	
$T_{h(NOE-D)}$	FMC_D[15-0] valid data after FMC_NOE high	0	-	
$T_{d(ALE-NOE)}$	FMC_ALE valid before FMC_NOE low	-	$3T_{HCLK}-1$	
$T_{h(NOE-ALE)}$	FMC_NOE high to FMC_ALE invalid	$3T_{HCLK}-0.5$	-	

1. Guaranteed by characterization results.

Table 105. Switching characteristics for NAND Flash write cycles⁽¹⁾

Symbol	Parameter	Min	Max	Unit
$T_{w(NWE)}$	FMC_NWE low width	$4T_{HCLK}-0.5$	$4T_{HCLK}+0.5$	ns
$T_{v(NWE-D)}$	FMC_NWE low to FMC_D[15-0] valid	5	-	
$T_{h(NWE-D)}$	FMC_NWE high to FMC_D[15-0] invalid	$2T_{HCLK}$	-	
$T_{d(D-NWE)}$	FMC_D[15-0] valid before FMC_NWE high	$5T_{HCLK}-1$	-	
$T_{d(ALE-NWE)}$	FMC_ALE valid before FMC_NWE low	-	$3T_{HCLK}-1$	
$T_{h(NWE-ALE)}$	FMC_NWE high to FMC_ALE invalid	$3T_{HCLK}-0.5$	-	

1. Guaranteed by characterization results.

6.3.29 OCTOSPI characteristics

Unless otherwise specified, the parameters given in [Table 106](#) and [Table 107](#) for OCTOSPI are derived from tests performed under the ambient temperature, f_{AHB} frequency and V_{DD} supply voltage conditions summarized in [Table 21: General operating conditions](#), with the following configuration:

- Output speed is set to $\text{OSPEEDRy}[1:0] = 11$
- Measurement points are done at CMOS levels: $0.5 \times V_{\text{DD}}$

Refer to [Section 6.3.14: I/O port characteristics](#) for more details on the input/output alternate function characteristics.

Table 106. OCTOSPI⁽¹⁾ characteristics in SDR mode⁽²⁾

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
F(QCK)	OCTOSPI clock frequency	1.71 V < V_{DD} < 3.6 V Voltage Range V1 $C_{\text{LOAD}} = 20$ pF	-	-	54	MHz
		2.7 V < V_{DD} < 3.6 V Voltage Range V1 $C_{\text{LOAD}} = 20$ pF	-	-	92	
		1.71 V < V_{DD} < 3.6 V Voltage Range V1 $C_{\text{LOAD}} = 15$ pF	-	-	58	
		1.71 V < V_{DD} < 3.6 V Voltage Range 2 $C_{\text{LOAD}} = 20$ pF	-	-	26	
$t_{\text{w(CKH)}}$	OCTOSPI clock high and low time Even division	Prescaler[7:0] = $n = 0, 1, 3, 5$	$t_{\text{CKY}}/2 - 0.5$	-	$t_{\text{CKY}}/2$	ns
$t_{\text{w(CKL)}}$			$t_{\text{CKY}}/2 - 0.5$	-	$t_{\text{CKY}}/2$	
$t_{\text{w(CKH)}}$	OCTOSPI clock high and low time Odd division	Prescaler[7:0] = $n = 2, 4, 6, 8$	$(n/2) * t_{\text{CK}} / (n+1) - 0.5$	-	$(n/2) * t_{\text{CK}} / (n+1)$	
$t_{\text{w(CKL)}}$			$(n/2+1) * t_{\text{CK}} / (n+1) - 0.5$	-	$(n/2+1) * t_{\text{CK}} / (n+1)$	
$t_{\text{s(IN)}}$	Data input setup time	Voltage Range V1	1	-	-	
		Voltage Range V2	0.5	-	-	
$t_{\text{h(IN)}}$	Data input hold time	Voltage Range V1	4.5	-	-	
		Voltage Range V2	9.5	-	-	
$t_{\text{v(OUT)}}$	Data output valid time	Voltage Range V1	-	1	3	
		Voltage Range V2	-	2	5	
$t_{\text{h(OUT)}}$	Data output hold time	Voltage Range V1	0	-	-	
		Voltage Range V2	0	-	-	

1. Values in the table applies to Octal and Quad SPI mode.

2. Guaranteed by characterization results.

Table 107. OCTOSPI characteristics in DTR mode (with DQS)⁽¹⁾/Octal and Hyperbus

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
F_{CK} $1/t_{(CK)}$	OCTOSPI clock frequency	1.71 V < V_{DD} < 3.6 V Voltage Range V1 $C_{LOAD} = 20$ pF	-	-	60 ⁽²⁾	MHz
		2.7 V < V_{DD} < 3.6 V Voltage Range V1 $C_{LOAD} = 20$ pF	-	-	90 ⁽²⁾⁽³⁾	
		1.71 V < V_{DD} < 3.6 V Voltage Range V1 $C_{LOAD} = 15$ pF	-	-	66 ⁽³⁾	
		1.71 V < V_{DD} < 3.6 V Voltage Range V2 $C_{LOAD} = 20$ pF	-	-	26 ⁽²⁾	
$t_{w(CKH)}$	OCTOSPI clock high and low time Even division	-	$t_{(CK)}/2-1$	-	$t_{(CK)}/2+0.5$	ns
$t_{w(CKL)}$			$t_{(CK)}/2-0.5$	-	$t_{(CK)}/2+0.5$	
$t_{w(CKH)}$	OCTOSPI clock high and low time Odd division	-	$(n/2)*t_{(CK)}/(n+1)-0.5$	-	$(n/2)*t_{(CK)}/(n+1)+0.5$	
$t_{w(CKL)}$			$(n/2+1)*t_{(CK)}/(n+1)-0.5$	-	$(n/2+1)*t_{(CK)}/(n+1)+0.5$	
$t_{v(CK)}$	Clock valid time	-	-	-	$t_{(CK)} + 2$	
$t_{h(CK)}$	Clock hold time	-	$t_{(CK)}/2 - 0.5$	-	-	
$V_{ODr(CK)}$ ⁽⁴⁾	CK,CK# crossing level on CK rising edge	$V_{DD}=1v8$	730	-	997	mV
$V_{ODf(CK)}$ ⁽⁴⁾	CK,CK# crossing level on CK falling edge	$V_{DD}=1v8$	788	-	1070	
$t_{w(CS)}$	Chip select high time	-	$3*t_{(CK)}$	-	-	ns
$t_{v(DQ)}$	Data input valid time	-	0	-	-	
$t_{v(DS)}$	Data storbe input valid time					
$t_{h(DS)}$	Data storbe input hold time					
$t_{v(RWDS)}$	Data storbe output valid time	-	-	-	$3*t_{(CK)}$	
$t_{sr(DQ)}, t_{sf(DQ)}$	Data input setup time	Voltage Range V1	-0.75	-	$t_{(CK)}/2 - 3.75$ ⁽⁵⁾	
		Voltage Range V2	-1	-	$t_{(CK)}/2 - 5$ ⁽⁵⁾	
$t_{hr(DQ)}, t_{hf(DQ)}$	Data input hold time	Voltage Range V1	3.75	-	-	
		Voltage Range V2	5	-	-	

Table 107. OCTOSPI characteristics in DTR mode (with DQS)⁽¹⁾/Octal and Hyperbus (continued)

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$t_{vr(OUT)}$ $t_{vf(OUT)}$	Data output valid time	Voltage Range V1 DHQC =0 DHQC =1 Pres=1,2 ...	-	5.75	7.75	ns
			-	$T_{pclk}/4 + 0.75$	$T_{pclk}/4 + 1.5$	
$t_{hr(OUT)}$ $t_{hf(OUT)}$	Data output hold time	Voltage Range V1 DHQC =0 DHQC =1 Pres=1,2 ...	0.5	-	-	
			$T_{pclk}/4 - 0.25$	-	-	
		Voltage Range V2 DHQC=0	5	-	-	

1. Guaranteed by characterization results.
2. Maximum frequency values are given for an RWDS to DQ skew of maximum +/-1.0ns
3. Activating DHQC is mandatory to reach this frequency.
4. All clk/clk# pairs are in line with HyperMemory AC differential crossing voltage margins except pairs including PB10 and PH7.
5. Data input setup time maximum does not take into account data level switching duration.

Figure 51. OCTOSPI timing diagram - SDR mode

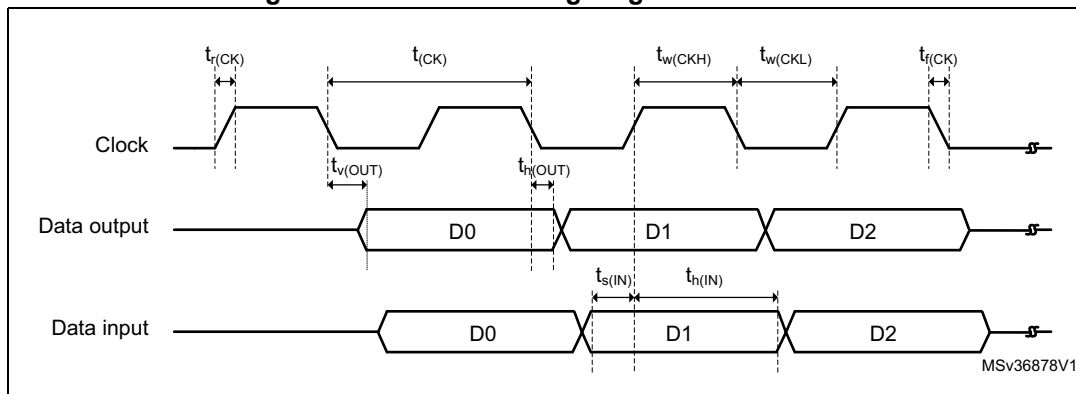


Figure 52. OCTOSPI timing diagram - DDR mode

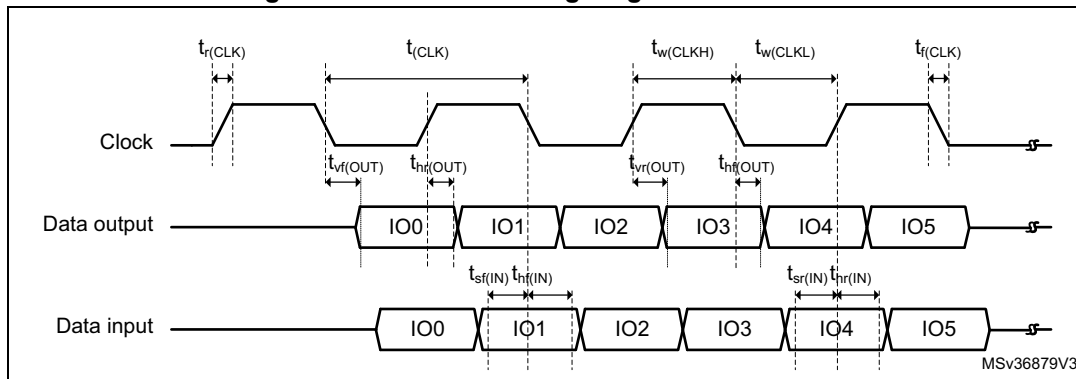


Figure 53. OCTOSPI Hyperbus clock

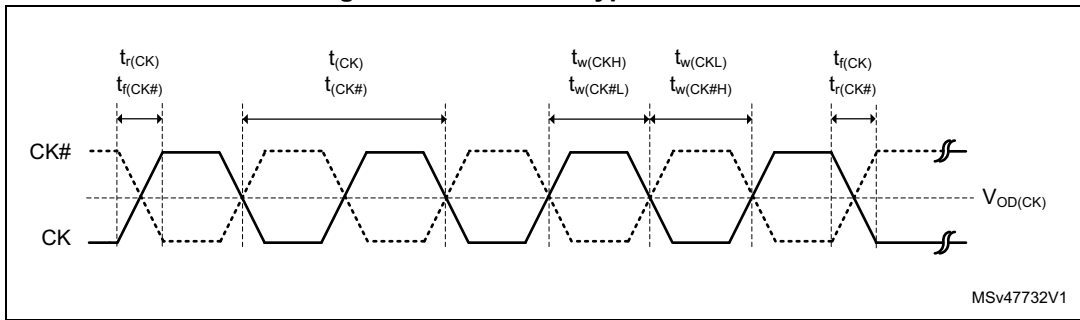


Figure 54. OCTOSPI Hyperbus read

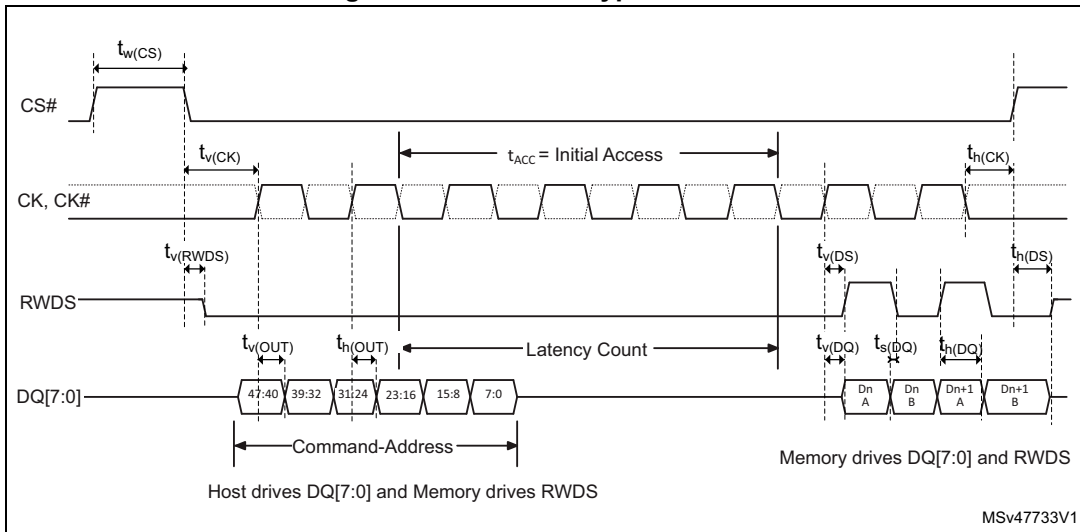


Figure 55. OCTOSPI Hyperbus read with double latency

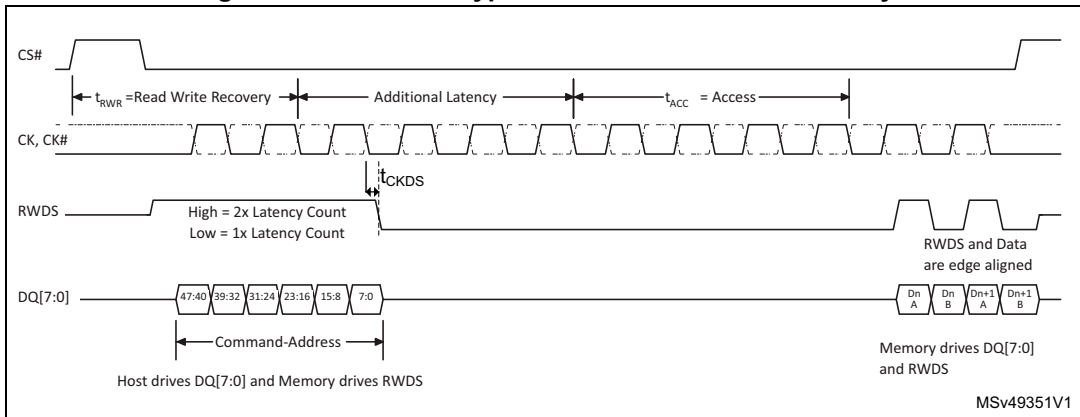
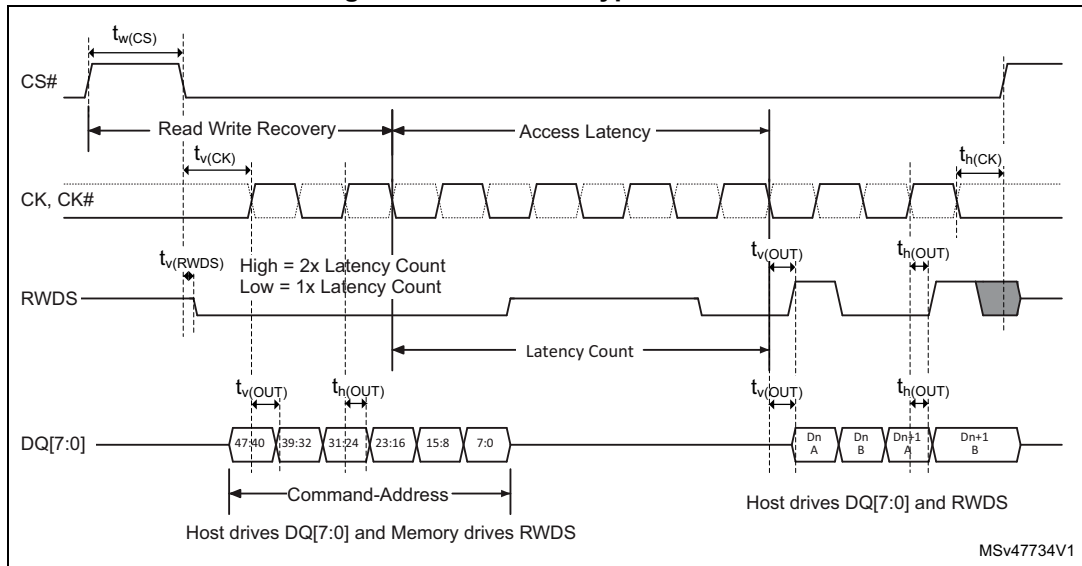


