

## **Battery Charger IC Series**

# **Boost DC/DC Charger** With Input Current Limiter

BD8664GW BD8665GW BD8668GW

## **General Description**

BD8664GW, BD8665GW and BD8668GW are lithium-ion battery charger IC's, suitable for charging 2S batteries from a 5V source, such as a USB port with DC/DC boost topology.

## Features

- CP/CV Charging
- Charge-On/ Off control available with EN pin
- Integrated Input Detection (VBUSOK)
- Integrated Power Good

**Typical Application Circuit** 

- Boost Switching Topology
- Low Ron integrated MOSFET
- Output Short Circuit Protection
- 0.4mm pitch Chip Scale Package (UCSP75M2)

## Applications

DVC, DSC, MID and other Lithium battery-powered portable devices

## **Key Specifications**

■ Input Current Accuracy ±2%(BD8664GW)

- ±3%(BD8665GW/BD8668GW)
- Charging Voltage Accuracy ±0.5%
- Selectable Input Current
  - 100mA/500mA/900mA/1500mA (max)
  - Charging frequency 1MHz (typ)
- Input Standby Current 71µA(typ)
- battery leakage current while charging is off 0µA(typ)

## Package

UCSP75M2

W(Typ) x D(Typ) x H(Max)

2.20mm x 2.20mm x 0.85mm

## Line Up

Charge	Deelvere	Pin	Orderable
Voltage	Package	number	Part Number
8.30V			BD8664GW
0.401/	USCP75M2	20	BD8665GW
8.40V		25	BD8668GW

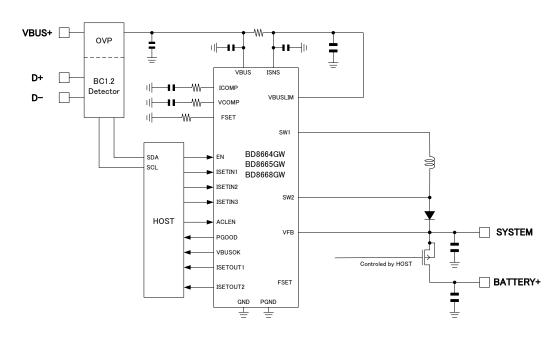


Figure 1. Typical Application

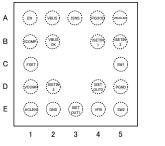
OProduct structure : Silicon monolithic integrated circuit OThis product has no designed protection against radioactive rays

## Contents

General Description1
Features1
Applications1
Key Specifications1
Package W(Typ) x D(Typ) x H(Max)1
Line Up1
Typical Application Circuit1
Pin Configuration (TOP VIEW)
Pin Description3
Block Diagram5
Absolute Maximum Ratings (Ta=25°C)6
Recommended Operating Conditions (Ta=-30 to +85°C)6
Electrical Characteristics
Typical Performance Curves9
Typical Performance Curves
Reference Data14
Reference Data

## Pin Configuration (TOP VIEW)

## BD8664GW / BD8665GW



	BD8668GW							
A	VBUS	VBUS	ISNS	PGOOD	VBUSLIM			
в	(ICOMP)		ISETIN 3	VBUS OK	SW1			
с	FSET	ISETIN 1	ISETIN 2	ISET OUT2	PGND			
D	GND	GND	VCOMPI	VFB	SW2			
E	ACLEN	GND	GND	VFB	ISET OUT1			
	1	2	3	4	5			

## **Pin Description**

## BD8664GW / BD8665GW

000040	W / BD86650	300	
No.	Name	I/O	Description
A1	EN	I	Charging ON/OFF
A2	VBUS	I	Power input
A3	ISNS	I	Current sensing
A4	PGOOD	0	Power GOOD output
A5	VBUSLIM	0	VBUS current limiter output
B1	ICOMP	0	Pin for phase compensation of constant current
B2	VBUSOK	0	VBUSOK output
B3	-	-	-
B4	ISETIN1	Ι	Current setting pin1
B5	ISETIN3	Ι	Current setting pin3
C1	FSET	Ι	Frequency setting pin
C2	-	-	-
C3	-	-	-
C4	-	-	-
C5	SW1	0	Inductor connection pin1
D1	VCOMP	0	Pin for phase compensation of constant current connection
D2	ISETIN2	I	Current setting pin2
D3	-	-	-
D4	ISETOUT2	0	Current setting output2
D5	PGND	I	Power GND(0.0V)
E1	ACLEN	I	Automatic current ON/OFF selection pin
E2	GND	I	GND(0.0V)
E3	ISETOUT1	0	Current setting output 1
E4	VFB	I	Feedback pin of CV charging voltage
E5	SW2	0	Inductor connection pin2

No upper ESD protection diodes are connected to ISETIN1, ISETIN2, ISETIN3, and EN.

## **Pin Description – continued**

## BD8668GW

No.	Name	I/O	Description
A1	VBUS	I	Power input
A2	VBUS	I	Power input
A3	ISNS	I	Current sensing
A4	PGOOD	0	Power GOOD output
A5	VBUSLIM	0	VBUS current limiter output
B1	ICOMP	0	Pin for phase compensation of constant current
B2	EN	I	Charging ON/OFF
B3	ISETIN3	I	Current setting pin3
B4	VBUSOK	0	VBUSOK output
B5	SW1	0	Inductor connection pin1
C1	FSET	I	Frequency setting pin
C2	ISETIN1	I	Current setting pin1
C3	ISETIN2	I	Current setting pin2
C4	ISETOUT2	0	Current setting output2
C5	PGND	I	Power GND (0.0V)
D1	GND	I	GND (0.0V)
D2	GND	I	GND (0.0V)
D3	VCOMP	0	Pin for phase compensation of constant current connection
D4	VFB	I	Feedback pin of CV charging voltage
D5	SW2	0	Inductor connection pin2
E1	ACLEN	I	Automatic current ON/OFF selection pin
E2	GND	I	GND (0.0V)
E3	GND	I	GND (0.0V)
E4	VFB	I	Feedback pin of CV charging voltage
E5	ISETOUT1	0	Current setting output 1
Nou	nnor ESD protocti	on diados ara conne	cted to ISETIN1. ISETIN2. ISETIN3. and EN.

No upper ESD protection diodes are connected to ISETIN1, ISETIN2, ISETIN3, and EN.

## **Block Diagram**

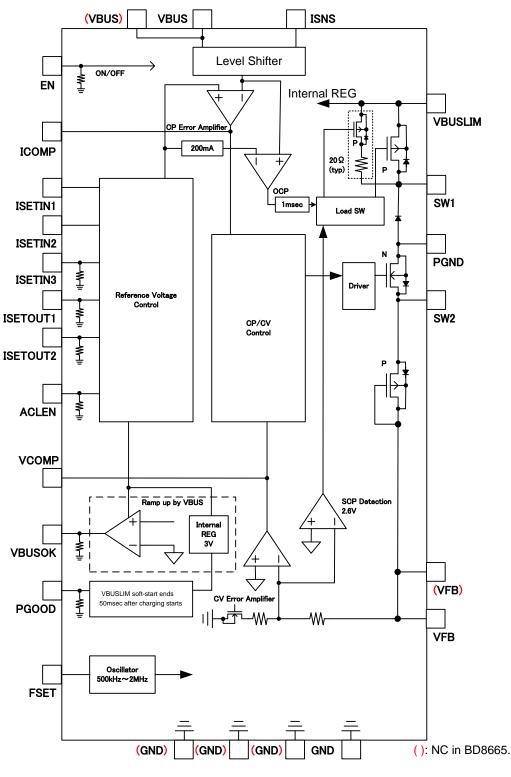


Figure 2. Block Diagram

## Absolute Maximum Ratings (Ta=25°C)

Item	Symbol	Range	Unit
VBUS Voltage	VVBUS	-0.3 to +7.0	V
VBUSLIM Voltage	Vvbuslim	-0.3 to VBUS+0.3 (Note 3)	V
VFB Voltage	V <sub>VFB</sub>	-0.3 to +13.0	V
SW1 Voltage	V <sub>SW1</sub>	-0.3 to VBUSLIM+0.3 (Note 4)	V
SW2 Voltage	V <sub>SW2</sub>	-0.3 to VFB+0.3	V
Terminal Voltage 1 (Note 1)	V <sub>INOUT1</sub>	-0.3 to VBUS+0.3 (Note 3)	V
Terminal Voltage 2 (Note 2)	VINOUT2	-0.3 to +6.0	V
Voltage Between Terminals (Note 5)	Vinout3	-0.3 to +0.3	V
Maximum Power Dissipation (Note 6)	Pd	1.00	W
Operating Temperature	Topr	-30 to +85	°C
Storage Temperature	Tstg	-55 to +150	°C
Junction Temperature	Tjmax	+150	°C

(Note 1) ISNS, FSET, VBUSOK, PGOOD, VCOMP, ICOMP, ISETOUT1, ISETOUT2

(Note 2) ACLEN, EN, ISETIN1, ISETIN2, ISETIN3

(Note 3) 7.0V against GND

(Note 4) 7.0V against PGND

(Note 5) GND-PGND, VBUS-ISNS

(Note 6) When mounted on 54mm x 62mm PCB. Pd decreases by 8mW per 1°C when Ta is 25°C or higher.

Caution: Operating the IC over the absolute maximum ratings may damage the IC. The damage can either be a short circuit between pins or an open circuit between pins and the internal circuitry. Therefore, it is important to consider circuit protection measures, such as adding a fuse, in case the IC is operated over the absolute maximum ratings.

## Recommended Operating Conditions (Ta=-30 to +85°C)

Item	Item Symbol		Value			Part No.	
nem	Gymbol	Min	Тур	Max	Unit	Tarrio.	
VBUS Voltage	V <sub>VBUS</sub>	4.1	5.0	5.5	V	-	
		B Voltage V <sub>VFB</sub> 0	0.0	8.4	10.0	V	BD8665GW/BD8668GW
VFB Voltage	V VFB	0.0	8.3	10.0	V	BD8664GW	

## **Electrical Characteristics**

			Value			- /
Item	Symbol	Min	Тур	Max	Unit	Conditions
VBUS Stand-by Current	IVBUS1	-	71	142	μA	V <sub>EN</sub> = 0.0V, Only VBUSOK is ON
VBUS Operational Current	IVBUS2	-	2	5	mA	No Switching
Battery Stand-by Current	I <sub>BATT1</sub>	-1	0	+1	μA	V <sub>EN</sub> = 0.0V
Battery Operational Current	IBATT2	-	60	120	μA	No Switching
Frequency 1	Fosc1	0.9	1.0	1.1	MHz	R <sub>FSET</sub> = 47kΩ
Frequency 2	F <sub>OSC2</sub>	(1.8)	2.0	(2.2)	MHz	$R_{FSET} = 22k\Omega$
FSET Output Voltage	VFSET	-	0.6	-	V	
<constant block="" control="" voltage=""></constant>				J	J	L
Constant Voltage Charging	V <sub>CV2</sub>	8.258	8.300	8.342	V	±0.5%, BD8664GW
Accuracy	V <sub>CV2</sub>	8.358	8.400	8.442	V	±0.5%, BD8665GW/BD8668GW
< VBUSLIM Current Control Block						· · · · ·
	IVBUSLIM1	6.0	8.0	10.0	mV	$V_{\text{ISETIN1}} = 0.0V,  V_{\text{ISETIN2}} = 0.0V$
	IVBUSLIM2	47.0	48.5	50.0	mV	VISETIN1 = 0.0V, VISETIN2 = 3.3V
		85.2	87.0	88.8	mV	VISETIN1 = 3.3V, VISETIN2 = 0.0V BD8664GW, ±2%
VBUSLIM Current Accuracy (VBUS-ISNS Voltage)	IVBUSLIM3	84.0	07.0	90.0	mV	$ V_{ISETIN1} = 3.3V,  V_{ISETIN2} = 0.0V \\ BD8665GW/BD8668GW, \pm 3\% $
	I <sub>VBUSLIM4</sub>	142.1	145.0	147.9	mV	VISETIN1 = 3.3V, VISETIN2 = 3.3V BD8664GW, ±2%
	TVBUSLIM4	140.0	140.0	150.0	mV	VISETIN1 = 3.3V, VISETIN2 = 3.3V BD8665GW/BD8668GW, ±3%
VBUSLIM Current Limiter Level (VBUS-ISNS Voltage)	IVBUSOCP	(+15)	+20	(+30)	mV	Against The Current Set By VBUSLIM
< PGOOD Block>						
PGOOD H Voltage	Vpgoodh	2.94	3.00	3.06	V	±2%
PGOOD L Voltage	Vpgoodl	-	0.0	-	V	
<vbusok pin=""></vbusok>	-					
VBUS Threshold 1	VVBUSOKTH1	3.9	4.0	4.1	V	$V_{BUS} = L$ to H
VBUS Threshold 2	VVBUSOKTH2	3.8	3.9	4.0	V	V <sub>BUS</sub> = H to L
VBUSOK L->H Delay Time	VVBUSOKDELAY	20	40	-	ms	
VBUSOK H Voltage	VVBUSOKH	2.94	3.00	3.06	V	±2%
VBUSOK L Voltage	VVBUSOKL	-	0.0	-	V	
<comparator block=""></comparator>	•					
VBUS UVLO Threshold	VVBUSUVLOON	3.40	3.60	3.80	V	
VBUS UVLO Unlock Threshold Voltage	VVBUSUVLOFF	3.50	3.70	3.90	V	
VFB Low Voltage Detection1	V <sub>VFBLV1</sub>	-	2.4	-	V	VFB = H to L
VFB Low Voltage Detection2	VVFBLV2	-	2.6	-	V	VFB = L to H
VFB Overvoltage Detection1	VVFBOV1	8.5	9.0	9.5	V	VFB = H to L
VFB Overvoltage Detection2	Vvfbov2	9.2	9.7	10.2	V	VFB = L to H
VBUS Current Automatic Selection Voltage 1	VVBUSAUTOTH	3.9	4.0	4.1	V	VBUS = H to L
VBUS Current Automatic Selection Voltage 1	Vvbusautoth	4.0	4.1	4.2	V	VBUS = L to H
<power mosfet=""></power>						·
VBUSLIM-SW1 PMOS Ron	Rsw1A	-	70	-	mΩ	I <sub>SW1</sub> = -10mA
VBUSLIM-SW1small PMOS Ron	Rsw1B	-	20	-	Ω	I <sub>SW1</sub> = -10mA
SW2-PGND NMOS Ron	R <sub>SW2</sub>	-	60	-	mΩ	$I_{SW2} = 10 \text{mA}$
SW2 Max Duty Width	TSW2MAXDUTY	-	80	-	ns	R <sub>FSET</sub> = 47kΩ
			00			