Figure 56. OCTOSPI Hyperbus write



Delay block

Unless otherwise specified, the parameters given in [Table 108](#) for delay block are derived from tests performed under the ambient temperature, f_{PCLKx} frequency and VDD supply voltage conditions summarized in [Table 22: Operating conditions at power-up / power-down](#) with the configuration shown in the table below.

Table 108. Dynamics characteristics: Delay Block characteristics

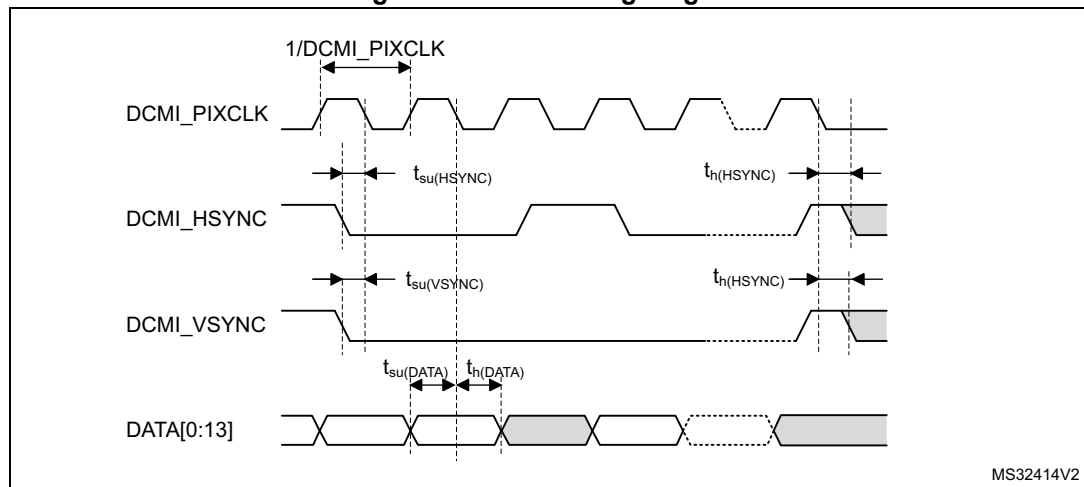
Symbol	Parameter	Conditions	Min	Typ	Max	Unit
t_{init}	Initial delay	-	700	875	1000	ps
t_{Δ}	Unit Delay	-	375	500	750	

6.3.30 Camera interface (DCMI) timing specifications

Unless otherwise specified, the parameters given in [Table 109](#) for DCMI are derived from tests performed under the ambient temperature, f_{HCLK} frequency and V_{DD} supply voltage summarized in [Table 20](#), with the following configuration:

- DCMI_PIXCLK polarity: falling
- DCMI_VSYNC and DCMI_HSYNC polarity: high
- Data format: 14 bits
- Capacitive load $C = 30 \text{ pF}$
- Measurement points are done at CMOS levels: $0.5 \times V_{DD}$

Figure 57. DCMI timing diagram



MS32414V2

Table 109. DCMI characteristics⁽¹⁾

Symbol	Parameter	Condition	Min	Max	Unit
-	Frequency ratio DCMI_PIXCLK/ f_{HCLK}	-	-	0.4	-
DCMI_PIXCLK	Pixel clock input	1.71 < VDD < 3.6 Voltage range V1	-	48	MHz
		1.71 < VDD < 3.6 Voltage range V2	-	10	
D _{pixel}	Pixel clock input duty cycle	-	30	70	%

Table 109. DCMI characteristics⁽¹⁾ (continued)

Symbol	Parameter	Condition	Min	Max	Unit
t _{su} (DATA)	Data input setup time	1.71 < VDD < 3.6 Voltage range V1	3	-	ns
		1.71 < VDD < 3.6 Voltage range V2	4	-	
t _h (DATA)	Data hold time	1.71 < VDD < 3.6 Voltage range V1	0.5	-	
		1.71 < VDD < 3.6 Voltage range V2	0.5	-	
t _{su} (HSYNC), t _{su} (VSYNC)	DCMI_HSYNC/DCMI_VSYNC input setup time	1.71 < VDD < 3.6 Voltage range V1	2	-	
		1.71 < VDD < 3.6 Voltage range V2	3	-	
t _h (HSYNC), t _h (VSYNC)	DCMI_HSYNC/DCMI_VSYNC input hold time	1.71 < VDD < 3.6 Voltage range V1	0.5	-	
		1.71 < VDD < 3.6 Voltage range V2	0.5	-	

1. Guaranteed by characterization results.

6.3.31 Parallel slave interface (PSSI)

Unless otherwise specified, the parameters given in [Table 110](#) for PSSI are derived from tests performed under the ambient temperature, fHCLK frequency and VDD supply voltage summarized in [Table 21](#), with the following configuration:

- PSSI_PDCK polarity: falling
- PSSI_RDY and PSSI_DE polarity: low
- Data format: 16 bits
- Output speed is set to OSPEEDRy[1:0] = 11
- Capacitive load C = 30 pF
- Measurement points are done at CMOS levels: 0.5 x V_{DD}

Table 110. PSSI receive characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min	Max	Unit
-	Frequency ratio PSSI_PDCK/f _{HCLK}	-	-	0.4	-
PSSI_PDCK	PSSI Clock input	1.71<VDD<3.6 Voltage Range V1	-	48	MHz
		1.71<VDD<3.6 Voltage Range V2	-	10	
D _{pixel}	PSSI Clock input duty cycle	-	30	70	%
t _{su} (DATA)	Data input setup time	1.71<VDD<3.6 Voltage Range V1	4.5	-	ns
		1.71<VDD<3.6 Voltage Range V2	6.5	-	
t _h (DATA)	Data input hold time	1.71<VDD<3.6 Voltage Range V1	0.5	-	
		1.71<VDD<3.6 Voltage Range V2	0.5	-	
t _{su} (DE)	DE input setup time	1.71<VDD<3.6 Voltage Range V1	1.5	-	
		1.71<VDD<3.6 Voltage Range V2	2	-	
t _h (DE)	DE input hold time	1.71<VDD<3.6 Voltage Range V1	0.5	-	
		1.71<VDD<3.6 Voltage Range V2	0.5	-	
tov(RDY)	RDY output valid time	1.71<VDD<3.6 Voltage Range V1	-	18	
		1.71<VDD<3.6 Voltage Range V2	-	21.5	
toh(RDY)	RDY output hold time	1.71<VDD<3.6 Voltage Range V1	7	-	
		1.71<VDD<3.6 Voltage Range V2	10.5	-	

1. Guaranteed by characterization results.

Table 111. PSSI transmit characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min	Max	Unit
-	Frequency ratio PSSI_PDCK/f _{HCLK}	-	-	0.4	-
PSSI_PDCK	PSSI Clock input	1.71<VDD<3.6 Voltage Range V1	-	25	MHz
		2.7<VDD<3.6 Voltage Range V1	-	37	
		1.71<VDD<3.6 Voltage Range V2	-	10	
D _{pixel}	PSSI Clock input duty cycle	-	30	70	%
t _{ov} (DATA)	Data output valid time	1.71<VDD<3.6 Voltage Range V1	-	18	ns
		2.7<VDD<3.6 Voltage Range V1	-	13.5	
		1.71<VDD<3.6 Voltage Range V2	-	20	
t _{oh} (DATA)	Data output hold time	1.71<VDD<3.6 Voltage Range V1	4.5	-	
		1.71<VDD<3.6 Voltage Range V2	6	-	
t _{ov} (DE)	DE output valid time	1.71<VDD<3.6 Voltage Range V1	-	16.5	
		2.7<VDD<3.6 Voltage Range V1	-	12.5	
		1.71<VDD<3.6 Voltage Range V2	-	19.5	
t _{oh} (DE)	DE output hold time	1.71<VDD<3.6 Voltage Range V1	6	-	
		1.71<VDD<3.6 Voltage Range V2	9	-	
tsu(RDY)	RDY input setup time	-	0	-	
th(RDY)	RDY input hold time	-	0	-	

1. Guaranteed by characterization results.

Figure 58. PSSI receive timing diagram

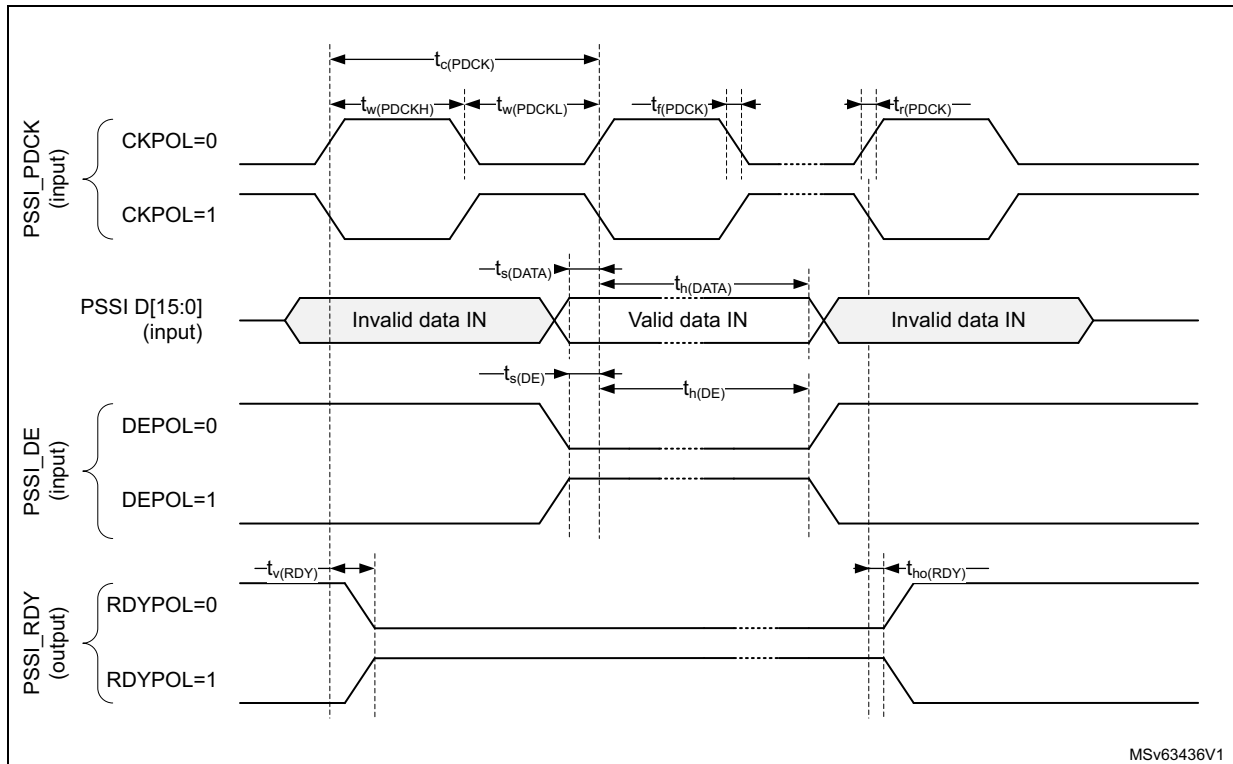
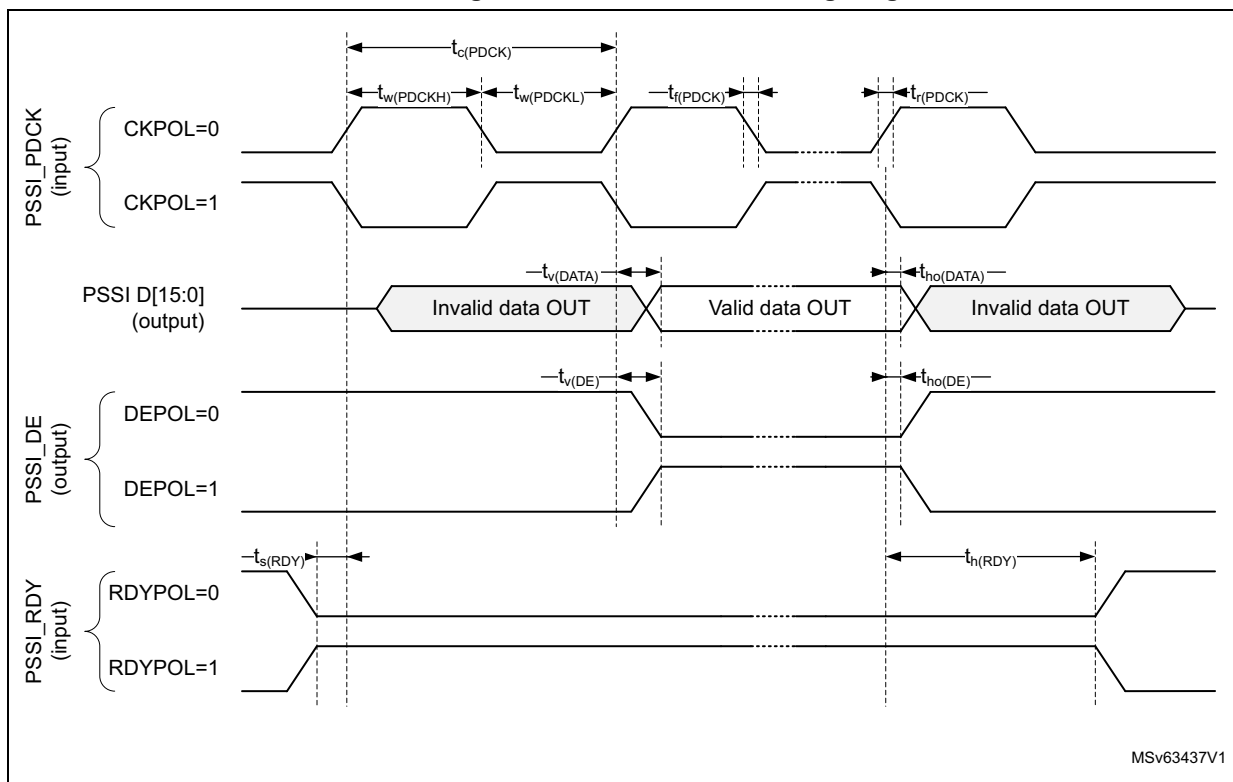


Figure 59. PSSI transmit timing diagram



6.3.32 LCD-TFT controller (LTDC) characteristics

Unless otherwise specified, the parameters given in [Table 112](#) for LCD-TFT are derived from tests performed under the ambient temperature, fHCLK frequency and VDD supply voltage summarized in [Table 21](#), with the following configuration:

- LCD_CLK polarity: high
- LCD_DE polarity: low
- LCD_VSYNC and LCD_HSYNC polarity: high
- Pixel formats: 24 bits
- Output speed is set to OSPEEDRy[1:0] = 11
- Capacitive load C = 30 pF
- Measurement points are done at CMOS levels: $0.5 \times V_{DD}$

Refer to [Section 6.3.14: I/O port characteristics](#) for more details on the input/output alternate function characteristics.