## **Electrical Characteristics - continued**

<input output=""/>						
EN/ISETIN1,2,3/ACLEN	VINL			0.4	V	
L Voltage	VINL	-	-	0.4	v	
EN/ISETIN1,2,3/ACLEN	V <sub>INH</sub>	2.5		5.5	V	
H Voltage	VINH	2.5	-	5.5	v	
ISETIN1,2 Input Current	linh	-	0	-	μA	No Pull-Down, Open Is Not Allowed.
EN/ACLEN/ISETIN3	RIN2	300	500	700	kΩ	
Pull-Down Resistor	RIN2	300	500	700	K12	
ISETOUT1,2 L Voltage	Voutl	-	0.0	-	V	
ISETOUT1,2 H Voltage	Vouth	2.94	3.00	3.06	V	

This product has no designed protection against radioactive rays. Pd is the maximum power. Please keep the current to meet power lower than the Pd.

## **Typical Performance Curves**

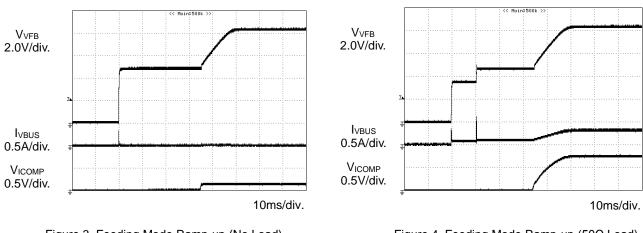
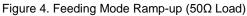


Figure 3. Feeding Mode Ramp-up (No Load)



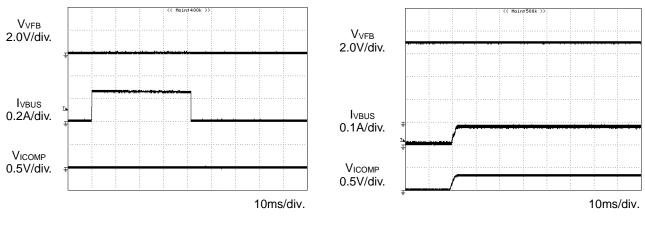
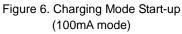
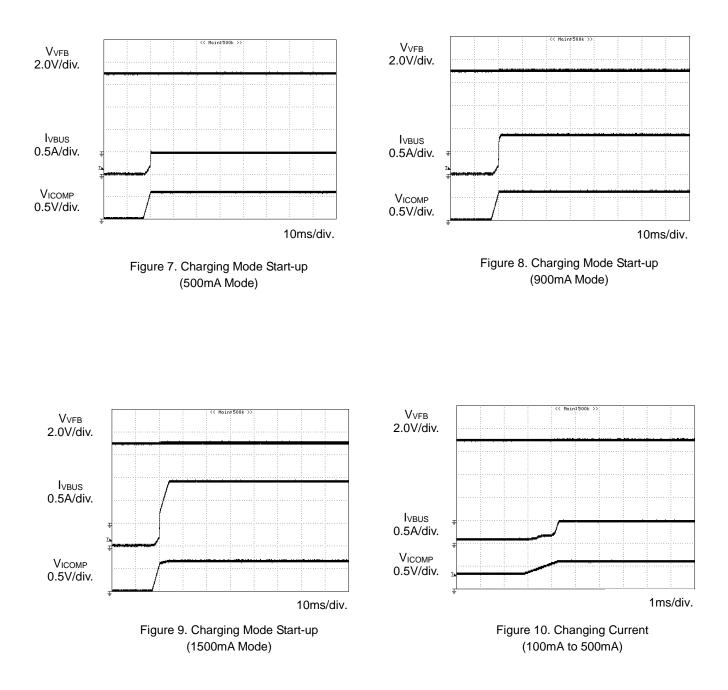
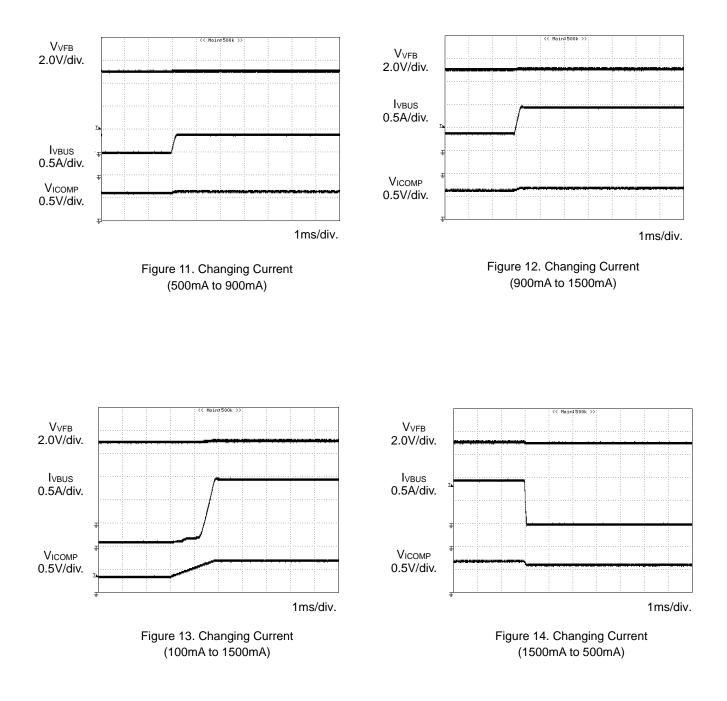
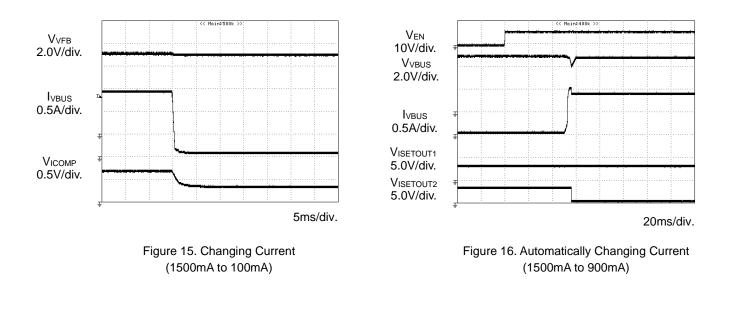


Figure 5. Start-up Waveform (VFB is Shorten to Ground)









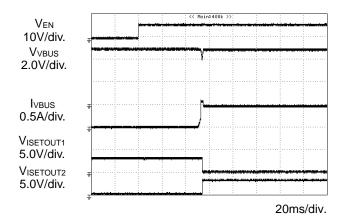


Figure 17. Automatically Changing Current (900mA to 500mA)