Table 112. LTDC characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min	Max	Unit
f _{CLK} D _{CLK}	LTDC clock output frequency	2.7 V < V _{DD} < 3.6 V	-	83	MHz
		1.71 V < V _{DD} < 3.6 V	-	50	
	LTDC clock output duty cycle	-	45	55	%
tw(CLKH) tw(CLKL)	Clock high time Clock low time	-	tw(CLK)/2-0.5	tw(CLK)/2+0.5	ns
t _v (DATA)	Data output valid time	-	-	5.5	
t _h (DATA)	Data output hold time	-	1	-	
t _v (HSYNC) t _v (VSYNC) t _v (DE)	HSYNC/VSYNC/DE output valid time	-	-	3	
t _h (HSYNC) t _h (VSYNC) t _h (DE)	HSYNC/VSYNC/DE output hold time	-	1	-	

1. Guaranteed by characterization results.

6.3.33 SD/SDIO/MMC card host interfaces (SDMMC)

Unless otherwise specified, the parameters given in [Table 113](#) for SDIO are derived from tests performed under the ambient temperature, f_{PCLKx} frequency and VDD supply voltage conditions summarized in [Table 21: General operating conditions](#) with the following configuration:

- Output speed is set to OSPEEDRy[1:0] = 11
- Capacitive load C = 30 pF
- Measurement points are done at CMOS levels: $0.5 \times V_{DD}$ Refer to [Section 6.3.14: I/O port characteristics](#) for more details on the input/output characteristics.

**Table 113. Dynamics characteristics:
SD / eMMC characteristics at VDD = 2.7 V to 3.6 V (1)**

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
fPP	Clock frequency in data transfer mode	-	0	-	64	MHz
-	SDIO_CK/fPCLK2 frequency ratio	-	-	-	8/3	-
tW(CKL)	Clock low time	fpp = 52 MHz	8.5	9.5	-	ns
tW(CKH)	Clock high time	fpp = 52 MHz	8.5	9.5	-	
CMD, D inputs (referenced to CK) in eMMC legacy/SDR/DDR and SD HS/SDR⁽²⁾/DDR⁽²⁾ mode						
tISU	Input setup time HS	-	1.5	-	-	ns
tIHD	Input hold time HS	-	2.0	-	-	
CMD, D outputs (referenced to CK) in eMMC legacy/SDR/DDR and SD HS/SDR⁽²⁾/DDR⁽²⁾ mode						
tOV	Output valid time HS	-	-	5.5	7	ns
tOH	Output hold time HS	-	4	-	-	
CMD, D inputs (referenced to CK) in SD default mode						
tISUD	Input setup time SD	-	0.5	-	-	ns
tIHD	Input hold time SD	-	1.5	-	-	
CMD, D outputs (referenced to CK) in SD default mode						
tOVD	Output valid default time SD	-	-	1	2	ns
tOHD	Output hold default time SD	-	0	-	-	

1. Guaranteed by characterization results.
2. For SD 1.8 V support, an external voltage converter is needed.

**Table 114. Dynamics characteristics:
eMMC characteristics at VDD = 1.71 V to 1.9 V⁽¹⁾⁽²⁾**

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
fPP	Clock frequency in data transfer mode	-	0	-	52	MHz
-	SDIO_CK/fPCLK2 frequency ratio	-	-	-	8/3	-
tW(CKL)	Clock low time	fpp = 52 MHz	8.5	9.5	-	ns
tW(CKH)	Clock high time	fpp = 52 MHz	8.5	9.5	-	
CMD, D inputs (referenced to CK) in eMMC mode						
tISU	Input setup time HS	-	0.5	-	-	ns
tIH	Input hold time HS	-	4.5	-	-	
CMD, D outputs (referenced to CK) in eMMC mode						
tOV	Output valid time HS	-	-	6	7.3	ns
tOH	Output hold time HS	-	4	-	-	

1. Guaranteed by characterization results.
2. Cload = 20 pF.

See the different SDMMC diagrams in [Figure 60](#), [Figure 61](#) and [Figure 62](#) below.

Figure 60. SDIO high-speed mode

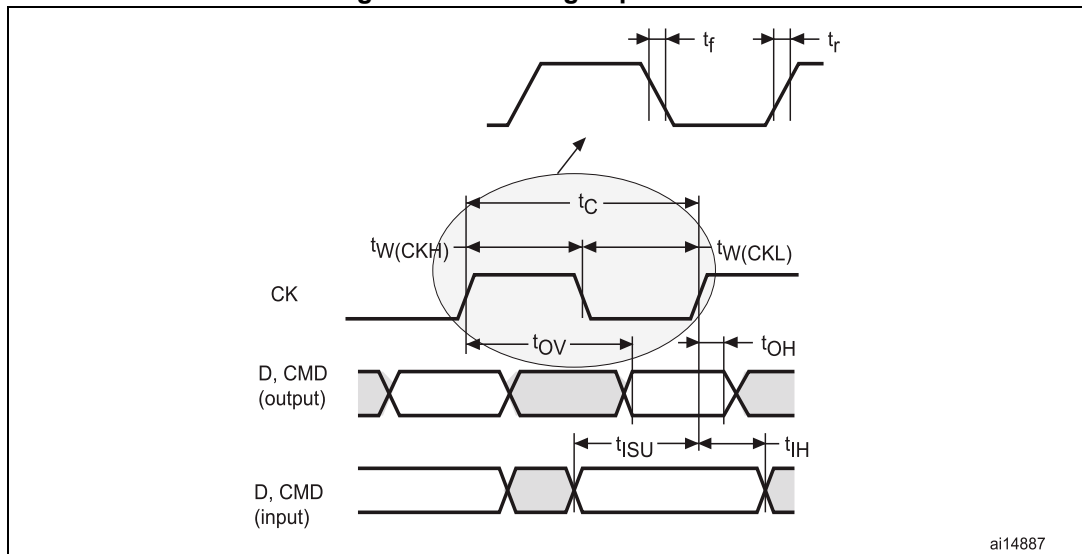


Figure 61. SD default mode

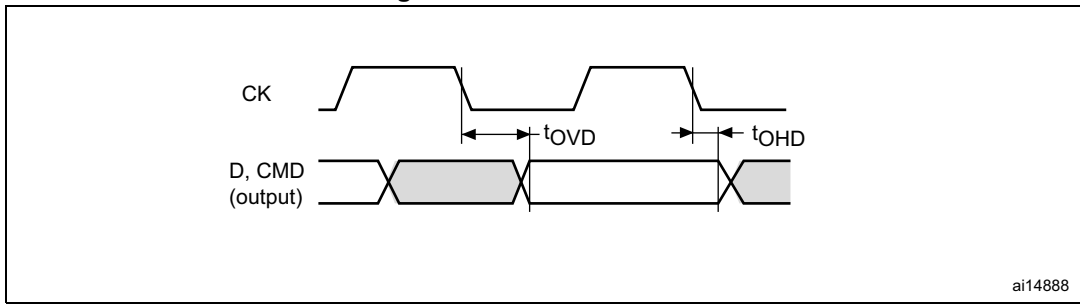
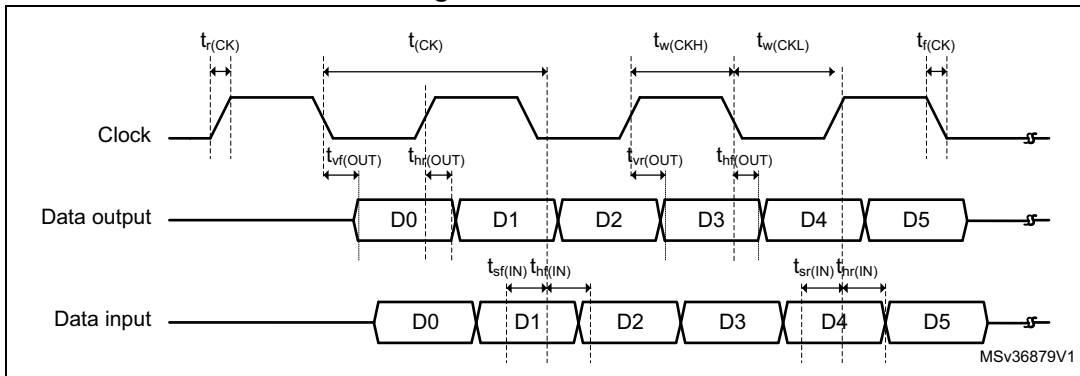


Figure 62. DDR mode



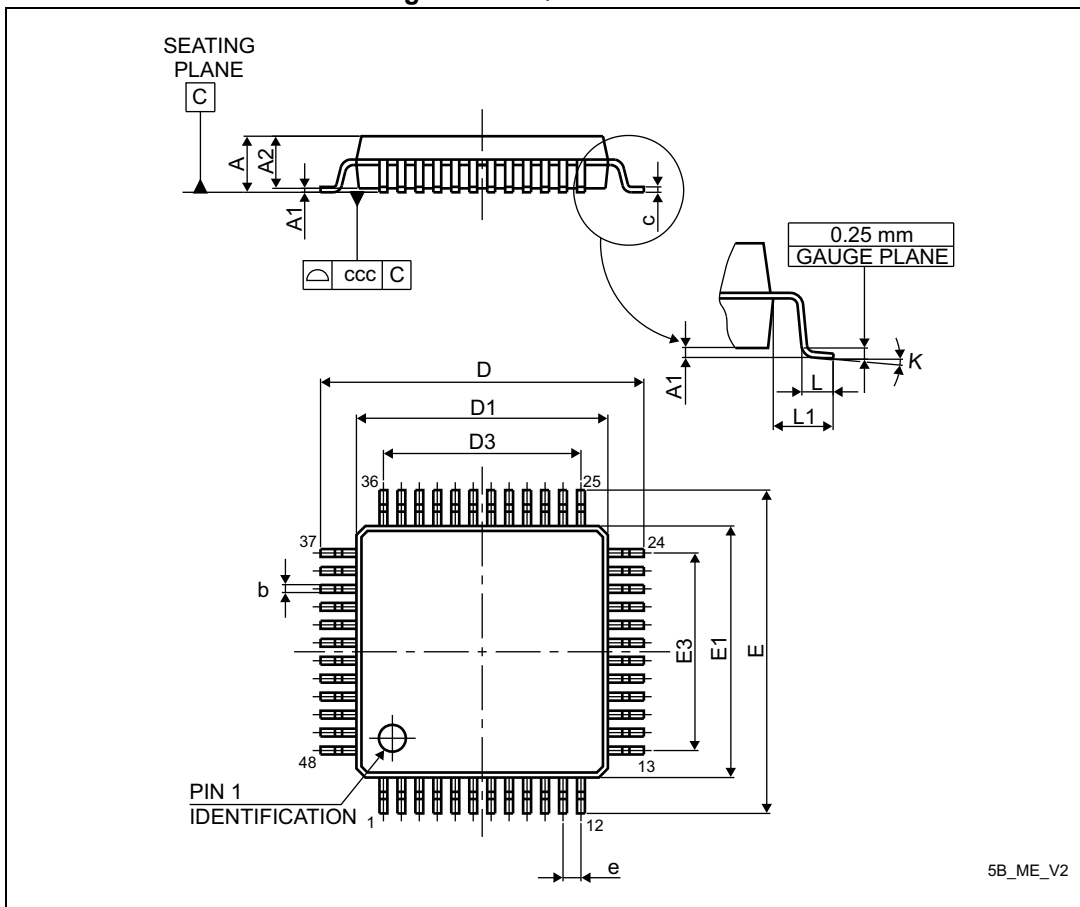
7 Package information

In order to meet environmental requirements, ST offers these devices in different grades of ECOPACK packages, depending on their level of environmental compliance. ECOPACK specifications, grade definitions and product status are available at: www.st.com. ECOPACK is an ST trademark.

7.1 LQFP48 package information

LQFP48 is a 48-pin, 7 x 7 mm low-profile quad flat package.

Figure 63. LQFP48 outline



1. Drawing is not to scale.

Table 115. LQFP48 mechanical data

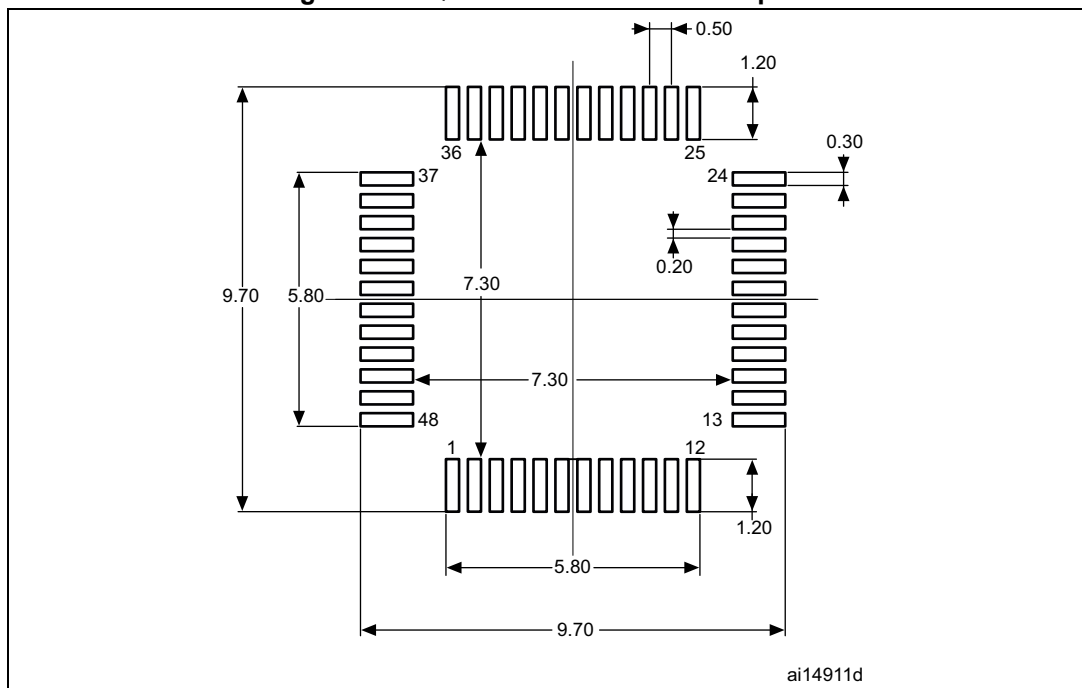
Symbol	millimeters			inches ⁽¹⁾		
	Min	Typ	Max	Min	Typ	Max
A	-	-	1.600	-	-	0.0630
A1	0.050	-	0.150	0.0020	-	0.0059
A2	1.350	1.400	1.450	0.0531	0.0551	0.0571

Table 115. LQFP48 mechanical data (continued)

Symbol	millimeters			inches ⁽¹⁾		
	Min	Typ	Max	Min	Typ	Max
b	0.170	0.220	0.270	0.0067	0.0087	0.0106
c	0.090	-	0.200	0.0035	-	0.0079
D	8.800	9.000	9.200	0.3465	0.3543	0.3622
D1	6.800	7.000	7.200	0.2677	0.2756	0.2835
D3	-	5.500	-	-	0.2165	-
E	8.800	9.000	9.200	0.3465	0.3543	0.3622
E1	6.800	7.000	7.200	0.2677	0.2756	0.2835
E3	-	5.500	-	-	0.2165	-
e	-	0.500	-	-	0.0197	-
L	0.450	0.600	0.750	0.0177	0.0236	0.0295
L1	-	1.000	-	-	0.0394	-
k	0°	3.5°	7°	0°	3.5°	7°
ccc	-	-	0.080	-	-	0.0031

1. Values in inches are converted from mm and rounded to 4 decimal digits.

Figure 64. LQFP48 recommended footprint



1. Dimensions are expressed in millimeters.

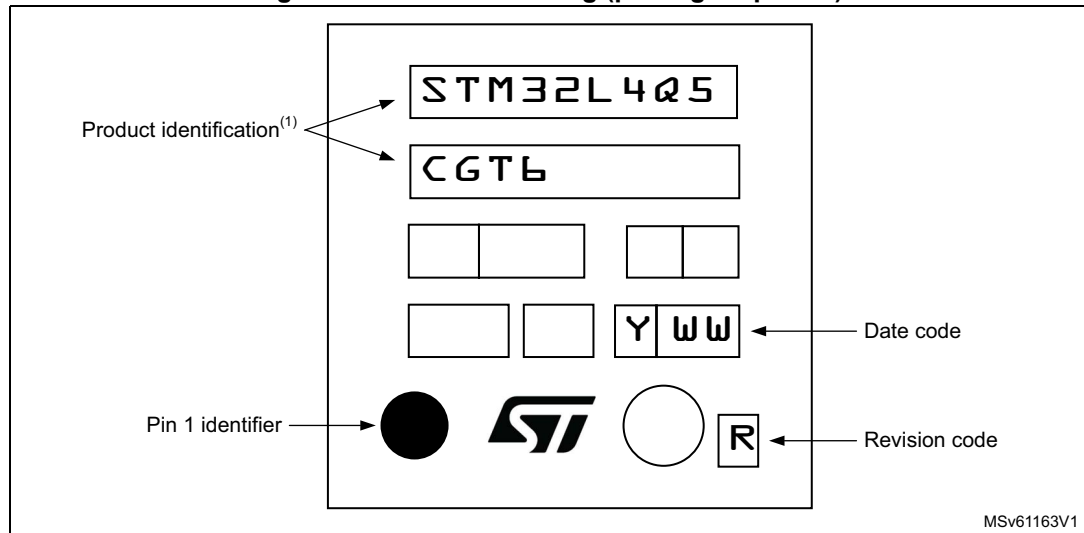
Device marking for LQFP48

The following figure gives an example of topside marking orientation versus pin 1 identifier location.

The printed markings may differ depending on the supply chain.

Other optional marking or inset/upset marks, which identify the parts throughout supply chain operations, are not indicated below.

Figure 65. LQFP48 marking (package top view)

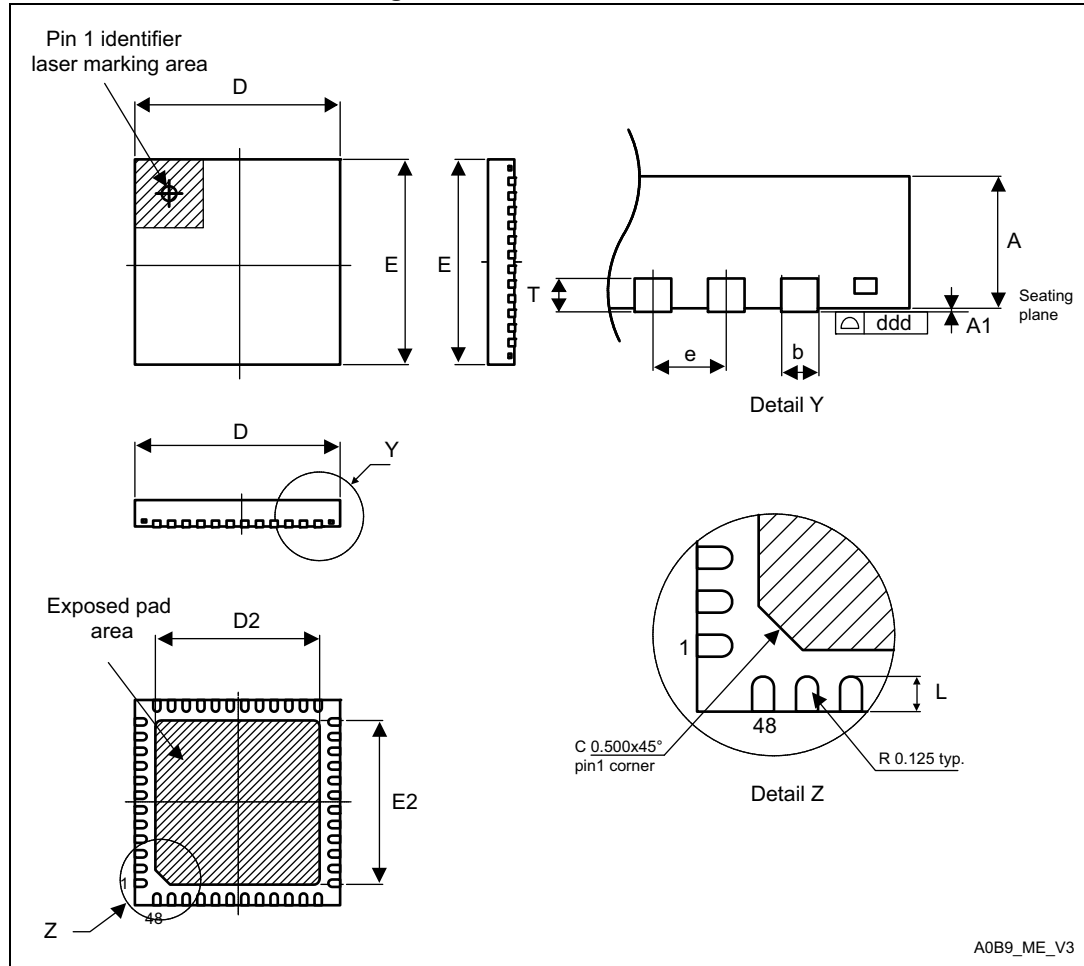


1. Parts marked as ES or E or accompanied by an engineering sample notification letter are not yet qualified and therefore not approved for use in production. ST is not responsible for any consequences resulting from such use. In no event will ST be liable for the customer using any of these engineering samples in production. ST's Quality department must be contacted prior to any decision to use these engineering samples to run a qualification activity.

7.2 UFQFPN48 package information

UFQFPN48 is a 48-lead, 7x7 mm, 0.5 mm pitch, ultra thin fine pitch quad flat package.

Figure 66. UFQFPN48 outline



1. Drawing is not to scale.
2. All leads/pads should also be soldered to the PCB to improve the lead/pad solder joint life.
3. There is an exposed die pad on the underside of the UFQFPN package. This back-side pad must be connected and soldered to PCB ground.

Table 116. UFQFPN48 mechanical data

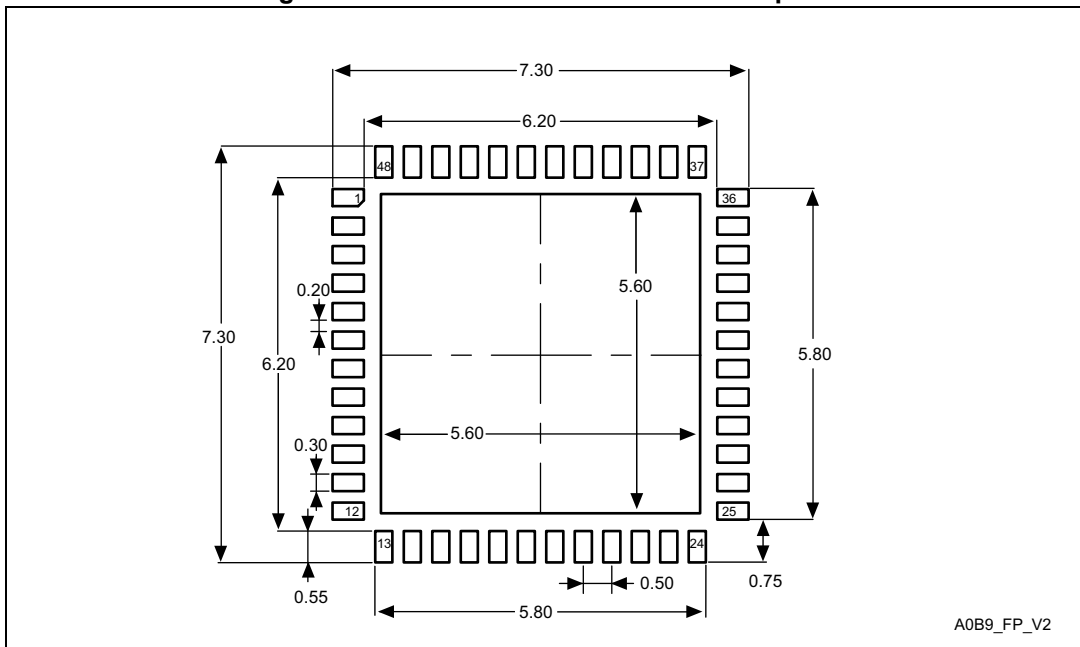
Symbol	millimeters			inches ⁽¹⁾		
	Min	Typ	Max	Min	Typ	Max
A	0.500	0.550	0.600	0.0197	0.0217	0.0236
A1	0.000	0.020	0.050	0.0000	0.0008	0.0020
D	6.900	7.000	7.100	0.2717	0.2756	0.2795
E	6.900	7.000	7.100	0.2717	0.2756	0.2795
D2	5.500	5.600	5.700	0.2165	0.2205	0.2244

Table 116. UFQFPN48 mechanical data (continued)

Symbol	millimeters			inches ⁽¹⁾		
	Min	Typ	Max	Min	Typ	Max
E2	5.500	5.600	5.700	0.2165	0.2205	0.2244
L	0.300	0.400	0.500	0.0118	0.0157	0.0197
T	-	0.152	-	-	0.0060	-
b	0.200	0.250	0.300	0.0079	0.0098	0.0118
e	-	0.500	-	-	0.0197	-
ddd	-	-	0.080	-	-	0.0031

1. Values in inches are converted from mm and rounded to 4 decimal digits.

Figure 67. UFQFPN48 recommended footprint



A0B9_FP_V2

1. Dimensions are expressed in millimeters.

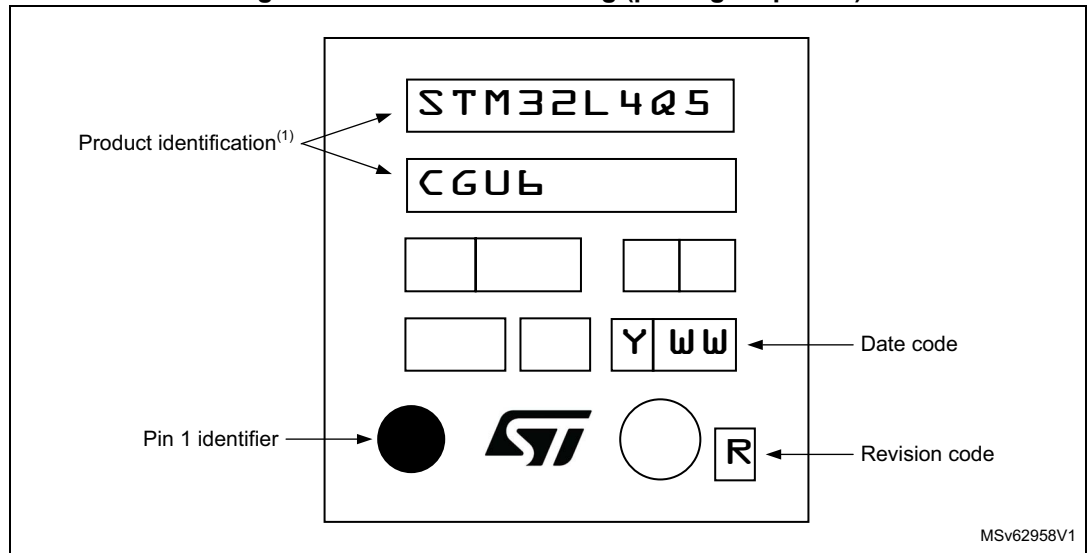
Device marking for UFQFPN48

The following figure gives an example of topside marking orientation versus pin 1 identifier location.

The printed markings may differ depending on the supply chain.

Other optional marking or inset/upset marks, which identify the parts throughout supply chain operations, are not indicated below.

Figure 68. UFQFPN48 marking (package top view)

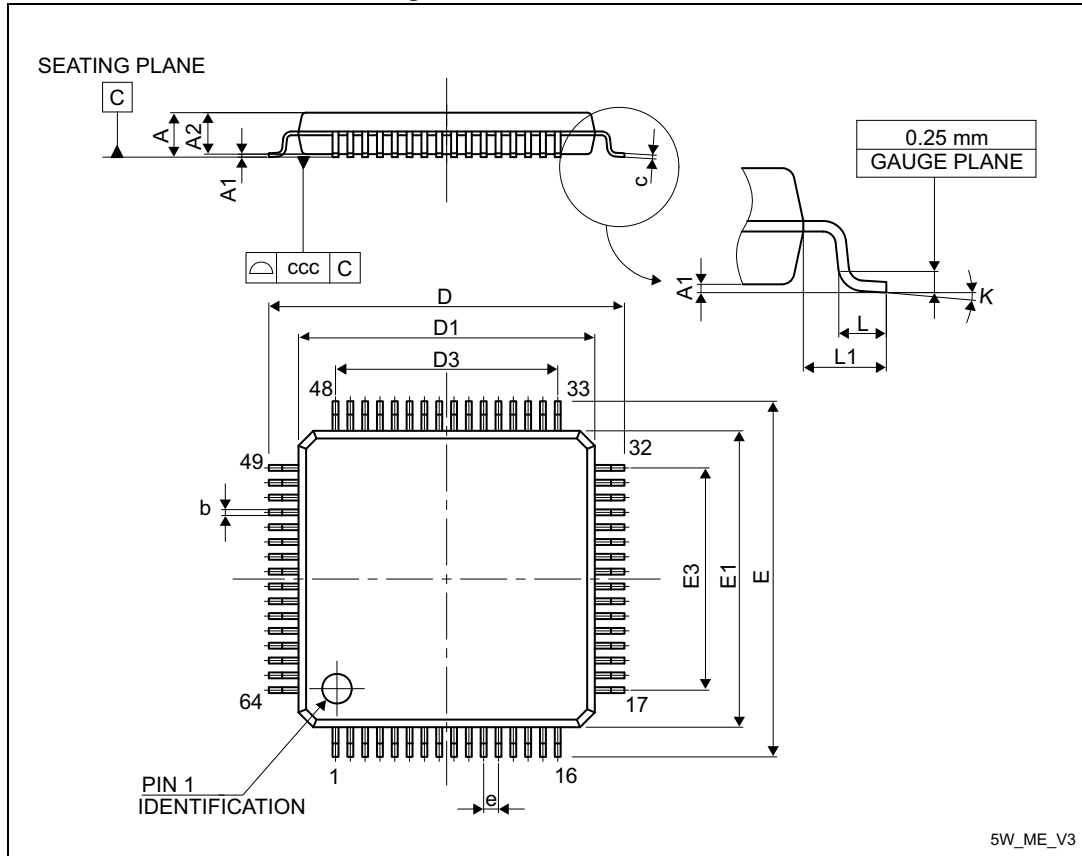


1. Parts marked as ES or E or accompanied by an engineering sample notification letter are not yet qualified and therefore not approved for use in production. ST is not responsible for any consequences resulting from such use. In no event will ST be liable for the customer using any of these engineering samples in production. ST's Quality department must be contacted prior to any decision to use these engineering samples to run a qualification activity.