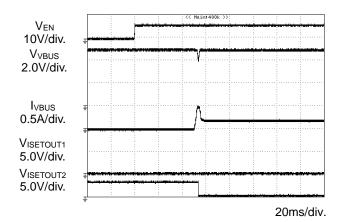


Figure 18. Automatically Changing Current (500mA to 100mA)

(Unless otherwise specified, V<sub>VBUS</sub>=5.0V V<sub>VFB</sub>=7.4V V<sub>ISETIN1,2,3</sub>=0V GND=PGND=0V Ta=25°C)

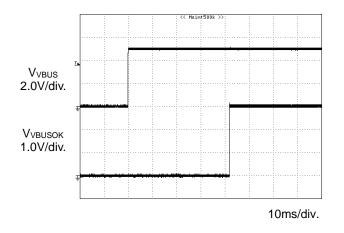
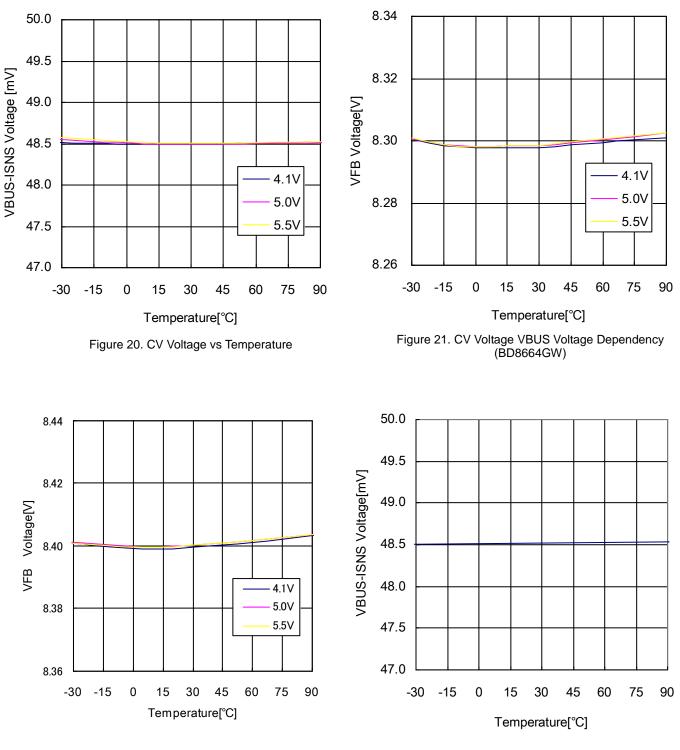


Figure 19. VBUSOK (L to H Delay Time)

## **Reference Data**

(Unless otherwise specified, VvBUS=5.0V VvFB=7.4V VISETIN1,2,3=0V GND=PGND=0V Ta=25°C )



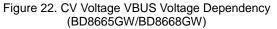
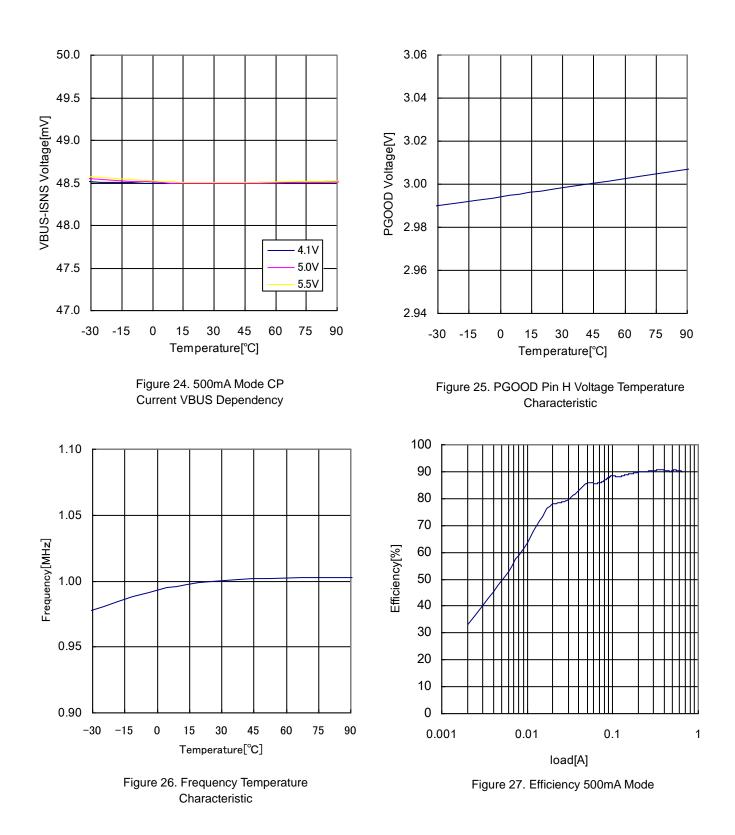


Figure 23. 500mA Mode CP Current Temperature Characteristic

## **Reference Data - continue**



## Block Descriptions

## 1. VBUS Input Detection Comparator Output (VBUSOK)

VBUS voltage can set VBUSOK. 40ms after detecting that VBUS is higher than 4.0V (typ), VBUSOK turns H (3.0V [typ]). In case VBUS ramps down, and reaches 3.9V (typ), VBUSOK turns L without any delay. The function works independently from the condition of EN and thermal shut down.

VBUS	VBUSOK
< 3.9V (typ)	L 0.0V (typ)
>= 4.0V (typ)	H 3.0V (typ)

## 2. Charging ON/OFF (EN)

ON/OFF is controlled with the EN pin. When EN is L, the IC enters shutdown mode (or USB suspend mode), the battery leakage current is set to  $0\mu$ A, and all other functions but VBUSOK turns off. EN pin is connected to a  $500k\Omega$  (typ) pull-down resistor.

## 3. USB Current Setting (ISETIN1,2,3)

ISET1	ISET2	ISET3	VBUS Input Current Settings
L	L	L	100mA (max)
L	L	Н	500mA (max)
L	Н	L	500mA (max)
L	Н	Н	500mA (max)
Н	L	L	900mA (max)
Н	L	Н	900mA (max)
Н	Н	L	1500mA (max)
Н	Н	Н	1500mA (max)

Note: Open is NOT allowed for ISETIN1,2 pins. No pull-down resistor is connected to ISETIN1, 2 pins. 500kΩ (typ) pull-down resistor is internally connected to ISETIN3.

## 4. VBUS Current Setting (ISETOUT1,2)

ISETOUT1	ISETOUT2	VBUS Current Settings
L	L	100mA (max)
L	Н	500mA (max)
Н	L	900mA (max)
Н	Н	1500mA (max)

Note:  $500k\Omega(typ)$  is connected internally to ISETOUT1,2. Even in the case VBUS=0V, the output stays stable.

## 5. Frequency Setting (FSET)

The PWM switching frequency can be set.

FSET	PWM
100kΩ	500kHz (typ)
47kΩ	1MHz (typ)
33kΩ	1.5MHz (typ)
22kΩ	2MHz (typ)

## 6. CV Control Soft-Start

If the system boots up with NO battery, CV control method suppresses the 8.3V (typ) in case of using BD8664GW and 8.4V (typ) in case of using BD8665GW/BD8668GW, on VFB pin, and enters "feeding mode". In this mode, it will take 40ms (typ) for the VFB to reach 8.4V (typ).

## 7. Load Switch Function

A PMOS load switch is integrated between VBUSLIM and SW1. When EN=L, the load switch turns off. If a low battery is connected, charging can be stopped. The integrated load resistors are  $20\Omega$  (typ) and  $70m\Omega$  (typ). The higher resistance is connected during start-up. After 10ms (typ), the lower resistance is connected if no short circuit is detected by VFB pin.

## 8. OCP for Load Switch

Through a sense resistor between VBUS and ISNS, over-current can be detected while the load switch is on. If the over-current is constantly detected for more than 1ms, the load switch turns off and latches, and PGOOD is set L. To unlatch, the IC must be rebooted by switching EN to low, then back to high; or set VBUS to a voltage lower than UVLO, then back to the operating VBUS voltage.

## 9. Battery Low Voltage Detection (Output SCP)

If during start-up, the battery's voltage is lower than VFB voltage or the output is shorted to ground, the low-side load switch will never be turned on, and the high-side load switch will be on for 80ms (typ). But, if the high-side load switch is kept on for more than 80ms, the load switch is turned off. This function is off after the PGOOD is turned H.

### 10. Power Good (PGOOD)

The IC is enabled by EN pin. After CV, CP, and soft start, PGOOD condition changes its state from L to H. Inversely, during thermal shutdown, overvoltage battery and OCP, PGOOD is L.

### 11. Battery Overvoltage Detection

Due to the VBUS current limiter, overvoltage can occur at VFB terminal during CP charging. This can cause damage to devices that are connected to the IC. To prevent this, overvoltage protection is integrated. Once overvoltage is detected, SW2 becomes Hi-Z, the error amp output and soft start are reset to default, and PGOOD is set L. Once VFB voltage is at a safe level, the IC automatically restarts with soft start.

### 12. Auto VBUS Current Setting

Once VBUS voltage exceeds 4.1V (typ), the VBUS current set to ISET1 to 3 pins are automatically changed from 1500mA (max) to 900mA (max), from 900mA (max) to 500mA (max), and from 500mA (max) to 100mA (max), while VBUS is continuously monitored. If ISET1 to 3 are changed after the auto change is done, the initial current that has been set to ISET1 to 3 will be employed again.

Notes:

- (1) If VBUS voltage remains lower than 4.1V and current is changed from 1500mA to 900mA, the current will not be changed to 500mA.
- (2) ACLEN has a pull-up resistor. The pin is L if it is open and auto setting becomes active.
- (3) It can be turned off by setting ACLEN to H.
- (4) It starts to work after the lower resistance load is turned on. Until PGOOD is changed to H, the bus current value is determined only by ISETIN1 to 3 and will not to be changed by VFB low voltage function.

### 13. Feeding Mode and Charging Mode

Feeding Mode: If the system boots up with NO battery to the IC, CV control method suppresses 8.3V (typ) for BD8664GW and 8.4V (typ) for BD8665GW/BD8668GW on VFB pin. During the feeding mode, the constant voltage is done by the VFB pin.

Charging Mode: If the system boots up with a battery to the IC, CP/CV control method is employed. During charging mode, the CV/CP function is applied to the battery. The two modes, however, are not internally controllable by the IC. An application that applies constant output voltage with CV charging is called "feeding mode" and another application that charges with CV/ CP charging is called "charging mode" in this technical note.

## **Timing Chart/Application Information**

1. VBUSOK/VBUS Threshold at Automatic Change Current Setting

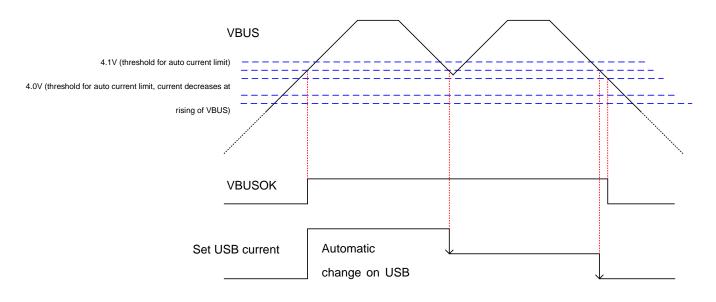
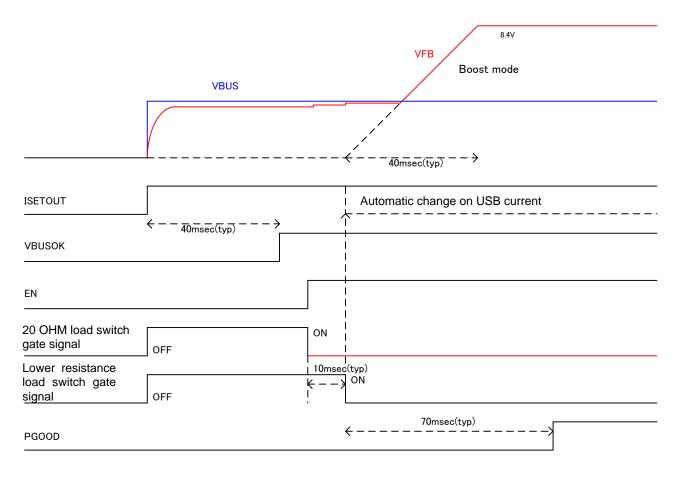
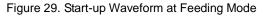


Figure 28. VBUSOK/VBUS Threshold at Automatic Change Current Setting waveform

## 1. Start-up Waveform at Feeding Mode (No Battery, Light Load)





## 2. Start-up Waveform at Feeding Mode (No Battery, $50\Omega$ Load)

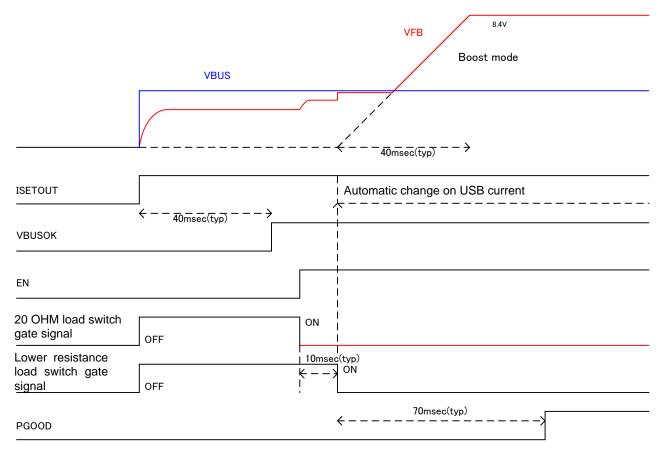


Figure 30. Start-up Waveform at Feeding Mode (No Battery, 50Ω Load)

## 3. Start-up Waveform at Feeding Mode(No battery, Heavy Load [Short to Ground])

	VBUS				
				Vout	
ISETOUT					
VBUSOK	<→ 40msec(typ)>				
EN					
20 OHM load switch gate signal	OFF	ON <del>(</del>	80msec(typ)		OFF
Lower resistance load switch gate signal	OFF				
PGOOD					