7.3 LQFP64 package information

LQFP64 is a 64-pin, 10 x 10 mm low-profile quad flat package.

Figure 69. LQFP64 outline



1. Drawing is not to scale.

Table 117. LQFP 64 mechanical data

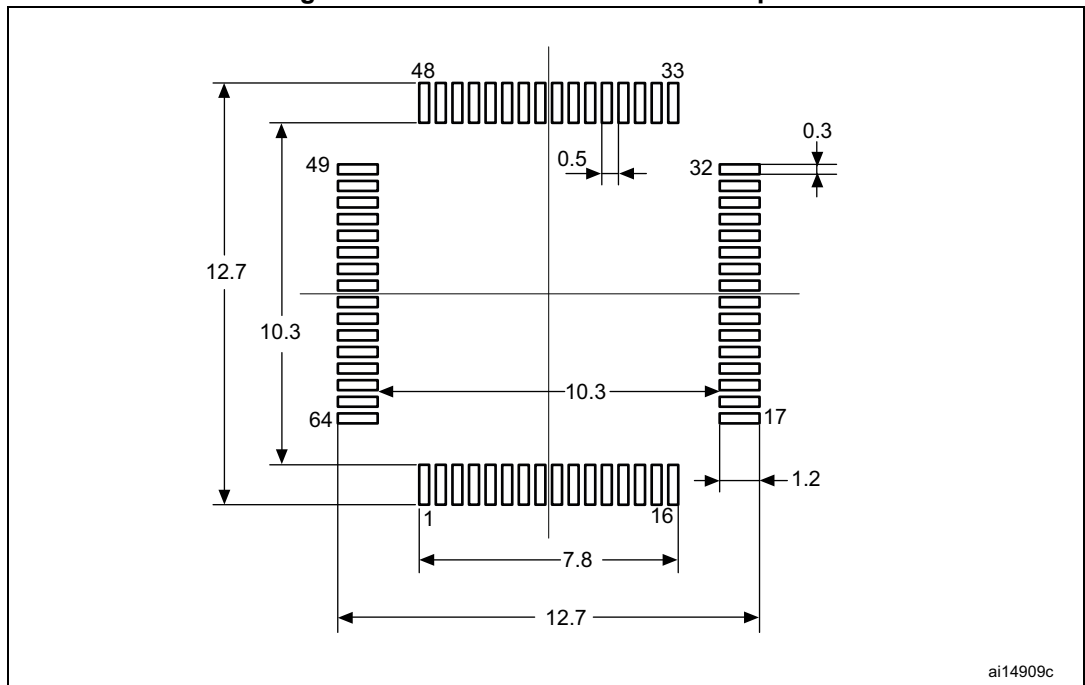
Symbol	millimeters			inches ⁽¹⁾		
	Min	Typ	Max	Min	Typ	Max
A	-	-	1.600	-	-	0.0630
A1	0.050	-	0.150	0.0020	-	0.0059
A2	1.350	1.400	1.450	0.0531	0.0551	0.0571
b	0.170	0.220	0.270	0.0067	0.0087	0.0106
c	0.090	-	0.200	0.0035	-	0.0079
D	-	12.000	-	-	0.4724	-
D1	-	10.000	-	-	0.3937	-
D3	-	7.500	-	-	0.2953	-
E	-	12.000	-	-	0.4724	-
E1	-	10.000	-	-	0.3937	-

Table 117. LQFP 64 mechanical data (continued)

Symbol	millimeters			inches ⁽¹⁾		
	Min	Typ	Max	Min	Typ	Max
E3	-	7.500	-	-	0.2953	-
e	-	0.500	-	-	0.0197	-
K	0°	3.5°	7°	0°	3.5°	7°
L	0.450	0.600	0.750	0.0177	0.0236	0.0295
L1	-	1.000	-	-	0.0394	-
ccc	-	-	0.080	-	-	0.0031

1. Values in inches are converted from mm and rounded to 4 decimal digits.

Figure 70. LQFP64 recommended footprint



ai14909c

1. Dimensions are expressed in millimeters.

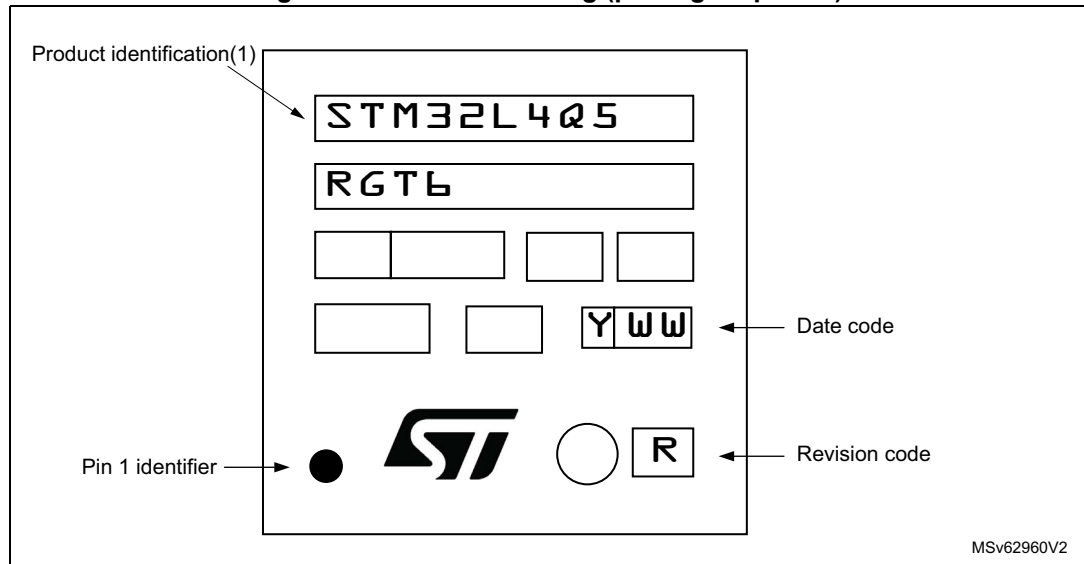
Device marking for LQFP64

The following figure gives an example of topside marking orientation versus pin 1 identifier location.

The printed markings may differ depending on the supply chain.

Other optional marking or inset/upset marks, which identify the parts throughout supply chain operations, are not indicated below.

Figure 71. LQFP64 marking (package top view)

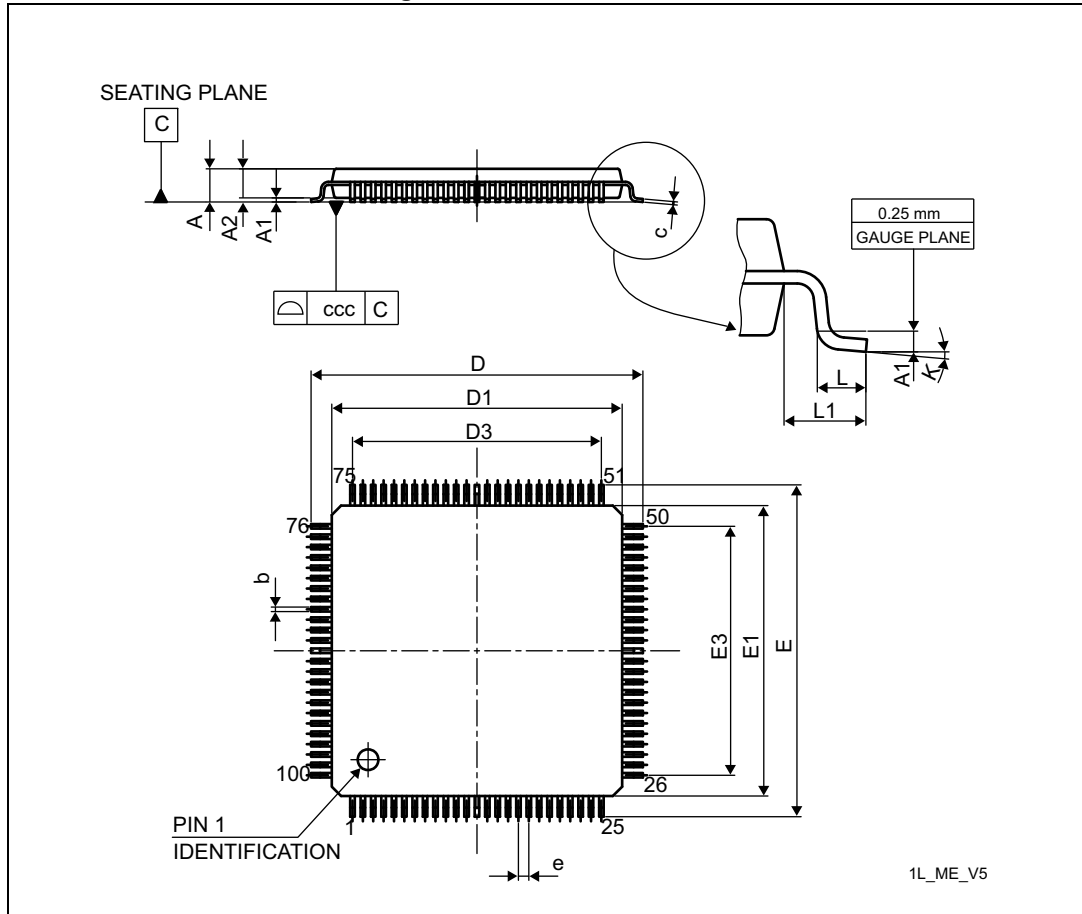


1. Parts marked as ES or E or accompanied by an engineering sample notification letter are not yet qualified and therefore not approved for use in production. ST is not responsible for any consequences resulting from such use. In no event will ST be liable for the customer using any of these engineering samples in production. ST's Quality department must be contacted prior to any decision to use these engineering samples to run a qualification activity.

7.4 LQFP100 package information

LQFP100 is a 100-pin, 14 x 14 mm low-profile quad flat package.

Figure 72. LQFP100 outline



1. Drawing is not to scale.

Table 118. LQFP100 mechanical data

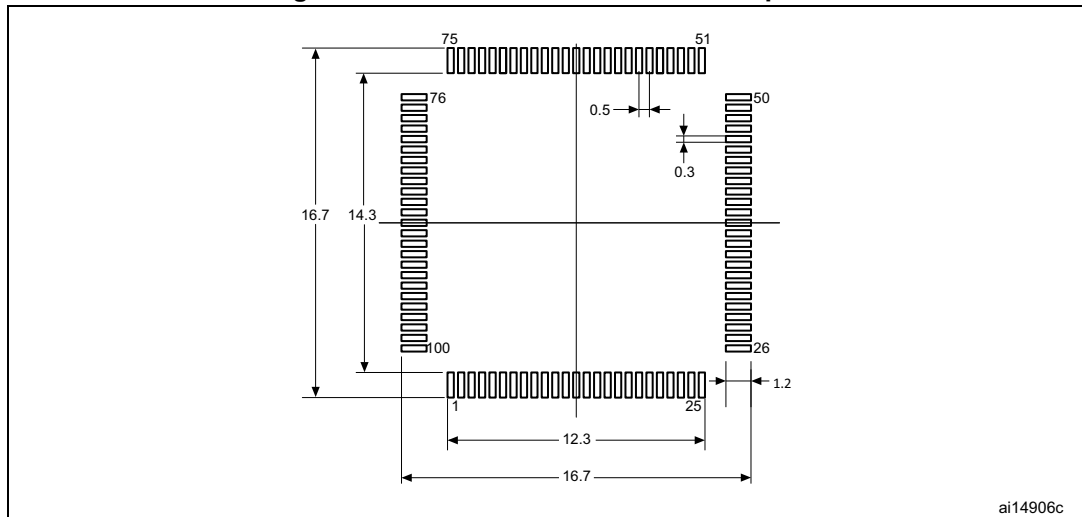
Symbol	millimeters			inches ⁽¹⁾		
	Min	Typ	Max	Min	Typ	Max
A	-	-	1.600	-	-	0.0630
A1	0.050	-	0.150	0.0020	-	0.0059
A2	1.350	1.400	1.450	0.0531	0.0551	0.0571
b	0.170	0.220	0.270	0.0067	0.0087	0.0106
c	0.090	-	0.200	0.0035	-	0.0079
D	15.800	16.000	16.200	0.6220	0.6299	0.6378
D1	13.800	14.000	14.200	0.5433	0.5512	0.5591
D3	-	12.000	-	-	0.4724	-

Table 118. LQFP100 mechanical data (continued)

Symbol	millimeters			inches ⁽¹⁾		
	Min	Typ	Max	Min	Typ	Max
E	15.800	16.000	16.200	0.6220	0.6299	0.6378
E1	13.800	14.000	14.200	0.5433	0.5512	0.5591
E3	-	12.000	-	-	0.4724	-
e	-	0.500	-	-	0.0197	-
L	0.450	0.600	0.750	0.0177	0.0236	0.0295
L1	-	1.000	-	-	0.0394	-
k	0.0°	3.5°	7.0°	0.0°	3.5°	7.0°
ccc	-	-	0.080	-	-	0.0031

1. Values in inches are converted from mm and rounded to 4 decimal digits.

Figure 73. LQFP100 recommended footprint



ai14906c

1. Dimensions are expressed in millimeters.

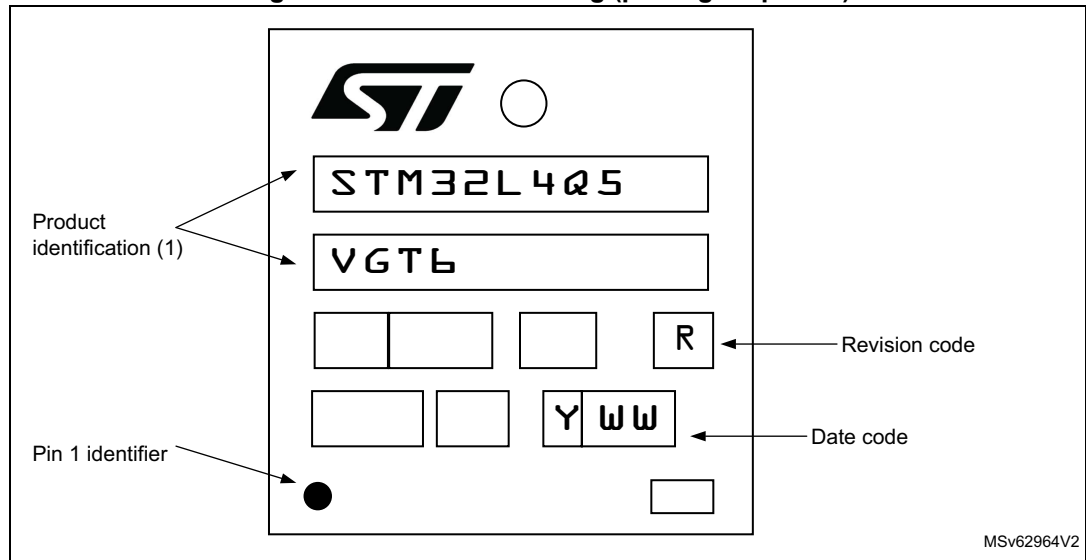
Device marking for LQFP100

The following figure gives an example of topside marking orientation versus pin 1 identifier location.