Figure 31. Start-up Waveform at Feeding Mode (No battery, Heavy Load [Short to Ground])

## 4. Voltage Waveform to PGOOD at Charging Mode (With Battery)

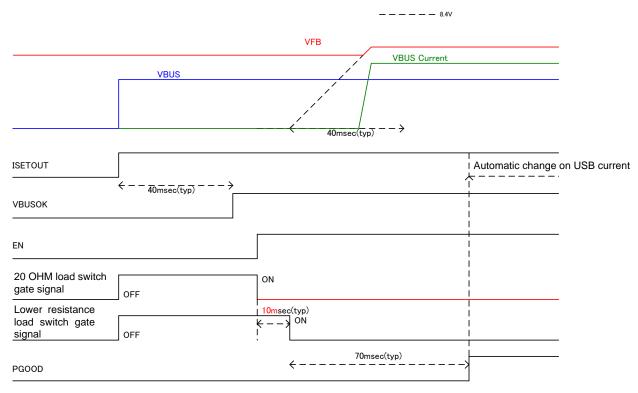


Figure 32. Voltage Waveform to PGOOD at Charging Mode (With Battery)

### 5. Operation from Feeding Mode to Charging Mode

During feeding mode, after the output started up with 8.3V(typ) for BD8664GW and 8.4V(typ) for BD8665GW/BD8668GW, if the battery has to be connected and the mode has to change to charging mode, set EN to L then H to enable CP charging. This turns PGOOD pin to L then H. Note that VBUS current may exceed the set value unless the EN is set L once.

### 6. Operation from Charging Mode To Feeding Mode

During charging mode, if the mode has to change to feeding mode, set the EN to L, detach the battery, then set EN to H again. This turns PGOOD pin to L, ramps up VFB to 8.3V (typ) for BD8664GW and 8.4V (typ) for BD8665GW/BD8668GW by feeding mode, and turns PGOOD to H, afterwards. Note that the overcurrent protection may occur unless the EN is set L like aforementioned VFB overvoltage detection waveform.

### 7. Battery Overvoltage Detection Waveform

During charging, if the battery is detached by a user, V<sub>OUT</sub> will go higher as the mode changes to feeding mode. In this scenario, to prevent damage to devices connected to this IC, OVP is integrated. PGOOD has to be turned off to L when OVP is detected. Soft-start is again implemented when V<sub>OUT</sub> goes low due to its output load.

In the application circuit example, note that the VFB node goes down to VBUS -1Vf, as determined by an external schottky diode.

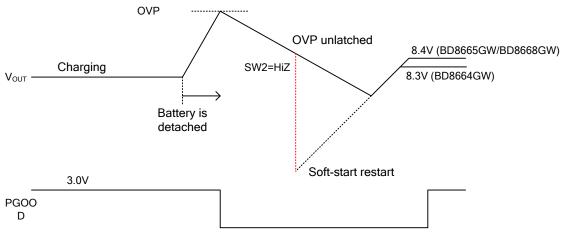


Figure 33. Battery Overvoltage Detection Waveform

### Precaution on Voltage Application between Constant-Voltage Charging Voltage and OVP 8.

When the voltage between constant-voltage charging voltage and OVP is applied to the VFB node, (e.g., An AC adapter is unplugged when the AC adapter voltage is applied to the VFB pin), the VFB terminal drops drastically, so avoid the above mentioned condition.

### **Application Components Selection** 1. Frequency Setting (FSET) Resistor

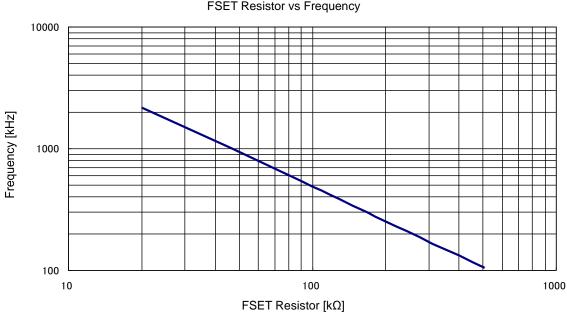


Figure 34.. Frequency Setting (FSET) Resistor

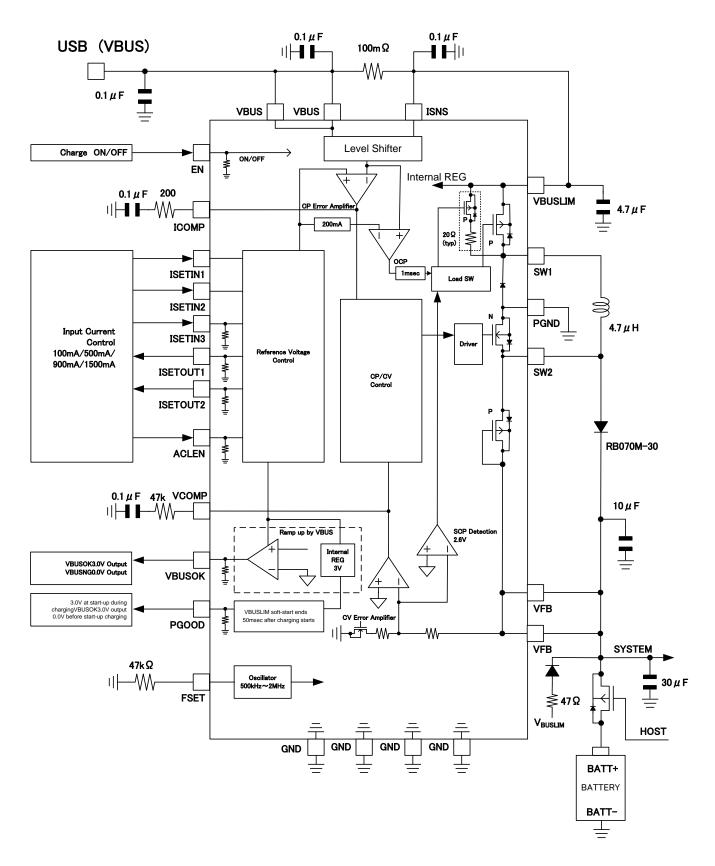
### Inductor Selection 2.

Inductance for the boost switching affects its ripple current and ripple current at feeding mode. The ripple voltage is inversely proportional to the inductance and switching frequency so that the inductance must be higher if the frequency is lower. In other words, the inductance can be smaller if the frequency is higher. However, if the inductance changes, since the LC cutoff frequency changes, the phase compensation of ICOMP and VCOMP may have to be changed.

PWM	Inductance	Output Capacitance	ICOMP Time Constant	VCOMP Coefficient
1MHz(typ)	4.7µH	40µF	200Ω, 0.1µF serial	47kΩ, 0.1µF serial
			1 1 1 4 1	

% If the external coefficient is changed from the designated value above, check the open-loop gain phase carefully.

## **Example of Recommended Circuit**





Pin No. (BD8668GW)	Pin Name	Pin Immediate Circuit	Function
A1/A2	VBUS		USB power input
A5	VBUSLIM		USB current limiter
D1/D2	GND		Ground
/E2/E3			
C5	PGND		Power Ground
		GND (sub)	
A3	ISNS	VBUS	Current detection amp input
A4 B4 C4 E5	PGOOD VBUSOK ISETOUT2 ISETOUT1	3V REF	Logic output (with pull-down resistor)
B1	ICOMP		Error amp output
D3	VCOMP		
B2 E1 B3	EN ACLEN ISETIN3		Logic input (with pull-down resistor)

## Input/Output Pin Immediate Circuit

Pin No.	Pin Name	Pin Immediate Circuit	Function
(BD8668GW)			
C2 C3	ISETIN1 ISETIN2	VBUSLIM	Logic input (without pull-down resistor)
B5	SW1	WBUSLIM	Load switch output Inductor connection1
C1	FSET	VBUSLIM	Frequency setting resistor terminal
D4 E4	VFB VFB	VFB	CV charging voltage feed-back terminal
D5	SW2	VFB	Boost switching terminal Inductor connection 2

## Input/Output Pin Immediate Circuit - continued

## **Thermal Reduction Characteristics**

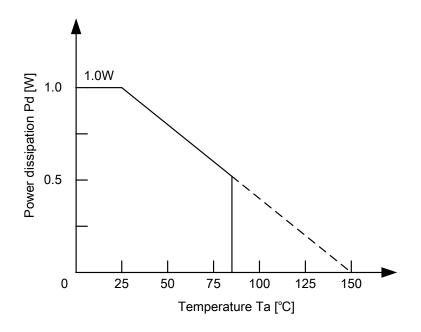


Figure 36.. Power Dissipation (Mounted on a 4-layer substrate board)

## **Operational Notes**

### 1. Reverse-Connection of Power Supply Connecter

Connecting the power supply in reverse polarity can damage the IC. Take precautions against reverse polarity when connecting the power supply, such as mounting an external diode between the power supply and the IC's power supply terminals.

## 2. Power Supply Line

Design the PCB layout pattern to provide low impedance supply lines. Separate the ground and supply lines of the digital and analog blocks to prevent noise in the ground and supply lines of the digital block from affecting the analog block. Furthermore, connect a capacitor to ground at all power supply pins. Consider the effect of temperature and aging on the capacitance value when using electrolytic capacitors.

### 3. Ground Potential

Ensure that no pins are at a voltage below that of the ground pin at any time, even during transient condition.

### 4. Ground Wiring Pattern

When using both small-signal and large-current ground traces, the two ground traces should be routed separately but connected to a single ground at the reference point of the application board to avoid fluctuations in the small-signal ground caused by large current. Also ensure that the ground traces of external components do not cause variation on the ground voltage. The ground lines must be as short and thick as possible to reduce line impedance.

## 5. Thermal Consideration

Should by any chance the power dissipation rating be exceeded, the rise in temperature of the chip may result in deterioration of the properties of the chip. The absolute maximum rating of the Pd stated in this specification is when the IC is mounted on a 70mm x 70mm x 1.6mm glass epoxy board. In case of exceeding this absolute maximum rating, increase the board size and copper area to prevent exceeding the Pd rating.

## 6. Recommended Operating Conditions

These conditions represent a range within which the expected characteristics of the IC can be approximately obtained. The electrical characteristics are guaranteed under the conditions of each parameter.

### 7. Inrush Current

When power is first supplied to the IC, it is possible that the internal logic may be unstable and inrush current may flow instantaneously due to the internal powering sequence and delays, especially if the IC has more than one power supply. Therefore, give special consideration to power coupling capacitance, power wiring, width of ground wiring, and routing of connections.

## 8. Operation Under Strong Electromagnetic Field

Operating the IC in the presence of a strong electromagnetic field may cause the IC to malfunction.

## 9. Testing on Application Boards

When testing the IC on an application board, connecting a capacitor directly to a low-impedance output pin may subject the IC to stress. Always discharge capacitors completely after each process or step. The IC's power supply should always be turned off completely before connecting or removing it from the test setup during the inspection process. To prevent damage from static discharge, ground the IC during assembly and use similar precautions during transport and storage.

### 10. Inter-pin Short and Mounting Errors

Ensure that the direction and position are correct when mounting the IC on the PCB. Incorrect mounting may result in damaging the IC. Avoid nearby pins being shorted to each other especially to ground, power supply and output pin. Inter-pin shorts could be due to many reasons such as metal particles, water droplets (in very humid environment) and unintentional solder bridge deposited in between pins during assembly to name a few.

## 11. Unused Input Terminals

Input terminals of an IC are often connected to the gate of a MOS transistor. The gate has extremely high impedance and extremely low capacitance. If left unconnected, the electric field from the outside can easily charge it. The small charge acquired in this way Is enough to produce a significant effect on the conduction through the transistor and cause unexpected operation of the IC. So, unless otherwise specified, unused input terminals should be connected to the power supply of ground line.

## **Operational Notes – continued**

### 12. Regarding the Input Pins of the IC

This monolithic IC contains P+ isolation and P substrate layers between adjacent elements in order to keep them isolated. P-N junctions are formed at the intersection of the P layers with the N layers of other elements, creating a parasitic diode or transistor. For example (refer to figure below):

When GND > Pin A and GND > Pin B, the P-N junction operates as a parasitic diode. When GND > Pin B, the P-N junction operates as a parasitic transistor.

Parasitic diodes inevitably occur in the structure of the IC. The operation of parasitic diodes can result in mutual interference among circuits, operational faults, or physical damage. Therefore, conditions that cause these diodes to operate, such as applying a voltage lower than the GND voltage to an input pin (and thus to the P substrate) should be avoided.

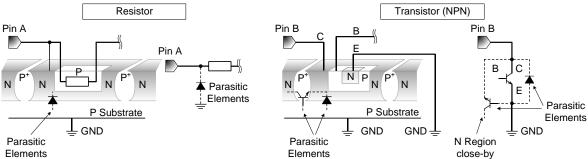


Figure 37. Example of monolithic IC structure

## 13. Thermal Shutdown Circuit (TSD)

This IC has a built-in thermal shutdown circuit that prevents heat damage to the IC. Normal operation should always be within the IC's power dissipation rating. If however the rating is exceeded for a continued period, the junction temperature (Tj) will rise which will activate the TSD circuit that will turn OFF all output pins. When the Tj falls below the TSD threshold, the circuits are automatically restored to normal operation.

Note that the TSD circuit operates in a situation that exceeds the absolute maximum ratings and therefore, under no circumstances, should the TSD circuit be used in a set design or for any purpose other than protecting the IC from heat damage.

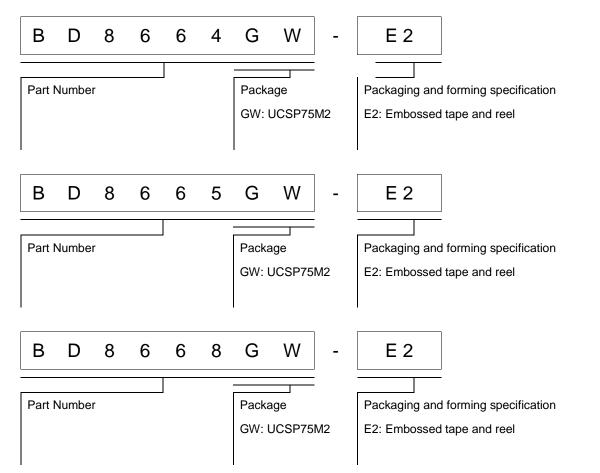
## 14. Thermal Consideration

Use a thermal design that allows for a sufficient margin by taking into account the permissible power dissipation (Pd) in actual operating conditions.