The printed markings may differ depending on the supply chain.

Other optional marking or inset/upset marks, which identify the parts throughout supply chain operations, are not indicated below.

Figure 74. LQFP100 marking (package top view)

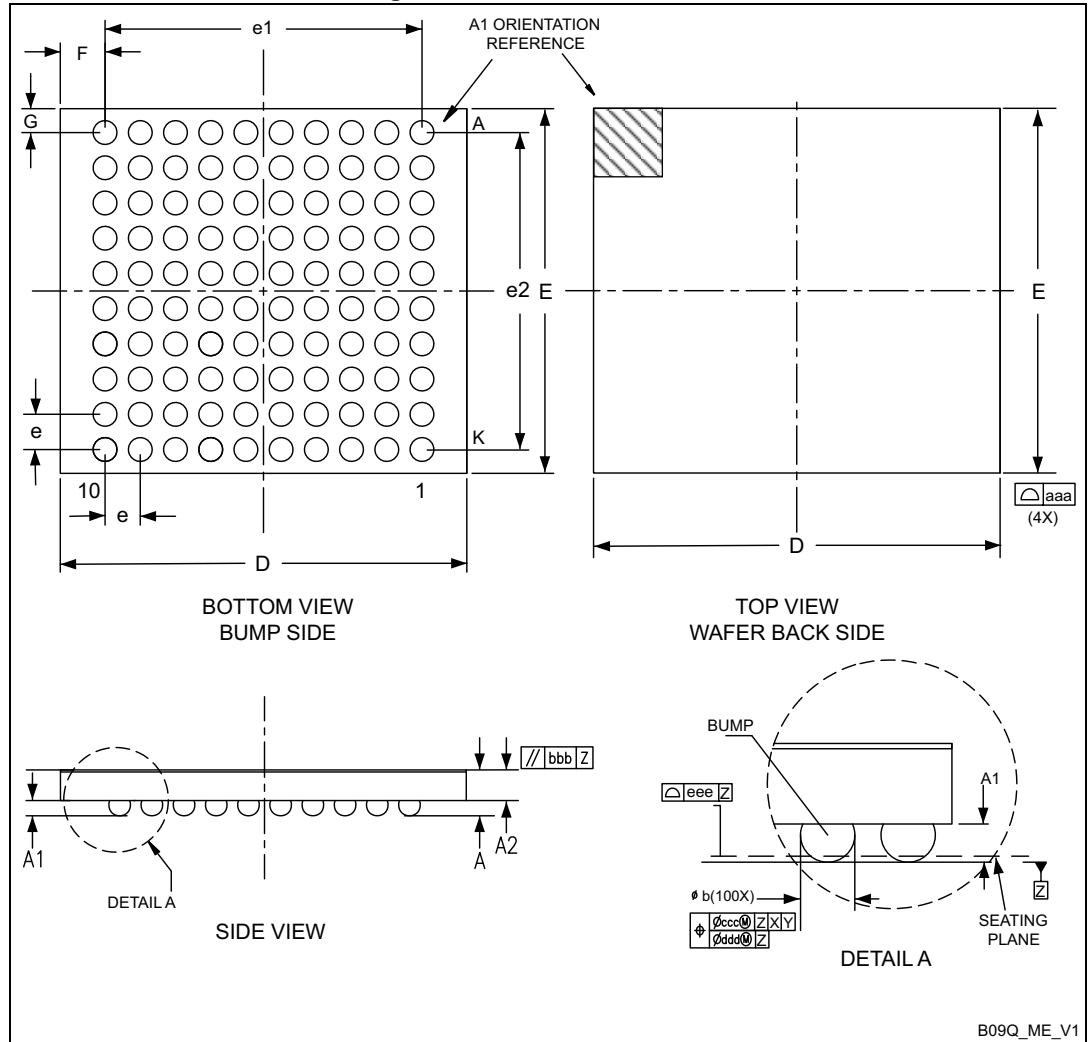


1. Parts marked as ES or E or accompanied by an engineering sample notification letter are not yet qualified and therefore not approved for use in production. ST is not responsible for any consequences resulting from such use. In no event will ST be liable for the customer using any of these engineering samples in production. ST's Quality department must be contacted prior to any decision to use these engineering samples to run a qualification activity.

7.5 WLCSP100 package information

WLCSP100 is a 100-ball, 4.437 x 4.456 mm, 0.4 mm pitch wafer level chip scale package.

Figure 75. WLCSP100 outline



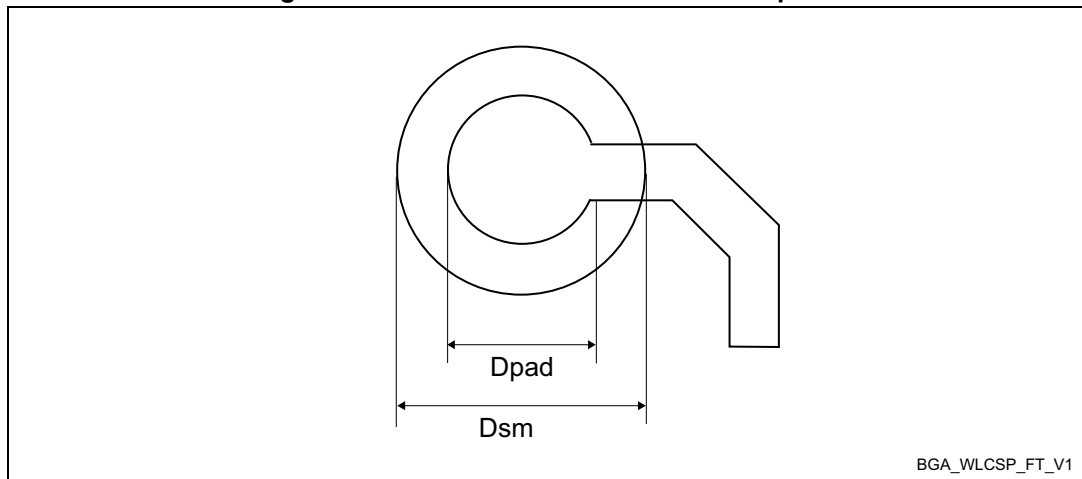
1. Drawing is not to scale.
2. Dimension is measured at the maximum bump diameter parallel to primary datum Z.
3. Primary datum Z and seating plane are defined by the spherical crowns of the bump.
4. Bump position designation per JESD 95-1, SPP-010.

Table 119. WLCSP100 mechanical data

Symbol	millimeters			inches ⁽¹⁾		
	Min	Typ	Max	Min	Typ	Max
A	0.520	0.550	0.580	0.0205	0.0217	0.0228
A1	0.155	0.170	0.185	0.0061	0.0067	0.0073
A2	0.355	0.380	0.405	0.0140	0.0150	0.0159
A3 ⁽²⁾	0.020	0.025	0.030	0.0008	0.0010	0.0012
Ø b ⁽³⁾	0.230	0.255	0.280	0.0091	0.0100	0.0110
D	4.417	4.437	4.457	0.1739	0.1747	0.1755
E	4.436	4.456	4.476	0.1746	0.1754	0.1762
e	-	0.400	-	-	0.0157	-
e1	-	3.600	-	-	0.1417	-
e2	-	3.600	-	-	0.1417	-
F	-	0.420	-	-	0.0165	-
G	-	0.430	-	-	0.0169	-
aaa	-	-	0.100	-	-	0.0039
bbb	-	-	0.100	-	-	0.0039
ccc	-	-	0.100	-	-	0.0039
ddd	-	-	0.050	-	-	0.0020
eee	-	-	0.050	-	-	0.0020

1. Values in inches are converted from mm and rounded to 4 decimal digits.
2. Back side coating.
3. Dimension is measured at the maximum bump diameter parallel to primary datum Z.

Figure 76. WLCSP100 recommended footprint



BGA_WLCSP_FT_V1

Table 120. WLCSP100 recommended PCB design rules (0.4 mm pitch)

Dimension	Recommended values
Pitch	0.4 mm
Dpad	0.225 mm
Dsm	0.290 mm
Stencil thickness	0.1 mm

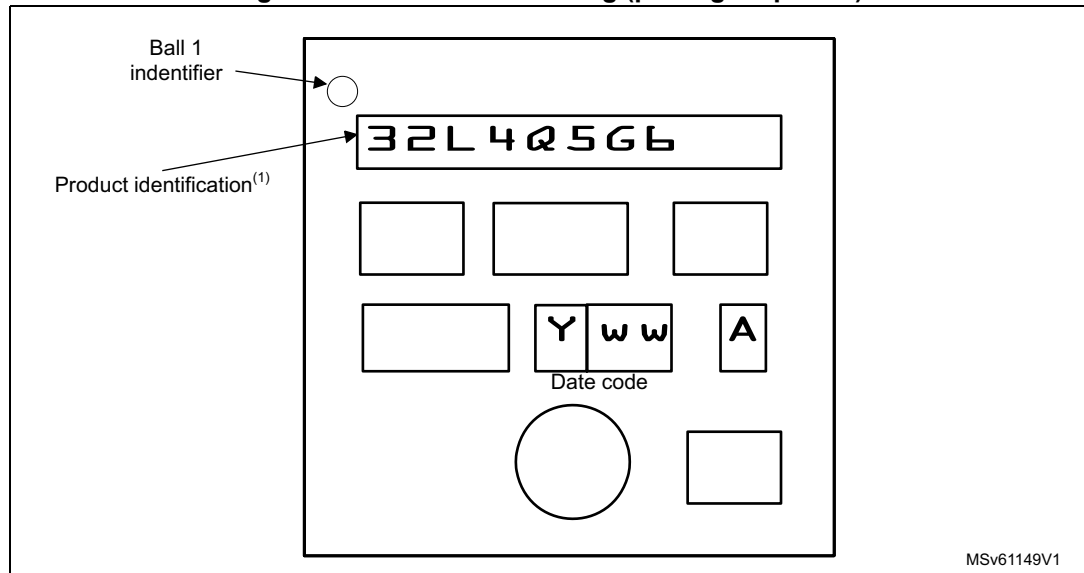
Device marking for WLCSP100

The following figure gives an example of topside marking orientation versus pin 1 identifier location.

The printed markings may differ depending on the supply chain.

Other optional marking or inset/upset marks, which identify the parts throughout supply chain operations, are not indicated below.

Figure 77. WLCSP100 marking (package top view)

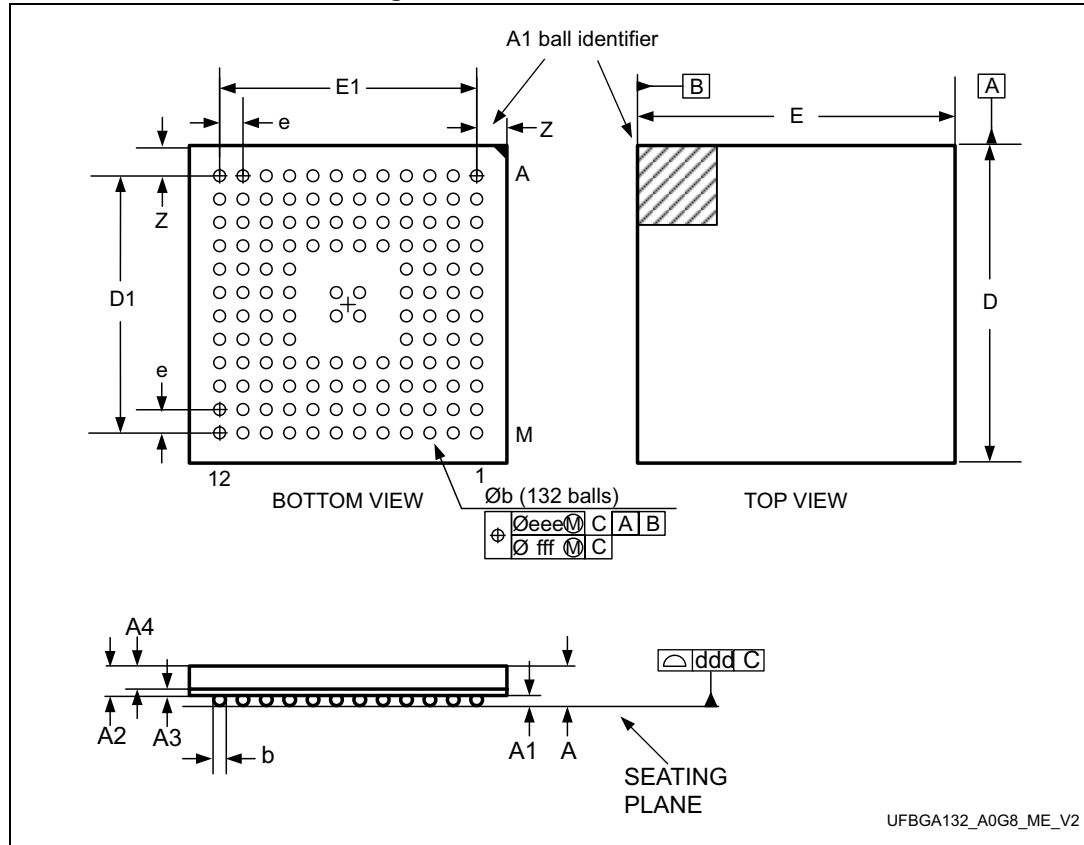


1. Parts marked as ES or E or accompanied by an engineering sample notification letter are not yet qualified and therefore not approved for use in production. ST is not responsible for any consequences resulting from such use. In no event will ST be liable for the customer using any of these engineering samples in production. ST's Quality department must be contacted prior to any decision to use these engineering samples to run a qualification activity.

7.6 UFBGA132 package information

UFBGA132 is a 132-ball, 7 x 7 mm ultra thin fine pitch ball grid array package.