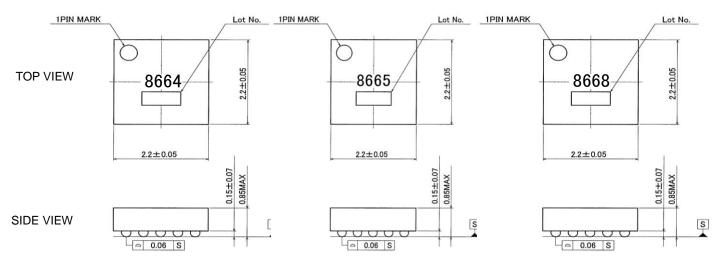
## 15. Capacitor between Logic Output and GND

The logic outputs are VBUSOK, PGOOD, ISETOUT1 and ISETOUT2. With a large capacitor connected between logic output and GND, it is possible that the logic output will short to 0V or GND and will cause the current from the capacitor to flow into the logic output, causing damage to IC. The capacitor between logic output and GND must be 0.1µF or less.

## **Ordering Information**



## **Marking Diagrams**



Part Number Marking	Package	Orderable Part Number	
BD8664GW	UCSP75M2	BD8664GW-E2	
BD8665GW	UCSP75M2	BD8665GW-E2	
BD8668GW	UCSP75M2	BD8668GW-E2	

## Datasheet

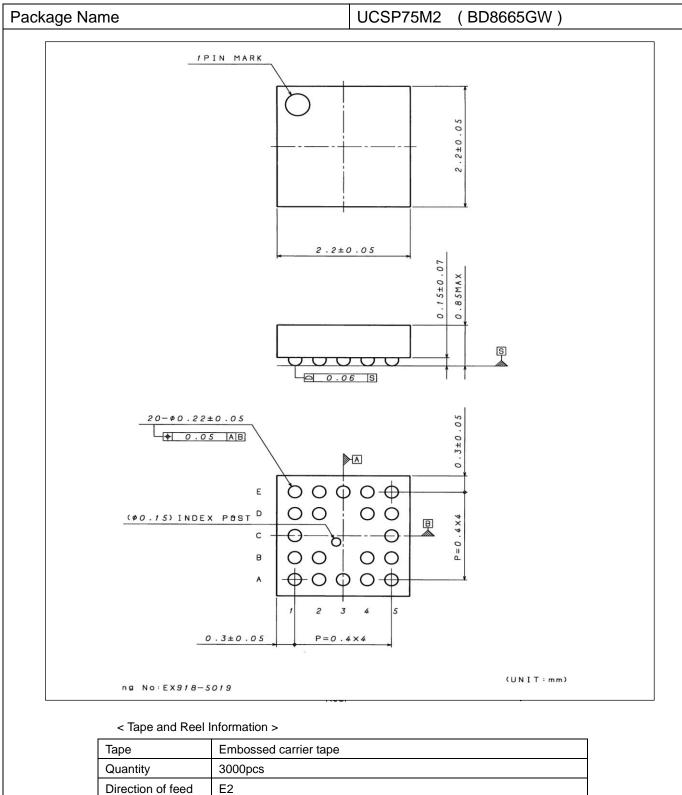
### Package Name UCSP75M2 (BD8664GW) 1 PIN MARK 2±0.05 2.2±0.05 0.15±0.07 0.85MAX S 000 0.06 S 20-\$0.22±0.05 3±0.05 + 0.05 AB A 0 Е D Ο $\odot$ Ο O O Ο O (\$0.15) INDEX POST D 4×4. 贝 С Ð ห P = 0. Ο в Ο Ο $\bigcirc$ 00 $\oplus$ $O \oplus$ A 2 3 4 0.3±0.05 $P=0.4\times4$ (UNIT:mm) ng No: EX918-5019 < Tape and Reel Information > Embossed carrier tape Tape Quantity 3000pcs Direction of feed E2 The direction is the pin 1 of product is at the upper left when you hold reel on the left hand and you pull out the tape on the right hand 0 0 0 0 0 0 0 0 0 0 0 0 1234 1234 1234 1234 1234 1234

## Physical Dimensions, Tape and Reel information

Reel

Mark PIN1 Direction of feed

## Physical Dimensions, Tape and Reel information



Mark PIN1

0

1234

0 0 0 0

1234

Reel

0

123

0 0 0

0

12:

The direction is the pin 1 of product is at the upper left when you hold reel on the left hand and you pull out the tape on the right hand

1234

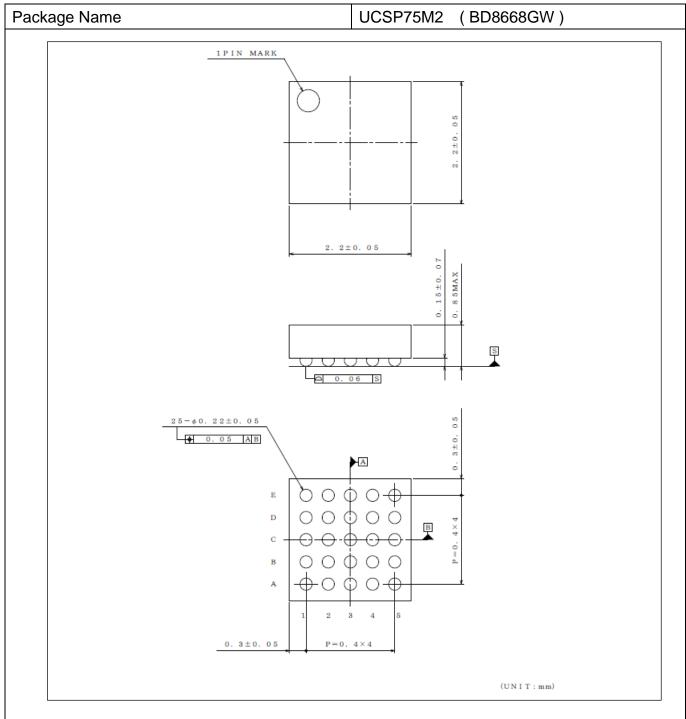
0

Direction of feed

0

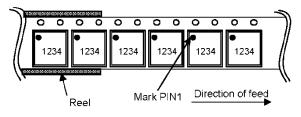
1234

## Physical Dimensions, Tape and Reel information



## < Tape and Reel Information >

Таре	Embossed carrier tape	
Quantity	3000pcs	
Direction of feed	E2	
	The direction is the pin 1 of product is at the upper left when you	
	hold reel on the left hand and you pull out the tape on the right hand	



## Revision History

Date	Revision	Changes	
17.Dec.2014	001	New Release	
7.Jul.2015	002	Condition of OCP is added in PGOOD function explanation.(pp16-17)	

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CLASSⅢ	CLASSⅢ	CLASS II b	
CLASSⅣ	CLASSIII	CLASSⅢ	CLASSII

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For details, please refer to ROHM Mounting specification

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  - [d] the Products are exposed to high Electrostatic
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