Figure 78. UFBGA132 outline



1. Drawing is not to scale.

Table 121. UFBGA132 mechanical data

Symbol	millimeters			inches ⁽¹⁾		
	Min	Typ	Max	Min	Typ	Max
A	-	-	0.600	-	-	0.0236
A1	-	-	0.110	-	-	0.0043
A2	-	0.450	-	-	0.0177	-
A3	-	0.130	-	-	0.0051	0.0094
A4	-	0.320	-	-	0.0126	-
b	0.240	0.290	0.340	0.0094	0.0114	0.0134
D	6.850	7.000	7.150	0.2697	0.2756	0.2815
D1	-	5.500	-	-	0.2165	-
E	6.850	7.000	7.150	0.2697	0.2756	0.2815
E1	-	5.500	-	-	0.2165	-

Table 121. UFBGA132 mechanical data (continued)

Symbol	millimeters			inches ⁽¹⁾		
	Min	Typ	Max	Min	Typ	Max
e	-	0.500	-	-	0.0197	-
Z	-	0.750	-	-	0.0295	-
ddd	-	0.080	-	-	0.0031	-
eee	-	0.150	-	-	0.0059	-
fff	-	0.050	-	-	0.0020	-

1. Values in inches are converted from mm and rounded to 4 decimal digits.

Figure 79. UFBGA132 recommended footprint

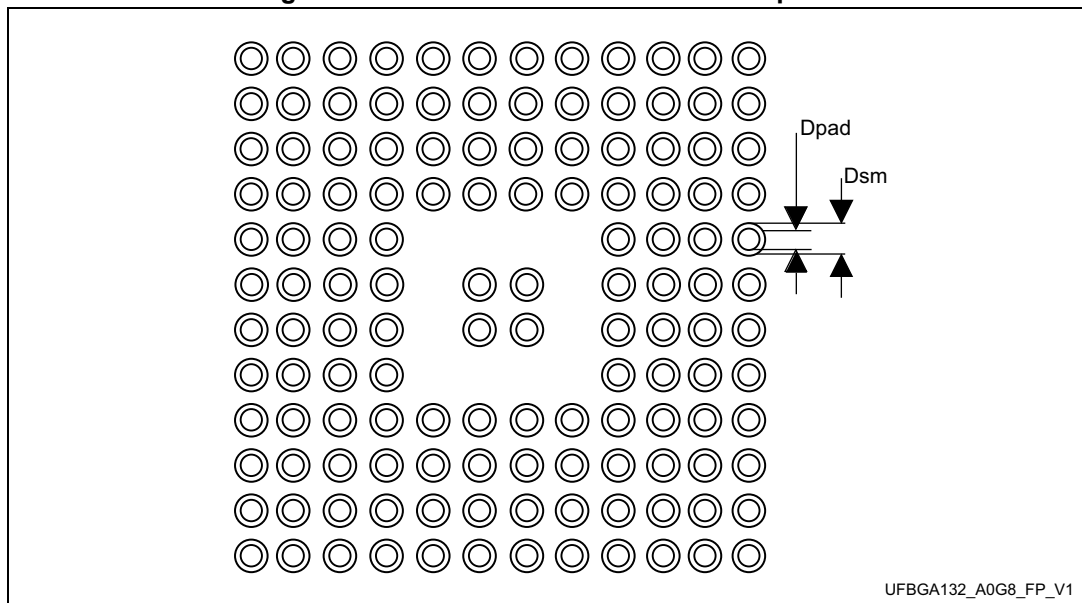


Table 122. UFBGA132 recommended PCB design rules (0.5 mm pitch BGA)

Dimension	Recommended values
Pitch	0.5 mm
Dpad	0.280 mm
Dsm	0.370 mm typ. (depends on the soldermask registration tolerance)
Stencil opening	0.280 mm
Stencil thickness	Between 0.100 mm and 0.125 mm
Pad trace width	0.100 mm
Ball diameter	0.280 mm

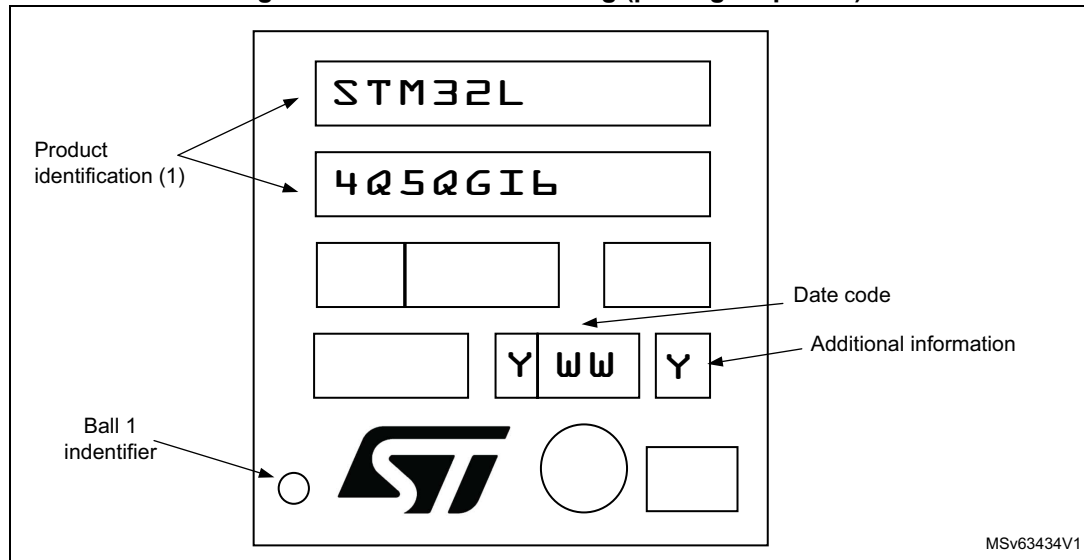
Device marking for UFBGA132

The following figure gives an example of topside marking orientation versus ball A1 identifier location.

The printed markings may differ depending on the supply chain.

Other optional marking or inset/upset marks, which identify the parts throughout supply chain operations, are not indicated below.

Figure 80. UFBGA132 marking (package top view)

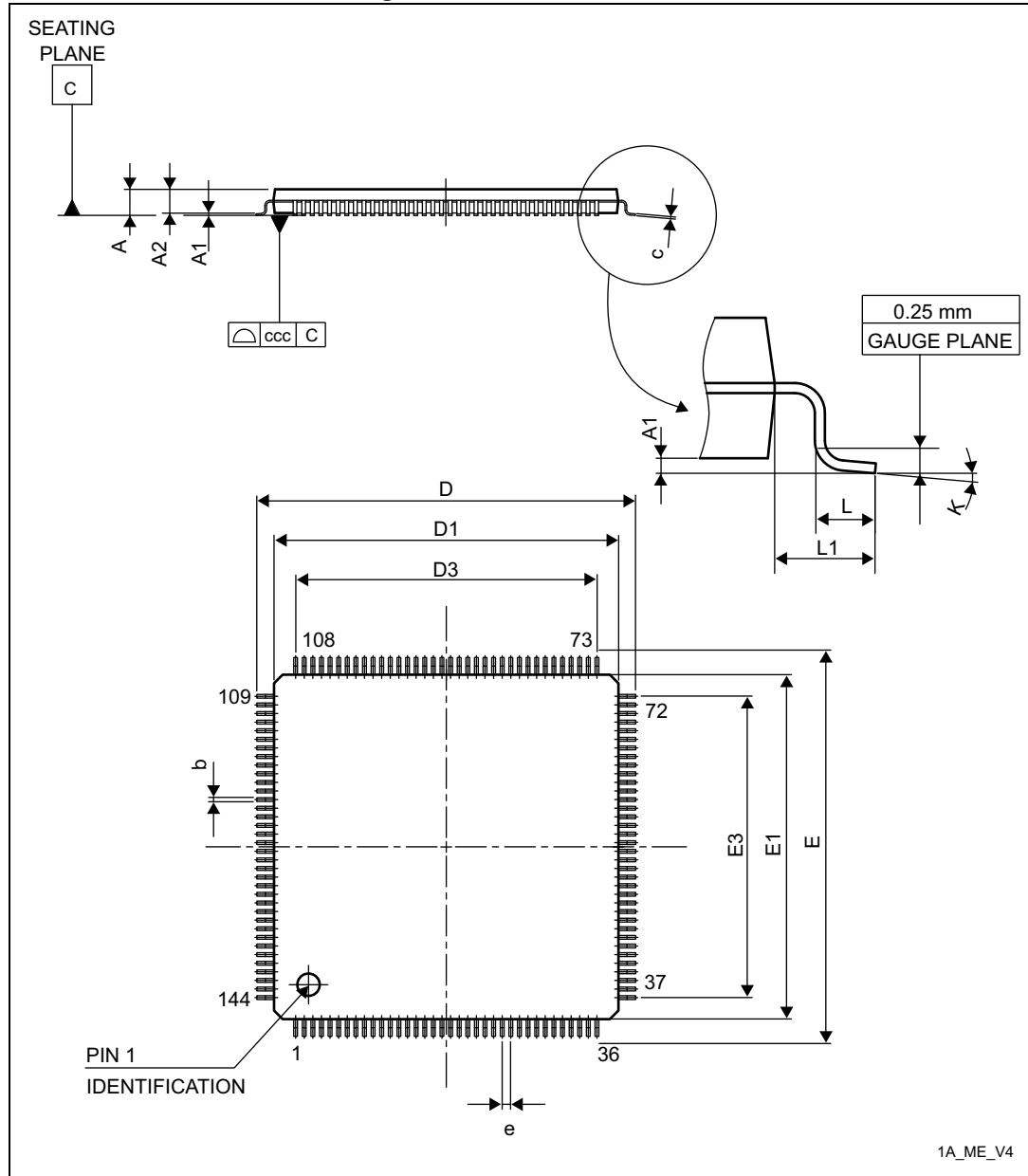


1. Parts marked as ES or E or accompanied by an engineering sample notification letter are not yet qualified and therefore not approved for use in production. ST is not responsible for any consequences resulting from such use. In no event will ST be liable for the customer using any of these engineering samples in production. ST's Quality department must be contacted prior to any decision to use these engineering samples to run a qualification activity.

7.7 LQFP144 package information

LQFP144 is a 144-pin, 20 x 20 mm low-profile quad flat package.

Figure 81. LQFP144 outline



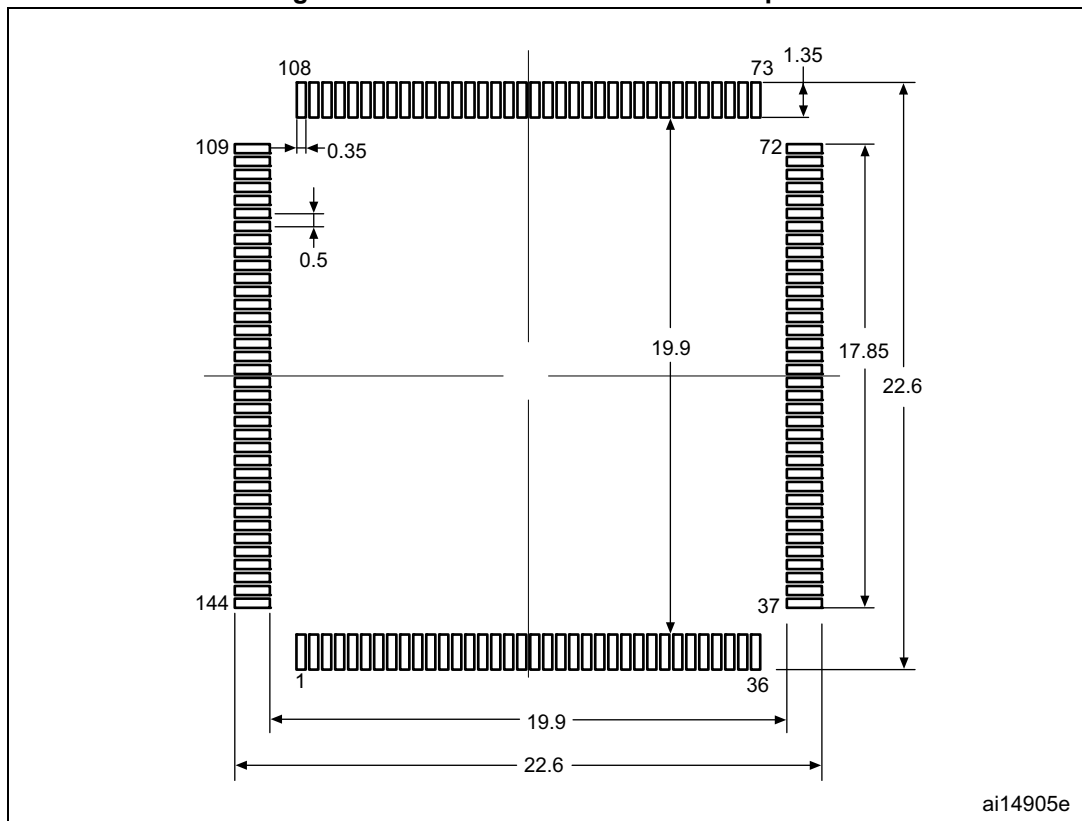
1. Drawing is not to scale.

Table 123. LQFP144 mechanical data

Symbol	millimeters			inches ⁽¹⁾		
	Min	Typ	Max	Min	Typ	Max
A	-	-	1.600	-	-	0.0630
A1	0.050	-	0.150	0.0020	-	0.0059
A2	1.350	1.400	1.450	0.0531	0.0551	0.0571
b	0.170	0.220	0.270	0.0067	0.0087	0.0106
c	0.090	-	0.200	0.0035	-	0.0079
D	21.800	22.000	22.200	0.8583	0.8661	0.8740
D1	19.800	20.000	20.200	0.7795	0.7874	0.7953
D3	-	17.500	-	-	0.6890	-
E	21.800	22.000	22.200	0.8583	0.8661	0.8740
E1	19.800	20.000	20.200	0.7795	0.7874	0.7953
E3	-	17.500	-	-	0.6890	-
e	-	0.500	-	-	0.0197	-
L	0.450	0.600	0.750	0.0177	0.0236	0.0295
L1	-	1.000	-	-	0.0394	-
k	0°	3.5°	7°	0°	3.5°	7°
ccc	-	-	0.080	-	-	0.0031

1. Values in inches are converted from mm and rounded to 4 decimal digits.

Figure 82. LQFP144 recommended footprint



1. Dimensions are expressed in millimeters.

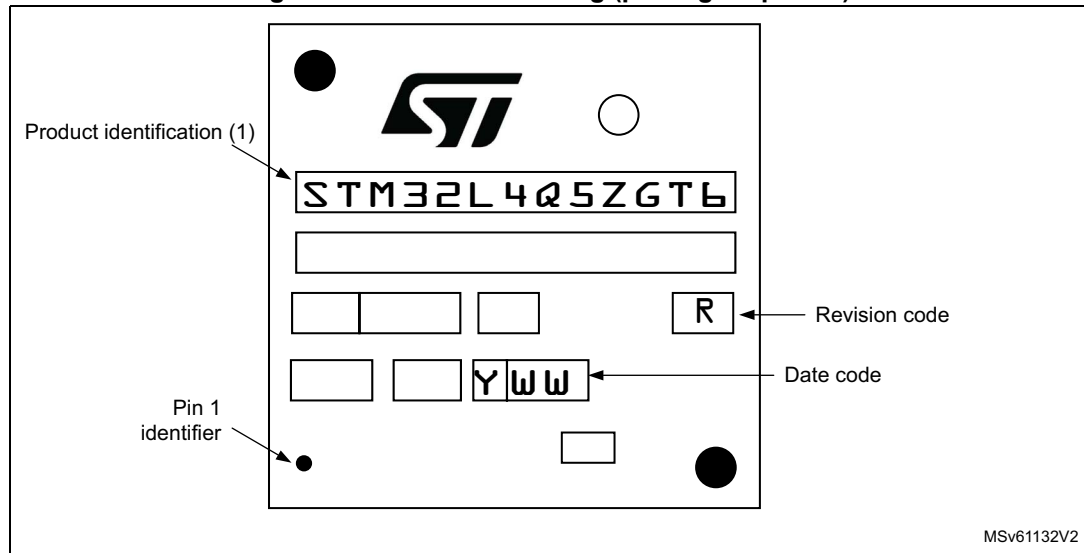
Device marking for LQFP144

The following figures gives an example of topside marking orientation versus pin 1 identifier location.

The printed markings may differ depending on the supply chain.

Other optional marking or inset/upset marks, which identify the parts throughout supply chain operations, are not indicated below.

Figure 83. LQFP144 marking (package top view)

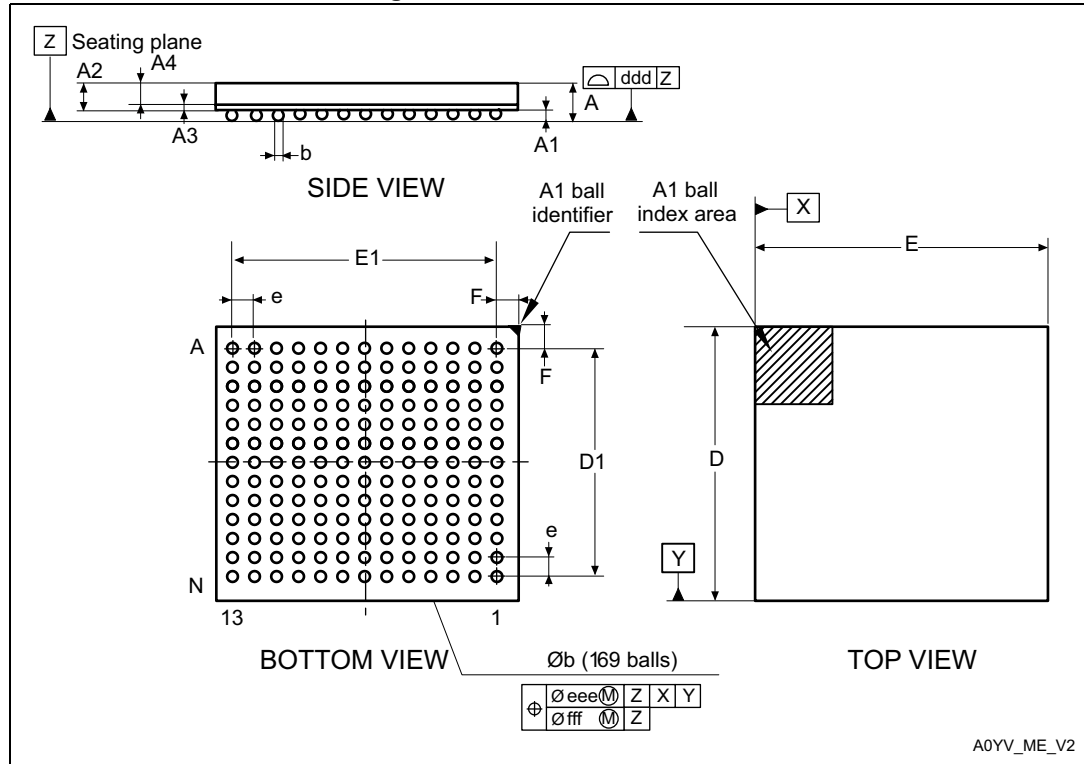


1. Parts marked as ES or E or accompanied by an engineering sample notification letter are not yet qualified and therefore not approved for use in production. ST is not responsible for any consequences resulting from such use. In no event will ST be liable for the customer using any of these engineering samples in production. ST's Quality department must be contacted prior to any decision to use these engineering samples to run a qualification activity.

7.8 UFBGA169 package information

UFBGA169 is a 169-ball, 7 x 7 mm, 0.50 mm pitch, ultra fine pitch ball grid array package.

Figure 84. UFBGA169 outline



1. Drawing is not to scale.

Table 124. UFBGA169 mechanical data

Symbol	millimeters			inches ⁽¹⁾		
	Min.	Typ.	Max.	Min.	Typ.	Max.
A	0.460	0.530	0.600	0.0181	0.0209	0.0236
A1	0.050	0.080	0.110	0.0020	0.0031	0.0043
A2	0.400	0.450	0.500	0.0157	0.0177	0.0197
A3	-	0.130	-	-	0.0051	-
A4	0.270	0.320	0.370	0.0106	0.0126	0.0146
b	0.230	0.280	0.330	0.0091	0.0110	0.0130
D	6.950	7.000	7.050	0.2736	0.2756	0.2776
D1	5.950	6.000	6.050	0.2343	0.2362	0.2382
E	6.950	7.000	7.050	0.2736	0.2756	0.2776
E1	5.950	6.000	6.050	0.2343	0.2362	0.2382
e	-	0.500	-	-	0.0197	-
F	0.450	0.500	0.550	0.0177	0.0197	0.0217

Table 124. UFBGA169 mechanical data (continued)

Symbol	millimeters			inches ⁽¹⁾		
	Min.	Typ.	Max.	Min.	Typ.	Max.
ddd	-	-	0.100	-	-	0.0039
eee	-	-	0.150	-	-	0.0059
fff	-	-	0.050	-	-	0.0020

1. Values in inches are converted from mm and rounded to 4 decimal digits.

Figure 85. UFBGA169 recommended footprint

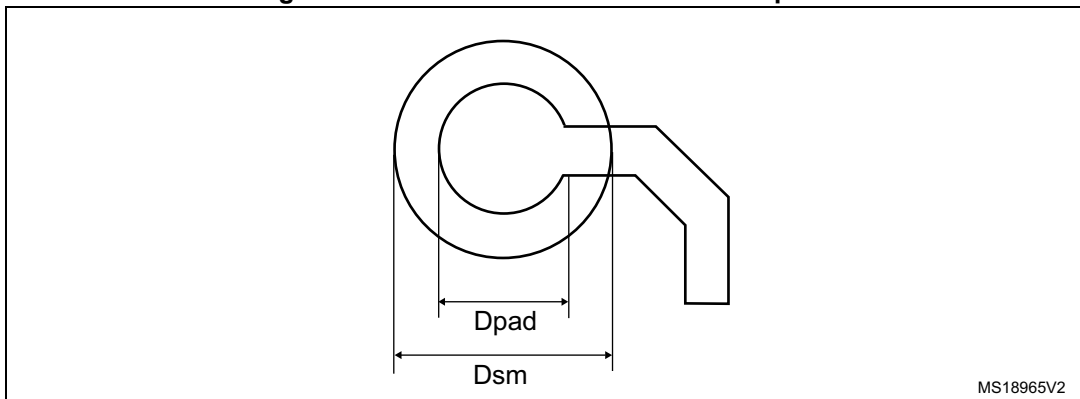


Table 125. UFBGA169 recommended PCB design rules (0.5 mm pitch BGA)

Dimension	Recommended values
Pitch	0.5 mm
Dpad	0.27 mm
Dsm	0.35 mm typ. (depends on the soldermask registration tolerance)
Solder paste	0.27 mm aperture diameter.

Note: Non-solder mask defined (NSMD) pads are recommended.

Note: 4 to 6 mils solder paste screen printing process.

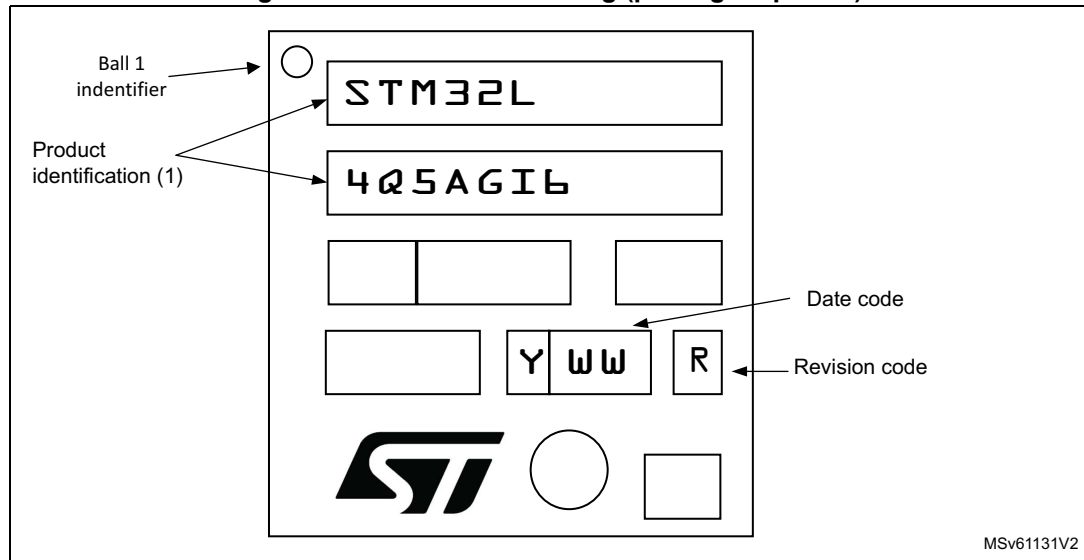
Device marking for UFBGA169

The following figure gives an example of topside marking orientation versus pin 1 identifier location.

The printed markings may differ depending on the supply chain.

Other optional marking or inset/upset marks, which identify the parts throughout supply chain operations, are not indicated below.

Figure 86. UFBGA169 marking (package top view)



MSv61131V2

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7.9 Thermal characteristics

The maximum chip-junction temperature, $T_J \text{ max}$, in degrees Celsius, may be calculated using the following equation:

$$T_J \text{ max} = T_A \text{ max} + (P_D \text{ max} \times \Theta_{JA})$$

Where:

- $T_A \text{ max}$ is the maximum ambient temperature in °C,
- Θ_{JA} is the package junction-to-ambient thermal resistance, in °C/W,
- $P_D \text{ max}$ is the sum of $P_{INT} \text{ max}$ and $P_{I/O} \text{ max}$ ($P_D \text{ max} = P_{INT} \text{ max} + P_{I/O} \text{ max}$),
- $P_{INT} \text{ max}$ is the product of I_{DD} and V_{DD} , expressed in Watts. This is the maximum chip internal power.

$P_{I/O} \text{ max}$ represents the maximum power dissipation on output pins where:

$$P_{I/O} \text{ max} = \sum (V_{OL} \times I_{OL}) + \sum ((V_{DDIOx} - V_{OH}) \times I_{OH}),$$

taking into account the actual V_{OL} / I_{OL} and V_{OH} / I_{OH} of the I/Os at low and high level in the application.

Table 126. Package thermal characteristics

Symbol	Parameter	Value	Unit
Θ_{JA}	Thermal resistance junction-ambient UFQFPN48 - 7 × 7 mm	24.8	°C/W
	Thermal resistance junction-ambient LQFP48 - 7 × 7 mm	50.7	
	Thermal resistance junction-ambient LQFP64 - 10 × 10 mm	49.2	
	Thermal resistance junction-ambient LQFP100 - 14 × 14 mm	47.8	
	Thermal resistance junction-ambient WLCSP100	35.8	
	Thermal resistance junction-ambient UFBGA132 - 7 × 7 mm	34.2	
	Thermal resistance junction-ambient LQFP144 - 20 × 20 mm	46.3	
	Thermal resistance junction-ambient UFBGA169 - 7 × 7 mm	35.2	

Table 126. Package thermal characteristics (continued)

Symbol	Parameter	Value	Unit
Θ_{JB}	Thermal resistance junction-board UFQFPN48 - 7 × 7 mm	12.9	°C/W
	Thermal resistance junction-board LQFP48 - 7 × 7 mm	35.7	
	Thermal resistance junction-board LQFP64 - 10 × 10 mm	37.7	
	Thermal resistance junction-board LQFP100 - 14 × 14 mm	39.6	
	Thermal resistance junction-board WLCSP100	NA	
	Thermal resistance junction-board UFBGA132 - 7 × 7 mm	11.4	
	Thermal resistance junction-board LQFP144 - 20 × 20 mm	41.8	
	Thermal resistance junction-board UFBGA169 - 7 × 7 mm	13.4	
Θ_{JC}	Thermal resistance junction-case UFQFPN48 - 7 × 7 mm	1.3	°C/W
	Thermal resistance junction-case LQFP48 - 7 × 7 mm	13.4	
	Thermal resistance junction-case LQFP64 - 10 × 10 mm	13.2	
	Thermal resistance junction-case LQFP100 - 14 × 14 mm	13	
	Thermal resistance junction-case WLCSP100	NA	
	Thermal resistance junction-case UFBGA132 - 7 × 7 mm	34.7	
	Thermal resistance junction-case LQFP144 - 20 × 20 mm	12.8	
	Thermal resistance junction-case UFBGA169 - 7 × 7 mm	34.7	

Reference document

JESD51-2 Integrated Circuits Thermal Test Method Environment Conditions - Natural Convection (Still Air). Available from www.jedec.org

8 Ordering information

Table 127. Ordering information scheme

Example:	STM32	L	4Q5	V	G	T	6	TR
Device family								
STM32 = Arm® based 32-bit microcontroller								
Product type								
L = ultra-low-power								
Device subfamily								
4Q5 = STM32L4Q5xx								
Pin count								
C = 48 pins								
R = 64 pins								
V = 100 pins/balls								
Q = 132 balls								
Z = 144 pins								
A = 169 balls								
Flash memory size								
G = 1 Mbyte of Flash memory								
Package								
I = UFBGA (7 x 7 mm)								
T = LQFP								
U = UFQFPN								
Y = WLCSP								
Temperature range								
6 = Industrial temperature range, -40 to 85 °C (105 °C junction)								
3 = Industrial temperature range, -40 to 125 °C (130°C junction)								
Packing								
TR = tape and reel								
xxx = programmed parts								

For a list of available options (speed, package, etc.) or for further information on any aspect of this device, please contact your nearest ST sales office.

9 Revision history

Table 128. Document revision history

Date	Revision	Changes
19-Dec-2019	1	Initial release.
10-Mar-2020	2	<p>Updated 'DAC1&2' in 'DAC1' in:</p> <ul style="list-style-type: none"> – Figure 1: STM32L4Q5xx block diagram. – Table 2: STM32L4Q5xx features and peripheral counts. – Table 4: STM32L4Q5xx modes overview. – Table 5: Functionalities depending on the working mode. – Table 6: STM32L4Q5xx peripherals interconnect matrix. – Table 42: STM32L4Q5xx peripheral current consumption. <p>Updated:</p> <ul style="list-style-type: none"> – Table 14: Legend/abbreviations used in the pinout table adding note. – Table 18: Voltage characteristics V_{REF+} data. <p>Updated current consumptions in:</p> <ul style="list-style-type: none"> – Table 25: Current consumption in Run and Low-power run modes, code with data processing running from Flash in single Bank, ART enable (Cache ON Prefetch OFF). – Table 26: Current consumption in Run and Low-power run modes, code with data processing running from Flash in dual bank, ART enable (Cache ON Prefetch OFF). – Table 27: Current consumption in Run and Low-power run modes, code with data processing running from Flash in single bank, ART disable. – Table 28: Current consumption in Run and Low-power run modes, code with data processing running from Flash in dual bank, ART disable. – Table 29: Current consumption in Run and Low-power run modes, code with data processing running from SRAM1. – Table 30: Typical current consumption in Run and Low-power run modes, with different codes running from Flash, ART enable (Cache ON Prefetch OFF). – Table 33: Current consumption in Sleep and Low-power sleep mode, Flash ON. – Table 34: Current consumption in Low-power sleep mode, Flash in power-down. – Table 35: Current consumption in Stop 2 mode, SRAM3 disabled. – Table 36: Current consumption in Stop 2 mode, SRAM3 enabled. – Table 37: Current consumption in Stop 1 mode. – Table 38: Current consumption in Stop 0 mode. – Table 39: Current consumption in Standby mode. – Table 40: Current consumption in Shutdown mode.

Table 128. Document revision history (continued)

Date	Revision	Changes
10-Mar-2020	2 (continued)	Updated: <ul style="list-style-type: none"> – Table 43: Low-power mode wakeup timings wakeup times. – Table 50: HSI16 oscillator characteristics. – Table 51: MSI oscillator characteristics. – Table 52: HSI48 oscillator characteristics. – Table 55: Flash memory characteristics I_{DD}. – Table 59: ESD absolute maximum ratings versus packages. – Table 67: Analog switches booster characteristics. – Table 76: VREFBUF characteristics V_{REFBUF_OUT}. – Section 7.5: WLCSP100 package information. – Table 126: Package thermal characteristics.

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